

*TMS570LS31x/21x Microcontroller*

**Silicon Revision B**

# Silicon Errata



Literature Number: SPNZ193D  
November 2012–Revised October 2013

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## ***TMS570LS31x/21x Microcontroller***

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This document describes the known exceptions to the functional specifications for the device.

### **1 Device Nomenclature**

To designate the stages in the product development cycle, TI assigns prefixes to the part numbers of all devices. Each commercial family member has one of three prefixes: TMX, TMP, or TMS (for example, **TMS570LS3137**). These prefixes represent evolutionary stages of product development from engineering prototypes (TMX) through fully qualified production devices/tools (TMS).

Device development evolutionary flow:

**TMX** — Experimental device that is not necessarily representative of the final device's electrical specifications.

**TMP** — Final silicon die that conforms to the device's electrical specifications but has not completed quality and reliability verification.

**TMS** — Fully-qualified production device.

TMX and TMP devices are shipped against the following disclaimer:

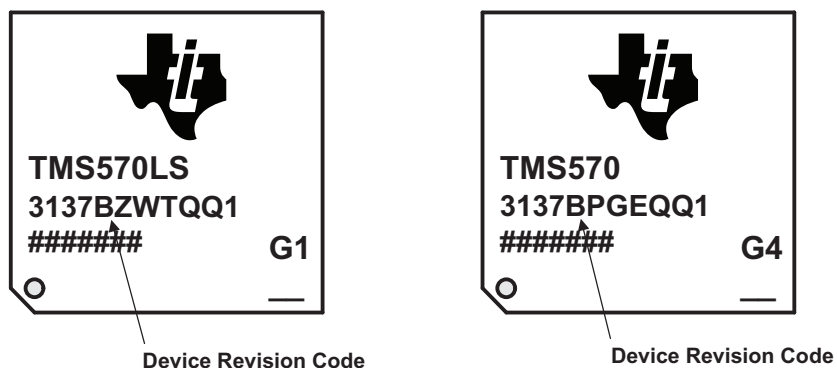
"Developmental product is intended for internal evaluation purposes."

TMS devices have been characterized fully, and the quality and reliability of the device have been demonstrated fully. TI's standard warranty applies.

Predictions show that prototype devices (TMX or TMP) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

## 2 Revision Identification

**Figure 1** provides examples of the TMS570LSx device markings. The device revision can be determined by the symbols marked on the top of the device.



**Figure 1. Device Revision Code Identification**

### 3 Revision History

This silicon errata revision history highlights the technical changes made from the previous revision of this document.

**Table 1. Revision History from Errata Document Revision C to Revision D**

Advisory Changes in Advisory List	Advisory ID
Added advisory(s)	DEVICE#B064, DEVICE#B065, DEVICE#B066, DEVICE#B069, PBIST#4, MIBSPI#139
Removed advisory(s)	None
Modified advisory(s)	None
Other	None

## 4 Known Design Exceptions to Functional Specifications

The following table lists the known exceptions to the functional specifications for the device.

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<b>DEVICE#142</b>	<b><i>CPU Abort Not Generated on Write to Unimplemented MCRC Space</i></b>
<b>Severity</b>	Low
<b>Expected Behavior</b>	A write to the illegal address region of the MCRC module will generate an abort
<b>Issue</b>	A CPU Abort does not get generated per the following condition:
<b>Conditions</b>	When a normal mode write to an illegal address region of MCRC register space is followed by a debug mode access.
<b>Implications</b>	When debugging, either a breakpoint on the instruction after the illegal write, or single stepping through the illegal write will not generate an abort.
<b>Workaround(s)</b>	None

