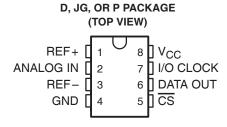
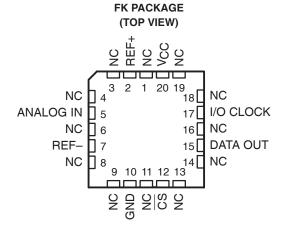
- 3.3-V Supply Operation
- 10-Bit-Resolution Analog-to-Digital Converter (ADC)
- Inherent Sample and Hold Function
- Total Unadjusted Error . . . ±1 LSB Max
- On-Chip System Clock
- Terminal Compatible With TLC1549 and TLC1549x
- Application Report Available[†]
- CMOS Technology

description

The TLV1549C, TLV1549I, and TLV1549M are 10-bit, switched-capacitor, successive-approximation, analog-to-digital converters. The devices have two digital inputs and a 3-state output [chip select (CS), input-output clock (I/O CLOCK), and data output (DATA OUT)] that provide a three-wire interface to the serial port of a host processor.

The sample-and-hold function is automatic. The converter incorporated in the device features differential high-impedance reference inputs that facilitate ratiometric conversion, scaling, and isolation of analog circuitry from logic and supply noise. A switched-capacitor design allows low-error conversion over the full operating free-air temperature range.





NC - No internal connection

The TLV1549C is characterized for operation from 0° C to 70° C. The TLV1549I is characterized for operation from -40° C to 85° C. The TLV1549M is characterized for operation over the full military temperature range of -55° C to 125° C.

AVAILABLE OPTIONS

		PACKA	(GE	
TA	SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	TLV1549CD	_	_	TLV1549CP
-40°C to 85°C	TLV1549ID	_	_	TLV1549IP
-55°C to 125°C	_	TLV1549MFK	TLV1549MJG	_



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

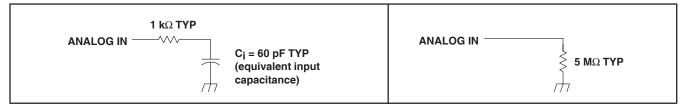
† Interfacing the TLV1549 10-Bit Serial-Out ADC to Popular 3.3-V Microcontrollers (SLAA005)



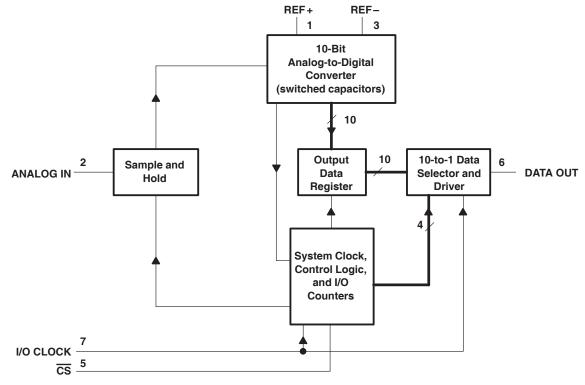
typical equivalent inputs

INPUT CIRCUIT IMPEDANCE DURING SAMPLING MODE

INPUT CIRCUIT IMPEDANCE DURING HOLD MODE



functional block diagram



Terminal numbers shown are for the D, JG, and P packages only.

Terminal Functions

TERMINA	۱L	1/0	DESCRIPTION
NAME	NO.	1/0	DESCRIPTION
ANALOG IN	2	Ι	Analog input. The driving source impedance should be \leq 1 k Ω . The external driving source to ANALOG IN should have a current capability \geq 10 mA.
<u>cs</u>	5	Ι	Chip select. A high-to-low transition on $\overline{\text{CS}}$ resets the internal counters and controls and enables DATA OUT and I/O CLOCK within a maximum of a setup time plus two falling edges of the internal system clock. A low-to-high transition disables I/O CLOCK within a setup time plus two falling edges of the internal system clock.
DATA OUT	6	0	This 3-state serial output for the A/D conversion result is in the high-impedance state when \overline{CS} is high and active when \overline{CS} is low. With a valid chip select, DATA OUT is removed from the high-impedance state and is driven to the logic level corresponding to the MSB value of the previous conversion result. The next falling edge of I/O CLOCK drives DATA OUT to the logic level corresponding to the next most significant bit, and the remaining bits are shifted out in order with the LSB appearing on the ninth falling edge of I/O CLOCK. On the tenth falling edge of I/O CLOCK, DATA OUT is driven to a low logic level so that serial interface data transfers of more than ten clocks produce zeroes as the unused LSBs.
GND	4	I	The ground return for internal circuitry. Unless otherwise noted, all voltage measurements are with respect to GND.
I/O CLOCK	7	I	The input/output clock receives the serial I/O CLOCK input and performs the following three functions: 1) On the third falling edge of I/O CLOCK, the analog input voltage begins charging the capacitor array and continues to do so until the tenth falling edge of I/O CLOCK. 2) It shifts the nine remaining bits of the previous conversion data out on DATA OUT. 3) It transfers control of the conversion to the internal state controller on the falling edge of the tenth clock.
REF+	1	I	The upper reference voltage value (nominally V_{CC}) is applied to REF+. The maximum input voltage range is determined by the difference between the voltage applied to REF+ and the voltage applied to REF
REF-	3	I	The lower reference voltage value (nominally ground) is applied to this REF
VCC	8	Ī	Positive supply voltage

