TMS320F28069F, TMS320F28068F, TMS320F28062F InstaSPIN[™]-FOC Software

Technical Reference Manual



Literature Number: SPRUHI9 February 2013



Contents

1	F2806	xF InstaSPIN™-FOC Enabled MCUs	4
2	FAST	Estimator Features	5
3	InstaS	PIN™-FOC Solution Features	5
4	InstaS	PIN-FOC Block Diagrams	6
5	Comp	aring FAST Estimator to Typical Solutions	8
6	FAST	Provides Sensorless FOC Performance	9
	6.1	FAST Estimator Replaces Mechanical Sensor	. 9
	6.2	Rotor Angle Accuracy Critical for Performance	11
	6.3	Phase Currents Key to Estimator Accuracy	11
7	Evalua	ating FAST and InstaSPIN-FOC Performance	12
8	Micro	controller Resources	12
	8.1	Memory Allocation and Utilization	19
	8.2	Pin Utilization	21
Арр	endix A	Definition of Terms and Acronyms	23

2



List of Figures

1	FAST - Estimating Flux, Angle, Speed, Torque - Automatic Motor Identification	4
2	Block diagram of Entire InstaSPIN-FOC Package in ROM	6
3	Block diagram of InstaSPIN-FOC in User Memory, with Exception of FAST in ROM	7
4	Sensored FOC System	10
5	Inverter Using the 3-Shunt Current Sampling Technique	12
6	Functional Block Diagram	15
7	Peripheral Blocks	16
8	Software Execution Clock Tree Provides Flexibility with Real-Time Scheduling	17
9	28069 Memory Map	20
10	2806xF Allocated Memory for InstaSPIN-FOC Library	21

List of Tables

1	FAST Estimator Compared to Typical Solutions	8
2	Hardware Features	13
3	CPU Cycles for FULL Implementation Executing from ROM and FLASH	17
4	CPU loading for FULL Implementation Executing from ROM and FLASH	18
5	CPU loading for FULL Implementation Executing from ROM and FLASH	19
6	2806xF Allocated Memory for InstaSPIN-FOC Library	21
7	User Memory and Stack Sizes	21
8	Pin Utilization Per Motor	22

3



TMS320F28069F, TMS320F28068F, TMS320F28062F InstaSPIN[™]-FOC Software

1 F2806xF InstaSPIN[™]-FOC Enabled MCUs

TMS320F2806xF are the first devices (69F, 68F, and 62F - 80 or 100 pin packages) from Texas Instruments that include the FAST[™] (Figure 1) estimator and additional motor control functions needed for cascaded speed and torque loops for efficient three-phase field-oriented motor control (FOC).

Together – with F2806xF peripheral drivers in user code – they enable a sensorless (also known as selfsensing) InstaSPIN-FOC solution which can identify, tune the torque controller and efficiently control your motor in minutes, without the use of any mechanical rotor sensors. This entire package is called InstaSPIN-FOC, which is made available in ROM. The user also has the option of executing all FOC functions in user memory (FLASH or RAM), which makes calls to the proprietary FAST estimator firmware in ROM. InstaSPIN-FOC was designed for flexibility, to accommodate a range of system software architectures. The range of this flexibility is shown in Figure 1 and Figure 2.

This document is a supplement to all standard TMS320F2806x documentation, including the standard device data sheet [*TMS320F2806x Piccolo Microcontrollers* (literature number <u>SPRS698</u>)], technical reference manual, and user's guides. An additional document included with the InstaSPIN-FOC documentation package is the *TMS320F2806xF InstaSPIN-FOC/InstaSPIN-MOTION User's Guide* (literature number <u>SPRUHJ1</u>), which covers the scope and functionality of:

- F2806xF devices
- F2806xF ROM contents
- FAST flux estimator

4

• InstaSPIN-FOC system solutions.

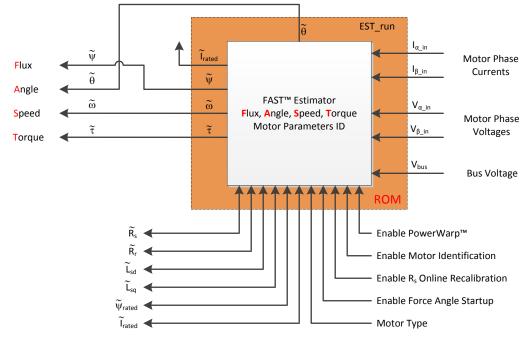


Figure 1. FAST - Estimating Flux, Angle, Speed, Torque - Automatic Motor Identification



FAST Estimator Features

2 FAST Estimator Features

- Unified observer structure which exploits the similarities between all motors that use magnetic flux for energy transduction
 - Both synchronous (BLDC, SPM, IPM), and asynchronous (ACIM) control are possible
 - Salient compensation for Interior Permanent Magnet motors: observer tracks rotor flux and angle correctly when Ls-d and Ls-q are provided
- Unique, high quality motor feedback signals for use in control systems
 - High-quality Flux signal for stable flux monitoring and field weakening
 - Superior rotor flux Angle estimation accuracy over wider speed range compared to traditional observer techniques independent of all rotor parameters for ACIM
 - Real-time low-noise motor shaft Speed signal
 - Accurate high bandwidth Torque signal for load monitoring and imbalance detection
- Angle estimator converges within first cycle of the applied waveform, regardless of speed
- Stable operation in all power quadrants, including generator quadrants
- Accurate angle estimation at steady state speeds below 1 Hz (typ) with full torque
- Angle integrity maintained even during slow speed reversals through zero speed
- Angle integrity maintained during stall conditions, enabling smooth stall recovery
- Motor Identification measures required electrical motor parameters of unloaded motor in under 2 minutes (typ)
- "On-the-fly" stator resistance recalibration (online Rs) tracks stator resistance changes in real time, resulting in robust operation over temperature. This feature can also be used as a temperature sensor of the motor's windings (basepoint calibration required)
- · Superior transient response of rotor flux angle tracking compared to traditional observers
- PowerWarp[™] adaptively reduces current consumption to minimize the combined (rotor and stator) copper losses to the lowest, without compromising ACIM output power levels