<b>DEVICE#179</b>	<b><i>CPU may Hang on STM Write to Memory on EMIF</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	CPU completes STM (store multiple) instruction
<b>Issue</b>	The CPU may hang on a STM instruction
<b>Condition</b>	<p>A STM instruction writes to an address in the EMIF followed by a write to a peripheral address mapped as strongly ordered</p> <p>OR</p> <p>A STM instruction writes to an address in the EMIF followed by another STM write to a peripheral which is mapped as "Strongly Ordered".</p>
<b>Implication(s)</b>	The CPU hangs waiting for the memory write to complete.
<b>Workaround(s)</b>	Do not map either the EMIF address space or any peripheral as "Strongly Ordered" when writing to the EMIF using the STM instruction. (burst mode)

<b>DEVICE#B053</b>	<b><i>CPU code execution could be halted on a device warm reset if the core power domain # 2 is disabled by software.</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The CPU code execution must start from the reset vector (address 0x00000000) upon a device warm reset and is not affected by the state of any switchable device power domain.
<b>Issue</b>	CPU code execution could be halted upon a warm reset if the core power domain # 2 has been disabled by software prior to the device warm reset.
<b>Condition</b>	The behavior is not dependent on any particular operating condition.
<b>Implication(s)</b>	CPU code execution is halted so that a system hang occurs. An external monitor must be present to prevent the system from entering an unsafe state when this happens.
<b>Workaround(s)</b>	The application must not disable the core power domain # 2 in software via the Power Management Module (PMM) registers, even if the modules inside this core power domain are not used in the application.

**DEVICE#B056** — *Error flags are incorrectly set in the Error Signaling Module upon a device warm reset for modules in core power domains that are disabled* www.ti.com

<b>DEVICE#B056</b>	<b><i>Error flags are incorrectly set in the Error Signaling Module upon a device warm reset for modules in core power domains that are disabled</i></b>
<b>Severity</b>	2-High
<b>Expected Behavior</b>	Once a core power domain is disabled (turned off) no error must be indicated to the ESM from a module in this power domain.
<b>Issue</b>	When a core power domain is disabled, a device warm reset causes error signals to be sent to the ESM that appear to be from the power domain that has been disabled. This is caused due to the incorrect power domain isolation implemented on these error signals. All such error signals are connected to ESM group1 channels.
<b>Condition</b>	The behavior is not dependent on any particular operating condition.
<b>Implication(s)</b>	If the application has enabled the generation of interrupts or the assertion of the nERROR output when any flag gets set in the ESM module's group1 status registers, then the CPU will receive an interrupt and/or the nERROR pin will be driven for an error condition ascribed to a module that is in a core power domain that has been disabled.
<b>Workaround(s)</b>	When a core power domain is disabled, the application must ensure that any error signal related to a module within this power domains is not configured to generate an interrupt to the CPU or assert the device nERROR output. Alternately, the application could clear the ESM group1 status flags upon each device warm reset condition before it enables interrupts and nERROR drive features for the ESM group1 channels.

<b>DEVICE#B058</b>	<b><i>Device I/Os Could Toggle After Warm Reset</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	All device I/Os are tri-stated (inputs) after a device warm reset. No toggle is expected on any of the device I/Os without software configuration.
<b>Issue</b>	All core power domains get enabled on a system reset. After the core domains are turned ON, the domain outputs are unstable for a few cycles after the system reset is released. This could cause the I/Os to toggle.
<b>Condition</b>	The behavior is not dependent on any particular operating condition.
<b>Implication(s)</b>	Any external circuit connected to the device I/Os could see activity on the connections to the device after a devcie warm reset.
<b>Workaround(s)</b>	There are no software or hardware workarounds for this issue.