detailed description

With chip select (\overline{CS}) inactive (high), the I/O CLOCK input is initially disabled and DATA OUT is in the high-impedance state. When the serial interface takes \overline{CS} active (low), the conversion sequence begins with the enabling of I/O CLOCK and the removal of DATA OUT from the high-impedance state. The serial interface then provides the I/O CLOCK sequence to I/O CLOCK and receives the previous conversion result from DATA OUT. I/O CLOCK receives an input sequence that is between 10 and 16 clocks long from the host serial interface. The first ten I/O clocks provide the control timing for sampling the analog input.

There are six basic serial interface timing modes that can be used with the TLV1549. These modes are determined by the speed of I/O CLOCK and the operation of \overline{CS} as shown in Table 1. These modes are: (1) a fast mode with a 10-clock transfer and \overline{CS} inactive (high) between transfers, (2) a fast mode with a 10-clock transfer and \overline{CS} active (low) continuously, (3) a fast mode with an 11- to 16-clock transfer and \overline{CS} inactive (high) between transfers, (4) a fast mode with a 16-bit transfer and \overline{CS} active (low) continuously, (5) a slow mode with an 11- to 16-clock transfer and \overline{CS} inactive (high) between transfers, and (6) a slow mode with a 16-clock transfer and \overline{CS} active (low) continuously.

The MSB of the previous conversion appears on DATA OUT on the falling edge of \overline{CS} in mode 1, mode 3, and mode 5, within 21 μs from the falling edge of the tenth I/O CLOCK in mode 2 and mode 4, and following the 16th clock falling edge in mode 6. The remaining nine bits are shifted out on the next nine falling edges of the I/O CLOCK. Ten bits of data are transmitted to the host serial interface through DATA OUT. The number of serial clock pulses used also depends on the mode of operation, but a minimum of ten clock pulses is required for conversion to begin. On the tenth clock falling edge, the internal logic takes DATA OUT low to ensure that the remaining bit values are zero if the I/O CLOCK transfer is more than ten clocks long.

Table 1 lists the operational modes with respect to the state of \overline{CS} , the number of I/O serial transfer clocks that can be used, and the timing on which the MSB of the previous conversion appears at the output.



SLAS071C - JANUARY 1993 - REVISED MARCH 1995

Table 1. Mode Operatio	Table 1	. Mode	Operation
------------------------	---------	--------	-----------

MODES	6	cs	NO. OF I/O CLOCKS	MSB AT DATA OUT	TIMING DIAGRAM
	Mode 1	High between conversion cycles	10	CS falling edge	Figure 6
Fast Modes	Mode 2	Low continuously	nuously 10 Withi		Figure 7
rasi Modes	Mode 3	High between conversion cycles	11 to 16 [‡]	CS falling edge	Figure 8
	Mode 4	Low continuously	16‡	Within 21 μs	Figure 9
Slow Modes	Mode 5	High between conversion cycles	11 to 16 [‡]	CS falling edge	Figure 10
Slow Modes	Mode 6	Low continuously	16 [‡]	16th clock falling edge	Figure 11

[†] This timing also initiates serial-interface communication.

All the modes require a minimum period of 21 μs after the falling edge of the tenth I/O CLOCK before a new transfer sequence can begin. During a serial I/O CLOCK data transfer, \overline{CS} must be active (low) so that the I/O CLOCK input is enabled. When \overline{CS} is toggled between data transfers (modes 1, 3, and 5), the transitions at \overline{CS} are recognized as valid only if the level is maintained for a minimum period of 1.425 μs after the transition. If the transfer is more than ten I/O clocks (modes 3, 4, 5, and 6), the rising edge of the eleventh clock must occur within 9.5 μs after the falling edge of the tenth I/O CLOCK; otherwise, the device could lose synchronization with the host serial interface and \overline{CS} has to be toggled to restore proper operation.

fast modes

The device is in a fast mode when the serial I/O CLOCK data transfer is completed within 21 μ s from the falling edge of the tenth I/O CLOCK. With a 10-clock serial transfer, the device can only run in a fast mode.