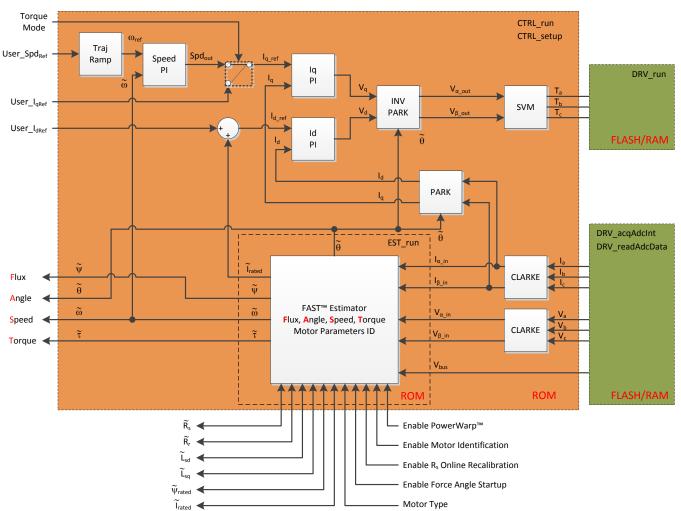
3 InstaSPIN[™]-FOC Solution Features

- Includes the Flux Angle Speed Torque (FAST) estimator, used to measure rotor flux (both magnitude and angle) in a sensorless field-oriented control (FOC) system
- Automatic torque (current) loop tuning, with option for user adjustments
- Automatic speed loop tuning provides stable operation for most applications. (Better transient response can be obtained by optimizing parameters for a particular application)
- Automatic or manual field weakening and field boosting
- Bus Voltage compensation
- Automatic offset calibration insures quality samples of feedback signals

5



InstaSPIN-FOC Block Diagrams



4 InstaSPIN-FOC Block Diagrams

Figure 2. Block diagram of Entire InstaSPIN-FOC Package in ROM



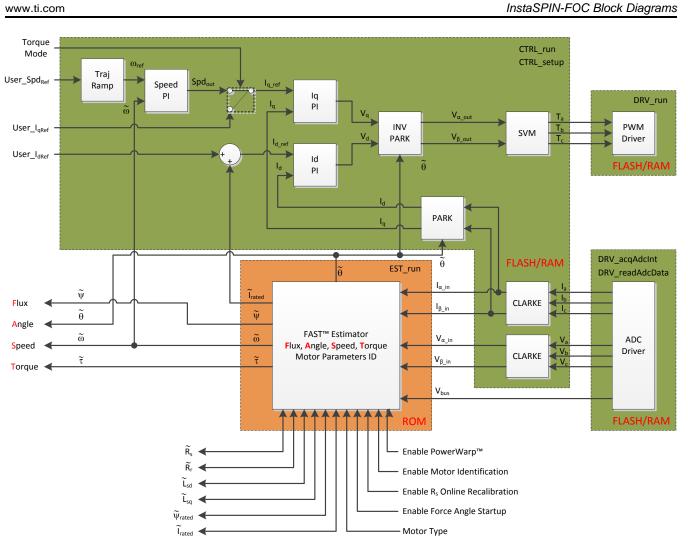


Figure 3. Block diagram of InstaSPIN-FOC in User Memory, with Exception of FAST in ROM



5 Comparing FAST Estimator to Typical Solutions

Table 1 shows a comparison of the FAST estimator and InstaSPIN-FOC solution to typical software sensors and FOC solutions.