<b>DEVICE#B059</b> <b><i>Pull Controls Not Correct for Modules in Power Domains That Are Disabled</i></b>	
<b>Severity</b>	4-Low
<b>Expected Behavior</b>	I/O signals used in communication protocols require pull-ups, as specified. These pull controls need to be consistent even if the core power domains containing these communication modules are disabled.
<b>Issue</b>	Some communication modules; SPI4, CAN3 and DMM, are part of core power domains that can be switched off. When a core power domain is switched off, all outputs from this domain are held at a defined "isolation value". This isolation value is incorrect for the following signals: SPI4nCS, SPI4nENA, DMMnENA, CAN3TX and CAN3RX. These signals are pulled down instead of being pulled up when the respective power domains containing the SPI4, DMM and CAN3 modules are disabled.
<b>Condition</b>	The behavior is not dependent on any particular operating condition.
<b>Implication(s)</b>	If the SPI4 or CAN3 modules are used to communicate with external circuitry, and if the core power domain containing these modules is disabled, then there is an issue in that the chip select output from the SPI4 and both the CAN3 connections to the transceiver are pulled down instead of being pulled up.
<b>Workaround(s)</b>	None

<b>DEVICE#B063</b>	<b><i>Unexpected PSCON Compare Error</i></b>
<b>Severity</b>	3 - Medium
<b>Expected Behavior</b>	No compare errors when disabling a logic power domain
<b>Issue</b>	The device may generate a PSCON compare error when disabling a logic power domain.
<b>Conditions</b>	<p>This problem might occur if:</p> <ol style="list-style-type: none"> <li>1. A logic power domain is disabled at reset by a factory OTP setting</li> <li>2. Software explicitly disables a power domain.</li> </ol>
<b>Implications</b>	ESM group 1 channel 38 and channel 39 errors may be generated.
<b>Workaround(s)</b>	<p>Before enabling the effect of ESM group 1 channels 38 or 39</p> <ol style="list-style-type: none"> <li>1. Disable power domains that are to be turned off (if no power domains are to be disabled by software, skip to step 2) <ul style="list-style-type: none"> <li>• Write to the PDCLK_DISx register to disable all clocks to the power domain</li> <li>• Write 0xA to the LOGICPDONx register to power down the domain</li> <li>• Poll for LOGICPDPWRSTATx to become 00. The power domain is now powered down (Allows for delay time of PowerGood signal)</li> </ul> </li> <li>2. Clear any PSCON compare error flags <ul style="list-style-type: none"> <li>• Write 0x000F0000 to LPDDCSTAT1 (0xFFFF00B0)</li> </ul> </li> <li>3. Clear ESM Group 1 flags 38 and 39 <ul style="list-style-type: none"> <li>• Write 0x000000C0 to ESMSR4 (0xFFFFF558)</li> </ul> </li> <li>4. Enable the effect of ESM group 1 channels 38 and 39</li> </ol> <p>The PSCON is now in lock step with its diagnostic partner, and any difference will result in an ESM error.</p>

**DEVICE#B064** — *Incorrect Write to External Memory using Store-Multiple (STMxx) CPU instruction when first target address is not aligned to a 64-bit boundary* www.ti.com

<b>DEVICE#B064</b>	<b><i>Incorrect Write to External Memory using Store-Multiple (STMxx) CPU instruction when first target address is not aligned to a 64-bit boundary</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	Writes to the external memory should happen correctly using any of the supported ARM/Thumb2 assembly instructions.
<b>Issue</b>	<p>Different issues are observed depending on the configuration of the external memory:</p> <ul style="list-style-type: none"> <li>• When external memory is configured as "normal" type <b>AND</b> first target address is not aligned to 64-bit boundary <ul style="list-style-type: none"> <li>– Data is not written at all</li> </ul> </li> <li>• When external memory is configured as "device" type <b>AND</b> first target address is not aligned to 64-bit boundary <ul style="list-style-type: none"> <li>– Data is written to the wrong address</li> </ul> </li> <li>• When external memory is configured as "strongly-ordered" type <b>AND</b> first target address is not aligned to 64-bit boundary <ul style="list-style-type: none"> <li>– Write access does not complete correctly causing the CPU to hang</li> </ul> </li> </ul>
<b>Condition</b>	CPU performs a store-multiple write operation to external memory AND the first target address in external memory is not aligned to a 64-bit boundary.
<b>Implication(s)</b>	Data is either not written, or written to the wrong address, or the CPU itself hangs depending on the configuration of the external memory.
<b>Workaround(s)</b>	Ensure that the first target address of a store-multiple instruction is aligned to a 64-bit boundary.