mode 1: fast mode, CS inactive (high) between transfers, 10-clock transfer

In this mode, \overline{CS} is inactive (high) between serial I/O-CLOCK transfers and each transfer is ten clocks long. The falling edge of \overline{CS} begins the sequence by removing DATA OUT from the high-impedance state. The rising edge of \overline{CS} ends the sequence by returning DATA OUT to the high-impedance state within the specified delay time. Also, the rising edge of \overline{CS} disables I/O CLOCK within a setup time plus two falling edges of the internal system clock.

mode 2: fast mode, CS active (low) continuously, 10-clock transfer

In this mode, \overline{CS} is active (low) between serial I/O-CLOCK transfers and each transfer is ten clocks long. After the initial conversion cycle, \overline{CS} is held active (low) for subsequent conversions. Within 21 μ s after the falling edge of the tenth I/O CLOCK, the MSB of the previous conversion appears at DATA OUT.

mode 3: fast mode, CS inactive (high) between transfers, 11- to 16-clock transfer

In this mode, \overline{CS} is inactive (high) between serial I/O-CLOCK transfers and each transfer can be 11 to 16 clocks long. The falling edge of \overline{CS} begins the sequence by removing DATA OUT from the high-impedance state. The rising edge of \overline{CS} ends the sequence by returning DATA OUT to the high-impedance state within the specified delay time. Also, the rising edge of \overline{CS} disables I/O CLOCK within a setup time plus two falling edges of the internal system clock.

mode 4: fast mode, $\overline{\text{CS}}$ active (low) continuously, 16-clock transfer

In this mode, \overline{CS} is active (low) between serial I/O-CLOCK transfers and each transfer must be exactly 16 clocks long. After the initial conversion cycle, \overline{CS} is held active (low) for subsequent conversions. Within 21 μ s after the falling edge of the tenth I/O CLOCK, the MSB of the previous conversion appears at DATA OUT.

slow modes

In a slow mode, the serial I/O CLOCK data transfer is completed after 21 μs from the falling edge of the tenth I/O CLOCK.



[‡] No more than 16 clocks should be used.

SLAS071C - JANUARY 1993 - REVISED MARCH 1995

mode 5: slow mode, CS inactive (high) between transfers, 11- to 16-clock transfer

In this mode, \overline{CS} is inactive (high) between serial I/O-CLOCK transfers and each transfer can be 11 to 16 clocks long. The falling edge of \overline{CS} begins the sequence by removing DATA OUT from the high-impedance state. The rising edge of \overline{CS} ends the sequence by returning DATA OUT to the high-impedance state within the specified delay time. Also, the rising edge of \overline{CS} disables I/O CLOCK within a setup time plus two falling edges of the internal system clock.

mode 6: slow mode, CS active (low) continuously, 16-clock transfer

In this mode, \overline{CS} is active (low) between serial I/O-CLOCK transfers and each transfer must be exactly 16 clocks long. After the initial conversion cycle, \overline{CS} is held active (low) for subsequent conversions. The falling edge of the sixteenth I/O CLOCK then begins each sequence by removing DATA OUT from the low state, allowing the MSB of the previous conversion to appear immediately at DATA OUT. The device is then ready for the next 16-clock transfer initiated by the serial interface.

analog input sampling

Sampling of the analog input starts on the falling edge of the third I/O CLOCK, and sampling continues for seven I/O CLOCK periods. The sample is held on the falling edge of the tenth I/O CLOCK.

converter and analog input

The CMOS threshold detector in the successive-approximation conversion system determines each bit by examining the charge on a series of binary-weighted capacitors (see Figure 1). In the first phase of the conversion process, the analog input is sampled by closing the S_C switch and all S_T switches simultaneously. This action charges all the capacitors to the input voltage.

In the next phase of the conversion process, all S_T and S_C switches are opened and the threshold detector begins identifying bits by identifying the charge (voltage) on each capacitor relative to the reference (REF–) voltage. In the switching sequence, ten capacitors are examined separately until all ten bits are identified and then the charge-convert sequence is repeated. In the first step of the conversion phase, the threshold detector looks at the first capacitor (weight = 512). Node 512 of this capacitor is switched to the REF+ voltage, and the equivalent nodes of all the other capacitors on the ladder are switched to REF–. If the voltage at the summing node is greater than the trip point of the threshold detector (approximately one-half V_{CC}), a bit 0 is placed in the output register and the 512-weight capacitor is switched to REF–. If the voltage at the summing node is less than the trip point of the threshold detector, a bit 1 is placed in the register and this 512-weight capacitor remains connected to REF+ through the remainder of the successive-approximation process. The process is repeated for the 256-weight capacitor, the 128-weight capacitor, and so forth down the line until all bits are determined.

With each step of the successive-approximation process, the initial charge is redistributed among the capacitors. The conversion process relies on charge redistribution to determine the bits from MSB to LSB.