Торіс	Typical Software Sensors and FOC Solutions	Fast Estimator and InstaSPIN-FOC Solution
Electrical Motor	Motor-model based observers heavily dependent on	Relies on fewer motor parameters.
Parameters	motor parameters.	Off-line parameter identification of motor – no data sheet required.
		On-line parameter monitoring and re-estimation of stator resistance.
Estimator Tuning	Complex observer tuning, done multiple times for speed/loads, for each motor.	No estimator tuning required. Once motor parameters are identified, it works the same way every time, across speed/torque dynamics.
Estimator Accuracy	Angle-tracking performance is typically only good at over 5-10Hz with challenges at higher speeds and compensation for field weakening. Dynamic performance influenced by hand tuning of observer; Motor stalls typically crash observer.	FAST provides reliable angle tracking which converges within one electrical cycle of the applied waveform, and can track at less than 1 Hz frequency (dependent on quality and resolution of analog sensing).
		Angle tracking exhibits excellent transient response (even with sudden load transients which can stall the motor, thus enabling a controlled restart with full torque).
Start-up	Difficult or impossible to start from zero speed.	InstaSPIN-FOC includes:
	Observer feedback at zero speed is not stable,	 Zero Speed start with forced-angle
	resulting in poor rotor angle accuracy and speed feedback.	 100% torque at start-up
		• FAST rotor flux angle tracking converges within one electrical cycle.
		FAST is completely stable through zero speed, providing accurate speed and angle estimation.
Current Loop	Tuning FOC current control is challenging – especially for novices.	Automatically sets the initial tuning of current controllers based on the parameters identified. User may update gains or use own controllers, if desired.
		The algorithm to fully tune the observer and torque controller takes less than 2 minutes.
Feedback Signals	System offsets and drifts are not managed.	FAST includes automatic hardware/software calibration and offset compensation.
		FAST requires 2-phase currents (3 for 100% and over-modulation), 3-phase voltages to support full dynamic performance, DCbus voltage for ripple compensation in current controllers.
		FAST includes an on-line stator resistance tracking algorithm.
Motor Types	Multiple techniques for multiple motors: standard back-EMF, Sliding Mode, Saliency tracking, induction flux estimators, or "mixed mode" observers.	FAST works with all 3-phase motor types, synchronous and asynchronous, regardless of load dynamics. Supports salient IPM motors with different Ls-d and Ls-q.
		Includes PowerWarp [™] for induction motors = energy savings.
Field-Weakening	Field-weakening region challenging for observers - as the Back-EMF signals grow too large, tracking and stability effected.	FAST estimator allows easy field weakening or field boosting applications due to the stability of the flux estimation in a wide range, including field weakening region.
Motor Temperature	Angle tracking degrades with stator temperature changes.	Angle estimation accuracy is improved from online stator resistance recalibration.
Speed Estimation	Poor speed estimation causes efficiency losses in the FOC system and less stable dynamic operation.	High quality low noise Speed estimator, includes slip calculation for induction motors.
Torque Estimation	Torque and vibration sensors typically required.	High bandwidth motor Torque estimator.

Table 1. FAST Estimator Compared to Typical Solutions



6 FAST Provides Sensorless FOC Performance

6.1 FAST Estimator Replaces Mechanical Sensor

Field-oriented control (FOC) of an electric motor results in superior torque control, lower torque ripple, and in many cases, improved efficiency compared to traditional AC control techniques. For best dynamic response, rotor flux referenced control algorithms are preferred to stator flux referenced techniques. To function correctly, these systems need to know the spacial angle of the rotor flux with respect to a fixed point on the stator frame (typically the magnetic axis of the phase A stator coil). This has traditionally been accomplished by a mechanical sensor (for example, encoder or resolver) mounted to the shaft of the motor. These sensors provide excellent angle feedback, but inflict a heavy toll on the system design. There are six major system impacts resulting from sensored angle feedback, as discussed below and illustrated in Figure 4:

- 1. The sensor itself is very expensive (often over \$2500 for a good resolver and several dollars for high volume integrated encoders).
- 2. The installation of the sensor requires skilled assembly, which increases labor costs.
- 3. The sensor often requires separate power supplies, which increases system costs and reduces reliability.
- 4. The sensor is the most delicate component of the system, which impacts system reliability, especially in harsh real-world applications.
- 5. The sensor feedback signals are brought back to the controller board via connectors, which also increases system costs and can significantly reduce reliability, depending on the type of connector.
- 6. The cabling required to bring the sensor signals back to the controller creates multiple challenges for the system designer:
 - Additional costs for the cable, especially if there is a substantial distance between the motor and controller.
 - Susceptibility to sources of noise, which requires adding expense to the cable with special shielding or twisted pairs.
 - The sensor and associated cabling must be earth grounded for safety reasons. This often adds additional cost to isolate these signals, especially if the processor which processes the sensor signals is not earth grounded.

In some applications where the motor is enclosed (for example, compressors), a sensored solution is impractical due to the cost of getting the feedback wires through the casing. For these reasons, designers of FOC systems are highly motivated to eliminate the sensor altogether, and obtain the rotor flux angle information by processing signals which are already available on the controller circuit board. For synchronous machines, most techniques involve executing software models of the motor being controlled to estimate the back-EMF waveforms (rotor flux), and then processing these sensed waveforms to extract an estimation of the rotor shaft angle, and a derivation of its speed. For asynchronous machines the process is a bit more complicated, as this software model (observer) must also account for the slip which exists between the rotor and rotor flux.

However, in both cases, performance suffers at lower speeds due to the amplitude of the back-EMF waveforms being directly proportional to the speed of the motor (assuming no flux weakening). As the back-EMF amplitude sinks into the noise floor, or if the ADC resolution cannot faithfully reproduce the small back-EMF signal, the angle estimation falls apart, and the motor drive performance suffers.

To solve the low-speed challenge, techniques have been created that rely on high frequency injection to measure the magnetic irregularities as a function of angle (that is, magnetic saliency) to allow accurate angle reconstruction down to zero speed. However, this introduces another set of control problems. First, the saliency signal is non-existent for asynchronous motors and very small for most synchronous machines (especially those with surface mount rotor magnets). For the motors that do exhibit a strong saliency signal (for example, IPM motors), the signal often shifts with respect to the rotor angle as a function of loading, which must be compensated. Finally, this angle measurement technique only works at lower speeds where the fundamental motor frequency does not interfere with the interrogation frequency. The control system has to create a mixed-control strategy, using high-frequency injection tracking at low speed, then move into Back-EMF based observers at nominal and high speeds.