<b>DEVICE#B065</b>	<b><i>RTP stuck after FIFO overflow</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The RTP is expected to automatically re-start transmission after an overflowed FIFO is drained.
<b>Issue</b>	Under certain situations, the RTP cannot recover automatically after a FIFO overflow. The RTP operation can only be resumed by a software module reset.
<b>Condition</b>	The FIFO overflows when there are too much data to be traced.
<b>Implication(s)</b>	RTP tracing could stop during operation when the FIFO overflows caused by a high data rate.
<b>Workaround(s)</b>	Enable halt on overflow. It will limit the usage and affect system operation.

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**DEVICE#B066**      ***Flash shut down when HCLK disable is requested***


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<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	Flash should continue to execute while FALLBACK is active and grace counters have not expired. The flash should therefore prevent the HCLK from disabling until the code has reached a boundary at which execution and data fetches have ceased for the given (grace period) duration.
<b>Issue</b>	HCLK disable request unconditionally disables the flash while CPU execution is still occurring
<b>Condition</b>	--
<b>Implication(s)</b>	If HCLK is disabled, the code execution will stop. Therefore, software must not disable HCLK while executing from flash.
<b>Workaround(s)</b>	none

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**DEVICE#B069**      ***More than one PCNT instruction on the same pin results in measurement error***


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**Severity**                      3 - Medium

**Expected Behavior**        It should be possible to use more than one Period/Pulse Count (PCNT) instruction to measure a single pin, as long as only one of the PCNT is configured for high resolution (hr\_lr=HIGH). For example, consider the following four code fragments.

**Code Fragment 1 - Should Be OK, But Fails Due to This Issue**

```
PC1    PCNT { hr_lr=HIGH, type=RISE2FALL, pin=2};
PC2    PCNT { hr_lr=LOW,  type=FALL2FALL, pin=2};
```

**Code Fragment 2 - Should Be OK, But Fails Due to This Issue**

```
PC1    PCNT { hr_lr=LOW,  type=RISE2FALL, pin=2};
PC2    PCNT { hr_lr=HIGH, type=FALL2FALL, pin=2};
```

**Code Fragment 3 - OK**

```
PC1    PCNT { hr_lr=LOW, type=RISE2FALL, pin=2};
PC2    PCNT { hr_lr=LOW, type=FALL2FALL, pin=2};
```

**Code Fragment 4 - Not Allowed**

```
PC1    PCNT { hr_lr=HIGH, type=RISE2FALL, pin=2};
PC2    PCNT { hr_lr=HIGH, type=FALL2FALL, pin=2};
```

Code fragment 4 is disallowed by the N2HET specification, because there is only one hi-res structure for each pin. Code Fragment 3 should work properly because both instructions measure the same pin with hr\_lr=low, so neither pin uses the hi-res structure. Code fragments 1 and 2 should work properly because only one of the two PCNT instructions are configured for hr\_lr=HIGH, and there is one hi-res structure available.

**Issue**                      Code Fragments 1 and 2 do not actually work properly. The addition of the loop resolution PCNT causes two issues.

First, a measurement error is introduced into the result of the PCNT instruction with hr\_lr=HIGH. Normally this instruction would return a result to within  $\pm\frac{1}{2}$  high resolution clock periods of the actual result, due to quantization noise. However another PCNT instruction on the same pin causes an error of up to  $\pm 1$  loop resolution period. Note that this error is greater than the normal loop resolution period error of  $\pm\frac{1}{2}$  loop resolution period; because the high-resolution bits also contribute to the error in this case.

Second, the PCNT instruction with hr\_lr=LOW should return a value with 0's in bit positions 6:0 (the high-resolution portion of the measurement result). This is the case when both PCNT instructions are set for hr\_lr=LOW (Code Fragment 3) but for Code Fragments 1 and 2 the loop resolution PCNT also returns a non-zero in bit positions 6:0.