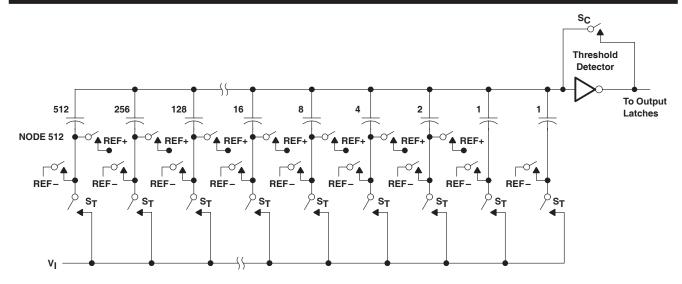


Figure 1. Simplified Model of the Successive-Approximation System

chip-select operation

The trailing edge of \overline{CS} starts all modes of operation, and \overline{CS} can abort a conversion sequence in any mode. A high-to-low transition on \overline{CS} within the specified time during an ongoing cycle aborts the cycle, and the device returns to the initial state (the contents of the output data register remain at the previous conversion result). Exercise care to prevent \overline{CS} from being taken low close to completion of conversion because the output data may be corrupted.

reference voltage inputs

There are two reference inputs used with the TLV1549: REF+ and REF-. These voltage values establish the upper and lower limits of the analog input to produce a full-scale and zero reading, respectively. The values of REF+, REF-, and the analog input should not exceed the positive supply or be lower than GND consistent with the specified absolute maximum ratings. The digital output is at full scale when the input signal is equal to or higher than REF+ and at zero when the input signal is equal to or lower than REF-.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

TLV1549C	
TLV1549I	
TLV1549M	0.5 V to 6 V
	0.3 V to V _{CC} + 0.3 V
	0.3 V to V _{CC} + 0.3 V
	V _{CC} + 0.1 V
	–0.1 V
	±20 mA
	±30 mA
TLV1549C	0°C to 70°C
TLV1549I	–40°C to 85°C
TLV1549M	–55°C to 125°C
	–65°C to 150°C
	260°C
	TLV1549I TLV1549M

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are with respect to ground with REF - and GND wired together (unless otherwise noted).



SLAS071C - JANUARY 1993 - REVISED MARCH 1995

recommended operating conditions

			MIN	NOM	MAX	UNIT
Supply voltage, V _{CC}			3	3.3	3.6	V
Positive reference voltage, V _{ref+} (see Note	2)			Vcc		V
Negative reference voltage, V _{ref} (see Note	2)			0		V
Differential reference voltage, V _{ref+} - V _{ref-}	(see Note 2)		2.5	Vcc	V _{CC} +0.2	V
Analog input voltage (see Note 2)			(V _{CC}	V
High-level control input voltage, VIH		V _{CC} = 3 V to 3.6 V	2			V
Low-level control input voltage, V _{IL}		V _{CC} = 3 V to 3.6 V			0.6	V
Clock frequency at I/O CLOCK (see Note 3)	(2.1	MHz		
Setup time, CS low before first I/O CLOCK↑,	t _{su(CS)} (see	Note 4)	1.425	i		μs
Hold time, CS low after last I/O CLOCK↓, th(CS)		(ns
Pulse duration, I/O CLOCK high, twH(I/O)			190)		ns
Pulse duration, I/O CLOCK low, t _{wL(I/O)}			190)		ns
Transition time, I/O CLOCK, t _{t(I/O)} (see Note	5 and Figure	5)			1	μs
Transition time, CS, t _{t(CS)}					10	μs
TLV1549			(70	°C
Operating free-air temperature, TA	TLV1549I	TLV1549I			85	°C
	TLV1549M		-55	i	125	°C

- NOTES: 2. Analog input voltages greater than that applied to REF + convert as all ones (1111111111), while input voltages less than that applied to REF convert as all zeros (0000000000). The TLV1549 is functional with reference voltages down to 1 V (V_{ref +} V_{ref -}); however, the electrical specifications are no longer applicable.
 - 3. For 11- to 16-bit transfers, after the tenth I/O CLOCK falling edge (≤ 2 V), at least one I/O CLOCK rising edge (≥ 2 V) must occur within 9.5 us.
 - 4. To minimize errors caused by noise at CS, the internal circuitry waits for a setup time plus two falling edges of the internal system clock after CS↓ before responding to the I/O CLOCK. Therefore, no attempt should be made to clock out the data until the minimum CS setup time has elapsed.
 - 5. This is the time required for the clock input signal to fall from V_{IL}max or to rise from V_{IL}max to V_{IL}max to V_{IL}max. In the vicinity of normal room temperature, the device functions with input clock transition time as slow as 1 μs for remote data-acquisition applications where the sensor and the A/D converter are placed several feet away from the controlling microprocessor.