9



FAST Provides Sensorless FOC Performance

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With any technique, the process of producing a stable software sensor is also extremely challenging, as this motor model (observer) is essentially its own control system that needs to be tuned per motor across the range of use. This tuning must be done with a stable forward control loop. Needed is a stable torque (and usually speed) loop to tune the observer, but how do you pre-tune your forward control without a functioning observer? One option is to use a mechanical sensor for feedback to create stable current and speed loops, and then tune your software sensor in parallel to the mechanical sensor. However, the use of a mechanical sensor is often not practical. This problem has delayed market use of software sensors for sensorless FOC control.

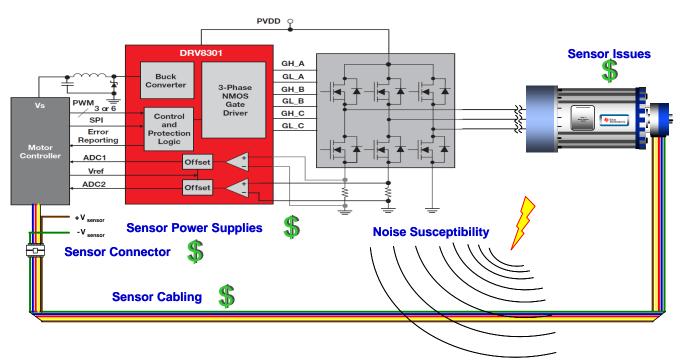


Figure 4. Sensored FOC System

In summary, these existing solutions all suffer from various maladies including:

- Poor low-speed performance (back-EMF and SMO)
- Poor high-speed performance (saliency observers)
- Poor dynamic response
- Calculation intensive (multi-modal observers)
- Parameter sensitivity
- Requirement for observer tuning.

The most recent innovation in the evolution of sensorless control is InstaSPIN-FOC. Available as a Ccallable library embedded in on-chip ROM on several TI processors, InstaSPIN-FOC was created to solve all of these challenges, and more. It reduces system cost and development time, while improving performance of three-phase variable speed motor systems. This is achieved primarily through the replacement of mechanical sensors with the proprietary FAST estimator. FAST is an estimator that:

- Works efficiently with all three phase motors, taking into account the differences between synchronous/asynchronous, salient/non-salient, and permanent/non-permanent/induced magnets.
- Dramatically improves performance and stability across the entire operating frequency and load range for a variety of applications.
- Removes the manual tuning challenge of traditional FOC systems:
 - Qbservers and estimators, completely removes required tuning.
 - Current loop regulators, dramatically reduces required tuning.



- Eliminates or reduces motor parameter variation effects.
- Automatically designs a stable and functional control system for most motors in under two minutes.

6.2 Rotor Angle Accuracy Critical for Performance

Why has the need for a precise estimation of the rotor flux angle driven many to use mechanical sensors?

For efficient control of three-phase motors, the objective is to create a rotating flux vector on the stator aligned to an ideal orientation with respect to the rotor in such a way that the rotor field follows the stator field while creating necessary torque and using the minimum amount of current.

- Stator: stationary portion of the motor connected to the microprocessor-controlled inverter.
- Ideal Orientation: 90 degrees for non-salient synchronous; slightly more for salient machines, and slightly less in asynchronous machines since part of the current vector is also used to produce rotor flux.
- **Rotor**: rotating portion of the motor, produces torque on the shaft to do work.

To achieve this, you need to extract the following information from the motor:

- Current being consumed by each phase.
- Precise relative angle of the rotor flux magnetic field (usually within ± 3 electrical degrees), so you can orient your stator field correctly.
- For speed loops, you also need to know rotor speed.

6.3 Phase Currents Key to Estimator Accuracy

Resistor shunt current measurement is a very reasonable technique for measuring phase current in a motor control inverter. There are three widely used examples, the 1-, 2-, and 3-shunt resistor measurements. While at first the 1- and 2-shunt techniques seem to reduce cost, they require much faster and more expensive amplifier circuits. These 1- and 2-shunt current measurements also limit the capability of the current feedback which will limit the ability of the drive to use the full voltage that is provided to the inverter. The 3-shunt technique is superior and not much different in cost due to the advantage of using cheap slow current amplifier circuits. For best performance and cost with the FAST and InstaSPIN-FOC, the 3-shunt technique is recommended.

For more details, see the *TMS320F2806xF* InstaSPIN-FOC/TMS320F2806xM InstaSPIN-MOTION User's *Guide* (literature number <u>SPRUHJ1</u>).



Evaluating FAST and InstaSPIN-FOC Performance

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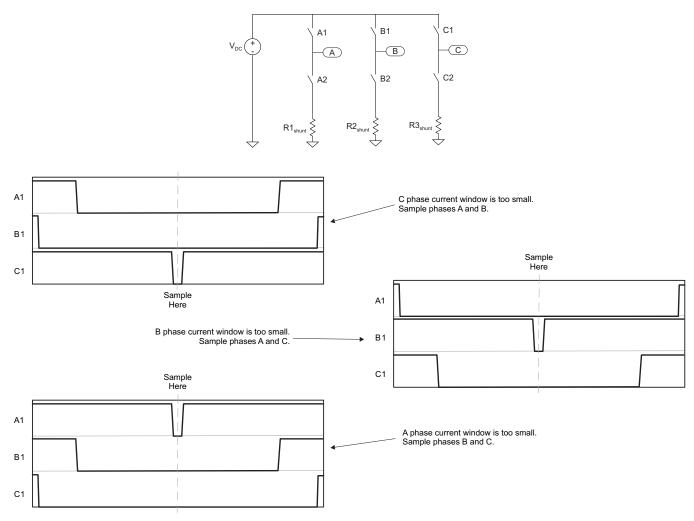


Figure 5. Inverter Using the 3-Shunt Current Sampling Technique

7 Evaluating FAST and InstaSPIN-FOC Performance

FAST and InstaSPIN-FOC performance data is being collected and will be provided in a future revision of this document.