**Conditions**                This problem occurs when more than one PCNT instruction is set for hr\_lr=HIGH on any given pin.

**Implications**             The impact is greatest when workaround option 1 cannot be applied due to the number of timer pins required by the application. If Option 1 cannot be applied, then the PCNT measurements on this pin are reduced to  $\pm\frac{1}{2}$  loop resolution period.

**Workaround(s)**            Option 1 - Use the HRSHARE feature and make both measurements with hr\_lr=HIGH. First, set the appropriate HRSHARE bit in the HETHRSH register. In the following example this means setting HETHRSH bit 1 - "HRSHARE3/2". This bit causes the input of device pin 2 to drive the N2HET pin inputs 2 and 3. Then modify the N2HET code sequence to use pin 3 for one of the PCNT instructions:

**Code Fragment 1 Modified for HRSHARE**

```
PC1    PCNT { hr_lr=HIGH, type=RISE2FALL, pin=2};
PC2    PCNT { hr_lr=HIGH, type=FALL2FALL, pin=3};
```

This option exceeds the original objective of because both PCNT measurements can be

made with high-resolution. The disadvantage of this workaround is that it requires the high-resolution structure of pin 3, leaving pin 3 only useable as a GPIO pin rather than as a timer pin.

Option 2 - Use only loop resolution mode PCNT instructions (as in Code Fragment 3). This will work properly while leaving pin 3 available for timing functions, but the resolution on both the period and duty cycle measurements is reduced to loop resolution.

---

**CORTEX-R4#26 (ARM ID-577077) Thumb STREXD Treated As NOP If Same Register Used For Both Source Operands**


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<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The STREXD instruction should work in Thumb mode when Rt and Rt2 are the same register.
<b>Issue</b>	The ARM Architecture permits the Thumb STREXD instruction to be encoded with the same register used for both transfer registers (Rt and Rt2). Because of this erratum, the Cortex-R4 processor treats such an encoding as UNPREDICTABLE and executes it as a NOP.
<b>Condition</b>	This occurs when the processor is in Thumb state and a STREXD instruction is executed which has Rt = Rt2. This instruction was new in ARM Architecture version 7 (ARMv7). It is not present in ARMv6T2 or other earlier architecture versions.
<b>Implication(s)</b>	If this occurs the destination register, Rd, which indicates the status of the instruction, is not updated and no memory transaction takes place. If the software is attempting to perform an exclusive read-modify-write sequence, then it might either incorrectly complete without memory being written, or loop forever attempting to complete the sequence.
<b>Workaround(s)</b>	This can be avoided by using two different registers for the data to be transferred by a STREXD instruction. This may involve copying the data in the transfer register to a second, different register for use by the STREXD.

---

**CORTEX-R4#27 (ARM ID-412027) Debug Reset Does Not Reset DBGDSCR When In Standby Mode**


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<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The debug reset input, PRESETDBGn, resets the processor's debug registers as specified in the ARMv7R Architecture. The debug reset is commonly used to set the debug registers to a known state when a debugger is attached to the target processor.
<b>Issue</b>	When the processor is in Standby Mode and the clock has been gated off, PRESETDBGn fails to reset the Debug Status and Control Register (DBGDSCR).
<b>Condition</b>	<ul style="list-style-type: none"> <li>• The DBGDSCR register has been written so that its contents differ from the reset values (most fields in this register reset to zero, though a few are UNKNOWN at reset)</li> <li>• The processor is in Standby Mode, and the clocks have been gated off, that is STANDBYWFI is asserted</li> <li>• The debug reset, PRESETDBGn, is asserted and deasserted while the processor clocks remain gated off</li> </ul>
<b>Implication(s)</b>	If this occurs, then after the reset the DBGDSCR register contains the values that it had before reset rather than the reset values. If the debugger relies on the reset values, then it may cause erroneous debug of the processor. For example, the DBGDSCR contains the ExtDCCmode field which controls the Data Communications Channel (DCC) access mode. If this field was previously set to Fast mode but the debugger assumes that it is in Non-blocking mode (the reset value) then debugger accesses to the DCC will cause the processor to execute instructions which were not expected.
<b>Workaround(s)</b>	This can be avoided by a workaround in the debug control software. Whenever the debugger (or other software) generates a debug reset, follow this with a write of zero to the DBGDSCR which forces all the fields to their reset values.