SLAS071C - JANUARY 1993 - REVISED MARCH 1995

electrical characteristics over recommended operating free-air temperature range, $V_{CC} = V_{ref+} = 3 \text{ V}$ to 3.6 V, I/O CLOCK frequency = 2.1 MHz (unless otherwise noted)

	PARAMETE	R	TEST CONI	DITIONS	MIN	TYP†	MAX	UNIT
Va	High-level output voltage		$V_{CC} = 3 V$,	$I_{OH} = -1.6 \text{ mA}$	2.4			٧
VOH	High-level output voltage		$V_{CC} = 3 \text{ V to } 3.6 \text{ V},$	$I_{OH} = -20 \mu A$	V _{CC} -0.1			٧
V-0.	Low-level output voltage		$V_{CC} = 3 V$,	I _{OL} = 1.6 mA			0.4	V
VOL	Low-level output voltage		$V_{CC} = 3 \text{ V to } 3.6 \text{ V},$	I _{OL} = 20 μA			0.1	٧
	Off-state (high-impedance-si	toto) output ourropt	$V_O = V_{CC}$	CS at V _{CC}			10	
loz	On-state (nigh-impedance-s	iale) output current	$V_{O} = 0$,	CS at V _{CC}			-10	μΑ
lіН	High-level input current		$V_I = V_{CC}$			0.005	2.5	μΑ
I _{IL}	Low-level input current		V _I = 0			-0.005	-2.5	μΑ
Icc	Operating supply current		CS at 0 V			0.4	2.5	mA
	Analas issuelas la sues auros		$V_I = V_{CC}$			1	4	
	Analog input leakage curren	τ	V _I = 0				-1	μΑ
	Maximum static analog refer	rence current into REF+	V _{ref+} = V _{CC} ,	V _{ref} _ = GND			10	μΑ
		TLV1549C, I (Analog)	During sample cycle			30	55	
		TLV1549M, (Analog)	During sample cycle			30		
Ci	Input capacitance	TLV1549C, I (Control)				5	15	pF
		TLV1549M, (Control)				5		

 $[\]dagger$ All typical values are at V_{CC} = 3.3 V, T_A = 25°C.

SLAS071C - JANUARY 1993 - REVISED MARCH 1995

operating characteristics over recommended operating free-air temperature range, $V_{CC} = V_{ref+} = 3 \text{ V}$ to 3.6 V, I/O CLOCK frequency = 2.1 MHz

	PARAMETER	TEST CONDITIONS	MIN MAX	UNIT
	Linearity error (see Note 6)		±1	LSB
	Zero error (see Note 7)	See Note 2	±1	LSB
	Full-scale error (see Note 7)	See Note 2	±1	LSB
	Total unadjusted error (see Note 8)		±1	LSB
tconv	Conversion time	See Figures 6-11	21	μs
t _C	Total cycle time (access, sample, and conversion)	See Figures 6-11 and Note 9	21 +10 I/O CLOCK periods	μs
t _V	Valid time, DATA OUT remains valid after I/O CLOCK \downarrow	See Figure 5	10	ns
td(I/O-DATA)	Delay time, I/O CLOCK↓ to DATA OUT valid	See Figure 5	240	ns
tPZH, tPZL	Enable time, CS↓ to DATA OUT (MSB driven)	See Figure 3	1.3	μs
tPHZ, tPLZ	Disable time, CS↑ to DATA OUT (high impedance)	See Figure 3	180	ns
^t r(bus)	Rise time, data bus	See Figure 5	300	ns
^t f(bus)	Fall time, data bus	See Figure 5	300	ns
td(I/O-CS)	Delay time, 10th I/O CLOCK \downarrow to $\overline{\text{CS}}$ \downarrow to abort conversion (see Note 10)		9	μs

- NOTES: 2. Analog input voltages greater than that applied to REF+ convert as all ones (1111111111), while input voltages less than that applied to REF- convert as all zeros (0000000000). The device is functional with reference voltages down to 1 V (V_{ref+} V_{ref-}); however, the electrical specifications are no longer applicable.
 - 6. Linearity error is the maximum deviation from the best straight line through the A/D transfer characteristics.
 - 7. Zero error is the difference between 0000000000 and the converted output for zero input voltage; full-scale error is the difference between 1111111111 and the converted output for full-scale input voltage.
 - 8. Total unadjusted error comprises linearity, zero, and full-scale errors.
 - 9. I/O CLOCK period = 1/(I/O CLOCK frequency). Sampling begins on the falling edge of the third I/O CLOCK, continues for seven I/O CLOCK periods, and ends on the falling edge of the tenth I/O CLOCK (see Figure 5).
 - 10. Any transitions of \overline{CS} are recognized as valid only if the level is maintained for a minimum of a setup time plus two falling edges of the internal clock (1.425 μs) after the transition.