8 Microcontroller Resources

The F2806xF microcontroller resources required by the InstaSPIN libraries are discussed in detail in the TMS320F2806xF InstaSPIN-FOC/TMS320F2806xM InstaSPIN-MOTION User's Guide (literature number SPRUHJ1).

Specifically for the library implementation and where the code is loaded and executed from, the following resources categories are discussed in this document:

- CPU Utilization
- Memory Allocation
- Stack Utilization
- Digital and Analog Pins Utilization

TEXAS INSTRUMENTS

Microcontroller Resources

Table 2. Hardware Features

			Table 2.	Hardware F	ealures			
FEATURE		TYPE		069F MHz)		068F MHz)	F280 (90 M	
Package Typ (PFP and PZ PN and PZ a	P are HTQFPs.		100-Pin PZ PZP	80-Pin PN PFP	100-Pin PZ PZP	80-Pin PN PFP	100-Pin PZ PZP	80-Pin PN PFP
Instruction cy	cle	-	11.1	1 ns	11.1	1 ns	11.1	1 ns
Floating-Poin	t Unit (FPU)		Y	es	Y	es	Ye	es
Viterbi, Comp (VCU)	blex Math, CRC Unit		Y	es	Y	es	No	
Control Law	Accelerator (CLA)	0	Y	es	N	lo	N	0
6-Channel DI	MA	0	Y	es	Y	es	Ye	es
On-chip Flasl	h (16-bit word)	-	12	8K	12	8K	64	ΙK
On-chip SAR	AM (16-bit word)	-	50	ЭК	50	ЭК	26	БК
Code security SARAM, and	/ for on-chip Flash, OTP blocks	_	Y	es	Y	es	Ye	es
Boot ROM (3	2K x 16)	-	Y	es	Y	es	Ye	es
One-time pro ROM (16-bit word)	grammable (OTP)	_	1	К	1	К	1	ĸ
ePWM output	ts	1	16	14	16	14	16	14
	on ePWM Channels	1	8	6	8	6	8	6
eCAP inputs		0		3		3	3	
High-resolutio	on capture modules	0	4	1	4	1	4	1
eQEP module	es	0	2	1	2	1	2	1
Watchdog timer		-	Y	es	Y	es	Ye	es
	MSPS		3.46		3.46		3.46	
	Conversion Time			289 ns	289 ns	289 ns	289 ns	289 ns
12-Bit ADC	Channels	3	16	12	16	12	16	12
	Temperature Sensor		Y	es	Y	es	Ye	es
	Dual Sample-and- Hold		Yes		Yes		Yes	
32-Bit CPU ti	mers	-	3		;	3	3	3
Comparators	with Integrated DACs	0	3		;	3	3	
Inter-integrate	ed circuit (I2C)	0		1	1		1	
Multichannel (McBSP)	Buffered Serial Port	1	1		1		1	
Enhanced Co (eCAN)	ontroller Area Network	0	1		1		1	
Serial Periph	eral Interface (SPI)	1	2		2		2	
(SCI)	unications Interface	0	:	2	2		2	2
	rial Bus (USB)	0		1	1		1	
2-pin Oscillator				1	1		1	
0-pin Oscillator			:	2	:	2	2	2
I/O pins (shared)	GPIO AIO	-	54	40 6	54	40	54	40
External interrupts		_		3	6 3		6	
Supply voltage		_		3 3 V		3 3 V	3.3	
Cabbiy volia	T: -40°C to 105°C	_	PZ	PN	PZ	PN	PZ 3.3	PN
Temperature		_	PZP	PFP	PZP	PFP	PZP	PFP
options	Q: -40°C to 125°C	_	PZP	PFP	PZP	PFP	PZP	PFP

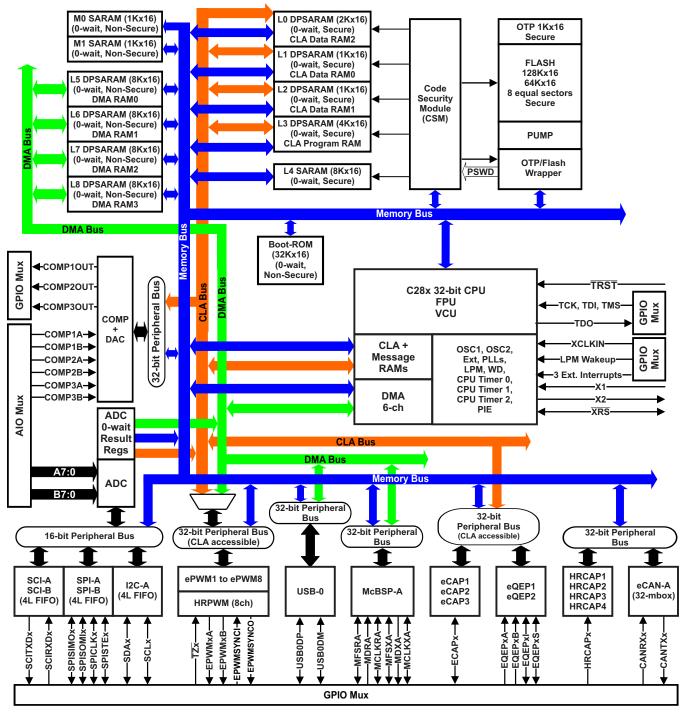


Microcontroller Resources

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FEATURE	TYPE	F28069F PE (90 MHz)		F28068F (90 MHz)		F28062F (90 MHz)	
Package Type (PFP and PZP are HTQFPs. PN and PZ are LQFPs.)		100-Pin PZ PZP	80-Pin PN PFP	100-Pin PZ PZP	80-Pin PN PFP	100-Pin PZ PZP	80-Pin PN PFP
Product status for T-temperature devices and S-temperature devices	-	TMS		TMS		TMS	
Product status for Q-temperature devices	atus for Q-temperature – TMX TMX		ТМХ				

Table 2. Hardware Features (continued)



A Not all peripheral pins are available at the same time due to multiplexing.