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**CORTEX-R4#33 (ARM ID-452032) Processor Can Deadlock When Debug Mode Enables Cleared**


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<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The Cortex-R4 processor supports two different debugging modes: Halt-mode and Monitor-mode, or both modes can be disabled. Bits [15:14] in the Debug Status and Control Register (DBGDSCR) control which, if any, mode is enabled. Debug events can be generated by the breakpoint unit, matching instructions executed. Deadlocks should not occur when the debug mode is changed.
<b>Issue</b>	If there are active breakpoints at the time when the debugging mode is changed, this can cause the processor to deadlock.
<b>Condition</b>	<ul style="list-style-type: none"> <li>• At least one breakpoint is programmed and active</li> <li>• Either halt-mode debugging or monitor mode debugging is enabled and the debug enable input, DBGEN is HIGH</li> <li>• An instruction which matches a breakpoint is fetched and executed</li> <li>• After the instruction is fetched, but before it has completed execution, both halt-mode and monitor-mode debugging are disabled by means of a write to DBGDSCR</li> </ul>
<b>Implication(s)</b>	Changing the debugging mode while breakpoints or watchpoints are active is UNPREDICTABLE. Therefore the above conditions cannot be met in normal (recommended) operation of the processor. However, if the conditions do occur, then the processor will deadlock, which is outside the bounds of permitted UNPREDICTABLE behavior.
<b>Workaround(s)</b>	None.

**CORTEX-R4#46 (ARM ID-599517) CP15 Auxiliary ID And Prefetch Instruction Accesses Are UNDEFINED**

<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	<p>The ARMv7-R architecture requires implementation of the following two features in CP15:</p> <ol style="list-style-type: none"> <li>1. An Auxiliary ID Register (AIDR), which can be read in privileged modes, and the contents and format of which are IMPLEMENTATION DEFINED.</li> <li>2. The operation to prefetch an instruction by MVA, as defined in the ARMv6 architecture, to be executed as a NOP.</li> </ol>
<b>Issue</b>	CP15 accesses to Auxiliary ID Register (AIDR) or an operation to prefetch an instruction by MVA will generate an UNDEFINED exception on Cortex-R4.
<b>Condition</b>	<p>Either of the following instructions is executed in a privileged mode:</p> <ul style="list-style-type: none"> <li>• MRC p15,1,&lt;Rt&gt;,c0,c0,7 ; Read IMPLEMENTATION DEFINED Auxiliary ID Register</li> <li>• MCR p15,0,&lt;Rt&gt;,c7,c13,1 ; NOP, was Prefetch instruction by MVA in ARMv6</li> </ul>
<b>Implication(s)</b>	If software attempts to read the AIDR as part of its process of identifying the processor on which it is running, it will encounter an unexpected UNDEFINED exception. If software attempts to execute an instruction prefetch using the ARMv6 CP15 prefetch operation, it will encounter an unexpected UNDEFINED exception.
<b>Workaround(s)</b>	In the first case, because the contents of the AIDR are IMPLEMENTATION DEFINED, it must always be read in conjunction with the Main ID Register (MIDR). A simple way of avoiding this would be to read and analyze the MIDR first and, if this indicates that the processor is Cortex-R4 skip the read of the AIDR. In the second case, any instruction to prefetch an instruction by MVA should be removed or replaced by a true NOP instruction.