PARAMETER MEASUREMENT INFORMATION

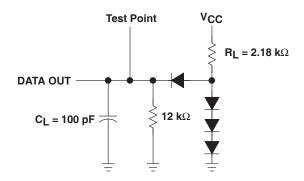


Figure 2. Load Circuit

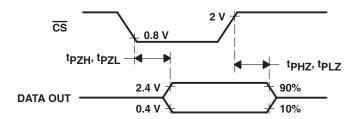


Figure 3. DATA OUT to Hi-Z Voltage Waveforms

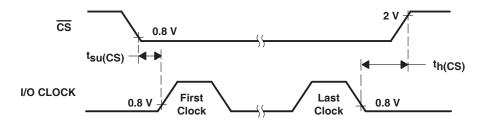


Figure 4. CS to I/O CLOCK Voltage Waveforms

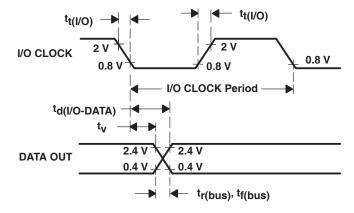


Figure 5. I/O CLOCK and DATA OUT Voltage Waveforms



PARAMETER MEASUREMENT INFORMATION

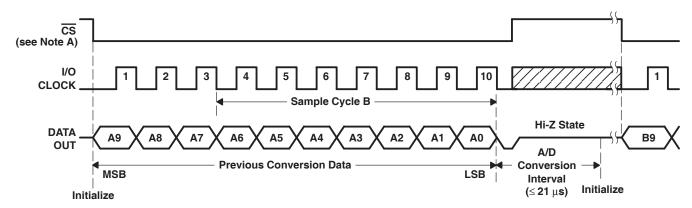


Figure 6. Timing for 10-Clock Transfer Using $\overline{\text{CS}}$

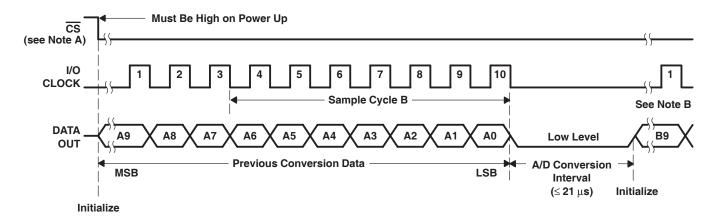


Figure 7. Timing for 10-Clock Transfer Not Using CS

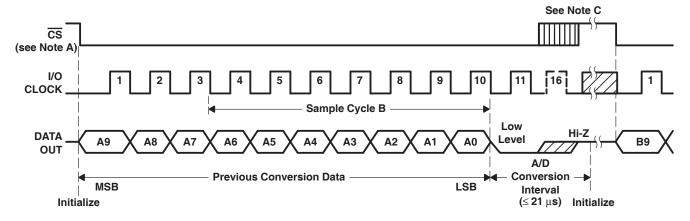


Figure 8. Timing for 11- to 16-Clock Transfer Using CS (Serial Transfer Completed Within 21 μs)

- NOTES: A. To minimize errors caused by noise at \overline{CS} , the internal circuitry waits for a setup time plus two falling edges of the internal system clock after \overline{CS} ↓ before responding to the I/O CLOCK. No attempt should be made to clock out the data until the minimum \overline{CS} setup time has elapsed
 - B. A low-to-high transition of $\overline{\text{CS}}$ disables I/O CLOCK within a maximum of a setup time plus two falling edges of the internal system clock.
 - C. The first I/O CLOCK must occur after the end of the previous conversion.



PARAMETER MEASUREMENT INFORMATION

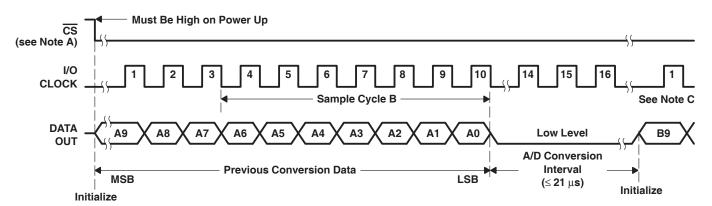


Figure 9. Timing for 16-Clock Transfer Not Using CS (Serial Transfer Completed Within 21 μs)

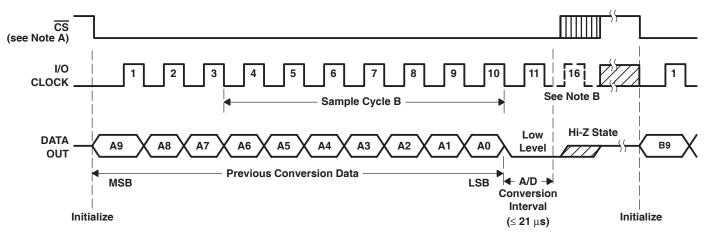


Figure 10. Timing for 11- to 16-Clock Transfer Using CS (Serial Transfer Completed After 21 μs)