Figure 6. Functional Block Diagram





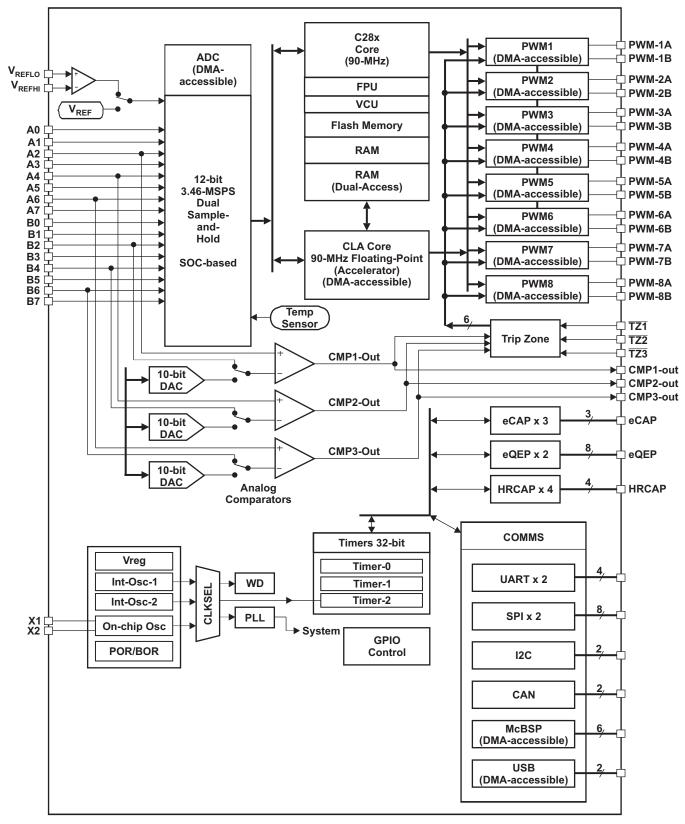


Figure 7. Peripheral Blocks

InstaSPIN-FOC provides flexibility throughout its design, including its software execution clock tree. Figure 8 illustrates the options available to the designer to manage the real-time scheduling of each of the major software functions. Balancing motor performance with CPU loading is not difficult, shortening system integration time.

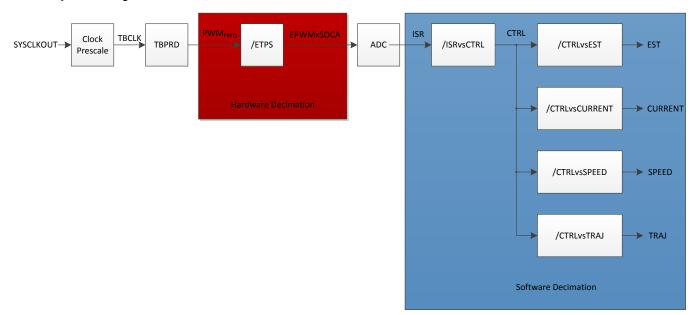


Figure 8. Software Execution Clock Tree Provides Flexibility with Real-Time Scheduling

Executing from single-cycle memory, total execution time for the full implementation of InstaSPIN-FOC will depend on the software execution clock tree. Table 3 shows the CPU cycles used when a full implementation of InstaSPIN is done, as well as users' code is loaded to FLASH. Note the impact of the software execution tree to total execution time. Table 4 shows the CPU loading and available MIPs for other system functions. The execution time does not change significantly from FULL to MIN implementations since the FAST block requires the largest number of CPU cycles and is in ROM for all implementations.

	CPU Cycles			Executed From		
Function Name	Min	Average	Max	ROM	RAM	FLASH
DRV_acqAdcInt	25	25	25	х	х	
DRV_readAdcData	108	108	108	х	х	\checkmark

		CPU Cycles		Ex	ecuted Fr	om
Function Name	Min	Average	Мах	ROM	RAM	FLASH
Ctrl_run				\checkmark	х	х
$\rm R_{s}$ Online Disabled, ISR vs CTRL = 1, CTRL vs EST = 1	2345	2355	2425			
CTRL vs EST = 2	1154	1760	2425			
CTRL vs EST = 3	1154	1562	2425			
ISR vs CTRL = 2, CTRL vs EST = 1	58	1207	2425			
CTRL vs EST = 2	58	909	2425			
CTRL vs EST = 3	58	810	2425			
ISR vs CTRL = 3, CTRL vs EST = 1	58	824	2425			
CTRL vs EST = 2	58	626	2425			
CTRL vs EST = 3	58	560	2425			
$\rm R_{s}$ Online Enabled, ISR vs CTRL = 1, CTRL vs EST = $\rm 1$	2807	2821	2894			
CTRL vs EST = 2	1154	1993	2894			
CTRL vs EST = 3	1154	1717	2894			
ISR vs CTRL = 2, CTRL vs EST = 1	58	1439	2894			
CTRL vs EST = 2	58	1025	2894			
CTRL vs EST = 3	58	887	2894			
ISR vs CTRL = 3, CTRL vs EST = 1	58	979	2894			
CTRL vs EST = 2	58	702	2894			
CTRL vs EST = 3	58	610	2894			
DRV_writePwmData	64	64	64	х	х	\checkmark
CTRL_setup	37	51	178	\checkmark	х	х

Table 4. CPU loading for FULL Implementation Executing from ROM and FLASH

2806xF CPU = 90 MHz			
Available MIPs = 90 MIPs PWM = 20 kHz	CPU Utilization [%]	MIPs Used [MIPS]	MIPS Available [MIPS]
R _s Online Disabled, ISR vs CTRL = 1, CTRL vs EST = 1	57.71	51.94	38.06
CTRL vs EST = 2	44.49	40.04	49.96
CTRL vs EST = 3	40.09	36.08	53.92
ISR vs CTRL = 2, CTRL vs EST = 1	32.2	28.98	61.02
CTRL vs EST = 2	25.58	23.02	66.98
CTRL vs EST = 3	23.38	21.04	68.96
ISR vs CTRL = 3, CTRL vs EST = 1	23.69	21.32	68.68
CTRL vs EST = 2	19.29	17.36	72.64
CTRL vs EST = 3	17.82	16.04	73.96
R_s Online Enabled, ISR vs CTRL = 1, CTRL vs EST = 1	68.07	61.26	28.74
CTRL vs EST = 2	49.67	44.7	45.3
CTRL vs EST = 3	43.53	39.18	50.82
ISR vs CTRL = 2, CTRL vs EST = 1	37.36	33.62	56.38
CTRL vs EST = 2	28.16	25.34	64.66
CTRL vs EST = 3	25.09	22.58	67.42
ISR vs CTRL = 3, CTRL vs EST = 1	27.13	24.42	65.58
CTRL vs EST = 2	20.98	18.88	71.12
CTRL vs EST = 3	18.93	17.04	72.96



Microcontroller Resources

2806xF CPU = 90 MHz Available MIPs = 90 MIPs PWM = 20 kHz	CPU Utilization [%]	MIPs Used [MIPS]	MIPS Available [MIPS]
R_s Online Disabled, ISR vs CTRL = 1, CTRL vs EST = 1	60.02	54.02	35.98
CTRL vs EST = 2	46.8	42.12	47.88
CTRL vs EST = 3	42.38	38.14	51.86
ISR vs CTRL = 2, CTRL vs EST = 1	33.49	30.14	59.86
CTRL vs EST = 2	26.87	24.18	65.82
CTRL vs EST = 3	24.67	22.2	67.8
ISR vs CTRL = 3, CTRL vs EST = 1	24.64	22.18	67.82
CTRL vs EST = 2	20.22	18.2	71.8
CTRL vs EST = 3	18.76	16.88	73.12
R_s Online Enabled, ISR vs CTRL = 1, CTRL vs EST = 1	70.42	63.38	26.62
CTRL vs EST = 2	52	46.8	43.2
CTRL vs EST = 3	45.87	41.28	48.72
ISR vs CTRL = 2, CTRL vs EST = 1	38.69	34.82	55.18
CTRL vs EST = 2	29.47	26.52	63.48
CTRL vs EST = 3	26.4	23.76	66.24
ISR vs CTRL = 3, CTRL vs EST = 1	28.09	25.28	64.72
CTRL vs EST = 2	21.96	19.76	70.24
CTRL vs EST = 3	19.91	17.92	72.08

Table 5. CPU loading for FULL Implementation Executing from ROM and FLASH

8.1 Memory Allocation and Utilization

Figure 9, Figure 10, and Table 6 show the memory map of the 28069, the location in ROM where the InstaSPIN-FOC library is located, and the required allocation of L8 RAM for the library to use. For a general memory map of these devices refer to the specific datasheet.