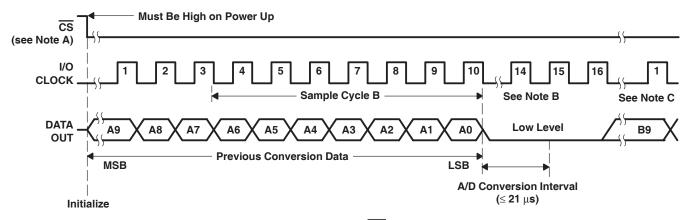
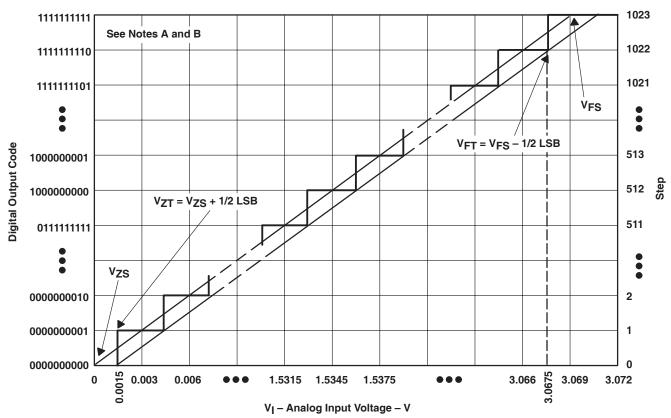


Figure 11. Timing for 16-Clock Transfer Not Using CS (Serial Transfer Completed After 21 μs)

- NOTES: A. To minimize errors caused by noise at \overline{CS} , the internal circuitry waits for a set up time plus two falling edges of the internal system clock after \overline{CS} ↓ before responding to the I/O CLOCK. No attempt should be made to clock out the data until the minimum \overline{CS} setup time has elapsed.
 - B. A low-to-high transition of $\overline{\text{CS}}$ disables I/O CLOCK within a maximum of a setup time plus two falling edges of the internal system clock.
 - C. The first I/O CLOCK must occur after the end of the previous conversion.



APPLICATION INFORMATION



- NOTES: A. This curve is based on the assumption that V_{ref+} and V_{ref-} have been adjusted so that the voltage at the transition from digital 0 to 1 (V_{ZT}) is 0.0015 V and the transition to full scale (V_{FT}) is 3.0675 V. 1 LSB = 3 mV.
 - B. The full-scale value (VFS) is the step whose nominal midstep value has the highest absolute value. The zero-scale value (VZS) is the step whose nominal midstep value equals zero.

Figure 12. Ideal Conversion Characteristics

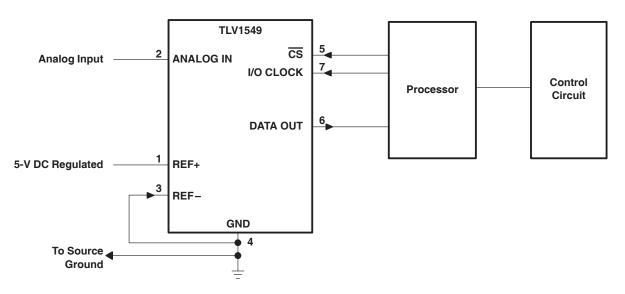


Figure 13. Typical Serial Interface



APPLICATION INFORMATION

simplified analog input analysis

Using the equivalent circuit in Figure 14, the time required to charge the analog input capacitance from 0 to V_S within 1/2 LSB can be derived as follows:

The capacitance charging voltage is given by

$$V_{C} = V_{S} \left(1 - e^{-t_{C}/R_{t}C_{i}} \right)$$
 (1)

where

$$R_t = R_s + r_i$$

The final voltage to 1/2 LSB is given by

$$V_C (1/2 LSB) = V_S - (V_S/2048)$$
 (2)

Equating equation 1 to equation 2 and solving for time t_c gives

$$V_{S} - (V_{S}/2048) = V_{S}(1 - e^{-t_{C}/R_{t}C_{i}})$$
 (3)

and

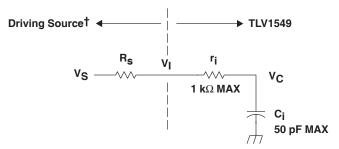
$$t_{c} (1/2 LSB) = R_{t} \times C_{i} \times ln(2048)$$

$$\tag{4}$$

Therefore, with the values given the time for the analog input signal to settle is

$$t_{\rm c} (1/2 \, \text{LSB}) = (R_{\rm S} + 1 \, \text{k}\Omega) \times 60 \, \text{pF} \times \ln(2048)$$
 (5)

This time must be less than the converter sample time shown in the timing diagrams.



V_I = Input Voltage at ANALOG IN

V_S = External Driving Source Voltage

R_S = Source Resistance

 r_i = Input Resistance

C_i = Equivalent Input Capacitance

† Driving source requirements:

- Noise and distortion for the source must be equivalent to the resolution of the converter.
- R_S must be real at the input frequency.

Figure 14. Equivalent Input Circuit Including the Driving Source







www.ti.com 21-Mar-2013

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing		Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
TLV1549CD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		V1549C	Samples
TLV1549CDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		V1549C	Samples
TLV1549CDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		V1549C	Samples
TLV1549CDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		V1549C	Samples
TLV1549CP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type		TLV1549CP	Samples
TLV1549CPE4	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type		TLV1549CP	Samples
TLV1549ID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		V1549I	Samples
TLV1549IDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		V1549I	Samples
TLV1549IP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type		TLV1549IP	Samples
TLV1549IPE4	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type		TLV1549IP	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)



PACKAGE OPTION ADDENDUM

21-Mar-2013

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Only one of markings shown within the brackets will appear on the physical device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

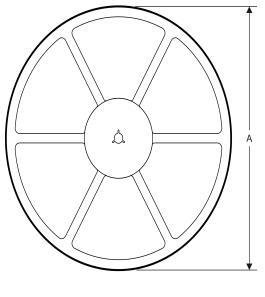
In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

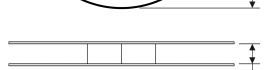
PACKAGE MATERIALS INFORMATION

www.ti.com 14-Jul-2012

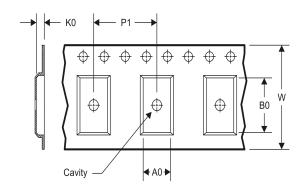
TAPE AND REEL INFORMATION

REEL DIMENSIONS





TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV1549CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

PACKAGE MATERIALS INFORMATION

www.ti.com 14-Jul-2012



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV1549CDR	SOIC	D	8	2500	367.0	367.0	35.0

IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All semiconductor products (also referred to herein as "components") are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its components to the specifications applicable at the time of sale, in accordance with the warranty in TI's terms and conditions of sale of semiconductor products. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by applicable law, testing of all parameters of each component is not necessarily performed.

TI assumes no liability for applications assistance or the design of Buyers' products. Buyers are responsible for their products and applications using TI components. To minimize the risks associated with Buyers' products and applications, Buyers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI components or services are used. Information published by TI regarding third-party products or services does not constitute a license to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of significant portions of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI components or services with statements different from or beyond the parameters stated by TI for that component or service voids all express and any implied warranties for the associated TI component or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyer acknowledges and agrees that it is solely responsible for compliance with all legal, regulatory and safety-related requirements concerning its products, and any use of TI components in its applications, notwithstanding any applications-related information or support that may be provided by TI. Buyer represents and agrees that it has all the necessary expertise to create and implement safeguards which anticipate dangerous consequences of failures, monitor failures and their consequences, lessen the likelihood of failures that might cause harm and take appropriate remedial actions. Buyer will fully indemnify TI and its representatives against any damages arising out of the use of any TI components in safety-critical applications.

In some cases, TI components may be promoted specifically to facilitate safety-related applications. With such components, TI's goal is to help enable customers to design and create their own end-product solutions that meet applicable functional safety standards and requirements. Nonetheless, such components are subject to these terms.

No TI components are authorized for use in FDA Class III (or similar life-critical medical equipment) unless authorized officers of the parties have executed a special agreement specifically governing such use.

Only those TI components which TI has specifically designated as military grade or "enhanced plastic" are designed and intended for use in military/aerospace applications or environments. Buyer acknowledges and agrees that any military or aerospace use of TI components which have *not* been so designated is solely at the Buyer's risk, and that Buyer is solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI has specifically designated certain components as meeting ISO/TS16949 requirements, mainly for automotive use. In any case of use of non-designated products, TI will not be responsible for any failure to meet ISO/TS16949.

Products Applications

Audio www.ti.com/audio Automotive and Transportation www.ti.com/automotive Communications and Telecom **Amplifiers** amplifier.ti.com www.ti.com/communications **Data Converters** dataconverter.ti.com Computers and Peripherals www.ti.com/computers **DLP® Products** www.dlp.com Consumer Electronics www.ti.com/consumer-apps

DSP **Energy and Lighting** dsp.ti.com www.ti.com/energy Clocks and Timers www.ti.com/clocks Industrial www.ti.com/industrial Interface interface.ti.com Medical www.ti.com/medical logic.ti.com Logic Security www.ti.com/security

Power Mgmt power.ti.com Space, Avionics and Defense www.ti.com/space-avionics-defense

Microcontrollers microcontroller.ti.com Video and Imaging www.ti.com/video

RFID www.ti-rfid.com

OMAP Applications Processors www.ti.com/omap TI E2E Community e2e.ti.com

Wireless Connectivity <u>www.ti.com/wirelessconnectivity</u>