1			
0x00 0000	Data Space Prog Space		
0x00 0000	M0 Vector RAM (Enabled if VMAP = 0)		
0x00 0040	M0 SARAM (1K x 16, 0-Wait)		
0x00 0800	M1 SARAM (1K x 16, 0-Wait)		
0x00 0D00	Peripheral Frame 0 PIE Vector - RAM (256 x 16) (Enabled if VMAP = 1, ENPIE = 1)		
0x00 0E00	Peripheral Frame 0	Reserved	
0x00 1400	CLA Registers		
0x00 1480	CLA-to-CPU Message RAM		
0x00 1500	CPU-to-CLA Message RAM		
0x00 1580	Reserved		
0x00 2000		erved	
0x00 4000		, veu	
0x00 5000	USB Control Registers ^(A) Peripheral Frame 3		
0 00 0000	(4K x 16, Protected) DMA-Accessible	Reserved	
0x00 6000	Peripheral Frame 1 (4K x 16, Protected)		
0x00 7000	Peripheral Frame 2 (4K x 16, Protected)		
0x00 8000	L0 DPSARAM (2K x 16) (0-Wait, Secure Zone + ECSL, CLA Data RAM2)		
0x00 8800	L1 DPSARAM (1K x 16) (0-Wait, Secure Zone + ECSL, CLA Data RAM 0)		
0x00 8C00	L2 DPSARAM (1K x 16) (0-Wait, Secure Zone + ECSL, CLA Data RAM 1)		
0x00 9000	L3 DPSARAM (4K x 16) (0-Wait, Secure Zone + ECSL, CLA Program RAM)		
0x00 A000	L4 SARAM (8K x 16) (0-Wait, Secure Zone + ECSL)		
0x00 C000	L5 DPSARAM (8K x 16) (0-Wait, DMA RAM 0)		
0x00 E000	L6 DPSARAM (8K x 16) (0-Wait, DMA RAM 1)		
0x01 0000	L7 DPSARAM (8K x 16) (0-Wait, DMA RAM 2)		
0x01 2000	L8 DPSARAM (8K x 16) (0-Wait, DMA RAM 3)		
0x01 4000	Reserved		
0x3D 7800	User OTP (1K x 16, Secure Zone + ECSL)		
0x3D 7BFA	Reserved		
0x3D 7C80	Calibration Data		
0x3D 7CC0	Get_mode function		
0x3D 7CD0	Reserved		
0x3D 7E80	PARTID		
	Calibrat	ion Data	
0x3D 7EB0	Calibration Data		
0x3D 8000	Reserved		
0x3F 7FF8	FLASH (128K x 16, 8 Sectors, Secure Zone + ECSL)		
0x3F 7FF8 0x3F 8000	128-Bit Password		
0x3F 8000 0x3F FFC0	Boot ROM (32K x 16, 0-Wait)		
	Vector (32 Vectors, I	Enabled if VMAP = 1)	

Figure 9. 28069 Memory Map

Features	2806xF
Maximum Number of Motors that can be controlled	2
FAST Version	1.6
ROM Library [size, hex, words]	4000
ROM Library Start [address, hex]	3F 8000
Library Required RAM [size, hex, words]	800
Library Start RAM [address, hex]	01 3800

Figure 10 highlights the pieces of ROM EXE-only memory used by the libraries. EXE-only is execute only memory where read access is not possible.

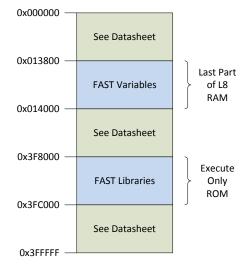


Figure 10. 2806xF Allocated Memory for InstaSPIN-FOC Library

Table 7 summarizes the memory used for the (4) most common configurations as shown in Figure 2 and Figure 3 (Full and Min implementations), with user memory optionally in FLASH or RAM. Note the code size increase as fewer functions in ROM are used.

Code Configurations		Memory Sizes (16bit Words)			Maximum Stack
ROM Code	User Code	RAM	Flash	Total	Used (16bit Words)
Full Implementation	RAM	0x1870	0x0000	0x1870	0x0120
Full Implementation	FLASH	0x001E	0x186C	0x188A	0x0120
Min Implementation	RAM	0x1F31	0x0000	0x1F31	0x0120
Min Implementation	FLASH	0x001E	0x1F2D	0x1F4B	0x0120

Table 7. User Memory and Stack Sizes

8.2 Pin Utilization

Flexibility in the design of InstaSPIN-FOC allows for multiple motors to be supported. Table 8 lists the minimum and maximum pins used per motor. Note that a F2806xF provides (14) ePWM outputs with the 100-pin package, and (12) with the 80-pin.



Microcontroller Resources

Table 8. Pin Utilization Per Motor

	Pin Name	Pins Usage Per Motor	
Pin Type		Min	Max
Digital	PWM1A	3	7
	PWM1B (Optional)	(Requires External Fault and External Complementary Mode	
	PWM2A	with Dead Time)	
	PWM2B (Optional)		
	PWM3A		
	PWM3B (Optional)		
	Trip Zone (Optional)		
Analog	IA	5	7
	IB	(Only two currents and no VBUS ripple compensation)	
	IC (Optional)		
	VA		
	VB		
	VC		
	VBUS (Optional)		



Appendix A Definition of Terms and Acronyms

- ACIM Alternating current induction motor.
- CCStudio Code Composer Studio.
- **FAST** Unified observer structure which exploits the similarities between all motors that use magnetic flux for energy transduction, automatically identifying required motor parameters and providing the following motor feedback signals:
 - High-quality Flux signal for stable flux monitoring and field weakening.
 - Superior rotor flux Angle estimation accuracy over wider speed range compared to traditional observer techniques independent of all rotor parameters for ACIM.
 - Real-time low-noise motor shaft Speed signal.
 - Accurate high bandwidth Torque signal for load monitoring and imbalance detection.
- FOC Field-oriented control.
- **Forced-Angle** Used for 100% torque at start-up until the FAST rotor flux angle tracker converges within first electrical cycle.
- **InstaSPIN-FOC** Complete sensorless FOC solution provided by TI on-chip in ROM on the 2806xF devices that can identify, tune and efficiently control your motor in minutes, without the use of any mechanical rotor sensors.
- **IPM** Interior permanent magnet motor.

Motor Parameters ID or Motor Commissioning -

- **PI** Proportional-integral regulator.
- **PMSM** Permanent magnet synchronous motor.
- **PowerWarp** Mode of operation used for AC induction motors (ACIM) that allows minimum current consumption.
- **Rs-Offline Recalibration** InstaSPIN-FOC feature that is used to recalibrate the stator resistance, Rs, when the motor is not running.
- **Rs-Online Recalibration** InstaSPIN-FOC feature that is used to recalibrate the stator resistance, Rs, while the motor is running in closed loop.
- **SVM** Space-vector modulation.

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