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**CORTEX-R4#54 (ARM ID-639819) *Instruction Causing A Data Watchpoint Match Incorrectly Traced***


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<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	When tracing a program execution using the ETM, an extra instruction should not be traced.
<b>Issue</b>	When tracing a program execution using the ETM, an extra instruction is traced if a data watchpoint matches and causes a debug exception. The extra instruction that is traced is the instruction which caused the data watchpoint to match.
<b>Condition</b>	<p>The extra instruction is traced if the following occurs:</p> <ul style="list-style-type: none"> <li>• A hardware watchpoint matches</li> <li>• DBGEN is asserted</li> <li>• Monitor debug-mode is enabled</li> </ul>
<b>Implication(s)</b>	Trace analysis tools will incorrectly consider the instruction causing a data watchpoint match to have executed. If any of the ETM address comparators are configured to match on address of the instruction and the exact match bit is set, the comparator will incorrectly fire. This might cause an unexpected trigger or change in any ETM resources which are configured to be sensitive to the address comparator.
<b>Workaround(s)</b>	If a data abort exception is taken and the cause was a data watchpoint, the instruction traced immediately before the entry to the exception handler was not executed and must be discarded.

**CORTEX-R4#55 (ARM ID-722412) — CPACR.ASEDIS And CPACR.D32DIS Incorrect When Configured With Floating Point**  
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**CORTEX-R4#55 (ARM ID-722412) CPACR.ASEDIS And CPACR.D32DIS Incorrect When Configured With Floating Point**

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<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	In implementations of the VFPv3-D16 architecture that do not also implement the Advanced SIMD functionality, the ASEDIS and D32DIS bits, (bits [31] and [30] respectively of the Coprocessor Access Control Register (CPACR)) are read-as-one/writes ignored (RAO/WI). This indicates that Advanced SIMD functionality and registers D16-D31 are disabled.
<b>Issue</b>	Because of this erratum, these bits read zero in implementations of Cortex- R4F which include the floating-point unit. When the floating point unit is not included, all bits of the CPACR correctly read-as-zero (RAZ).
<b>Condition</b>	In an implementation of Cortex-R4F with the floating-point unit included, the CPACR is read.
<b>Implication(s)</b>	If software uses the CPACR to determine if Advanced SIMD functionality and registers D16-D31 are available, it will incorrectly determine that they are. Any attempt to use these features will lead to an unexpected UNDEFINED exception.
<b>Workaround(s)</b>	Because the Cortex-R4F processor never includes Advanced SIMD functionality or registers D16-D31, this can be correctly determined by reading the part number from one of the ID registers rather than examining ASEDIS and D32DIS in the CPACR.

**CORTEX-R4#56 (ARM ID-736960) *Debug Halt Exceptions Always Shown As Cancelling On ETM Interface***


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<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	When tracing program execution using the ETM, the instruction executed immediately before an external debug halt request should be traced.
<b>Issue</b>	However, when tracing program execution using the ETM, the instruction executed immediately before an external debug halt request will not be traced. If this instruction is a partially-executed multi-cycle instruction, for example a load-multiple which has transferred some, but not all of its registers, then this is correct. However, when the instruction has completed execution, this is erroneous.
<b>Condition</b>	<p>The erratum occurs if:</p> <ul style="list-style-type: none"> <li>• Tracing is enabled</li> <li>• The processor enters debug halt state either because EDBGRRQ was asserted or because of a write to the DRCR,</li> <li>• At the time it stopped executing instructions in order to enter debug halt state, the processor was not part-way through executing a load or store multiple instruction</li> </ul>
<b>Implication(s)</b>	Trace analysis tools will incorrectly consider the instruction executed immediately before the debug halt state entry to have not executed. If any of the ETM address comparators are configured to match on address of the instruction and the exact match bit is set, the comparator will fail to fire. This might have a knock-on effect to an expected trigger event or change in any ETM resources which are configured to be sensitive to the address comparator.
<b>Workaround(s)</b>	None.

**CORTEX-R4#57 (ARM ID-737195) Conditional VMRS APSR\_Nzcv, FPSCR May Evaluate With Incorrect Flags**

<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	A conditional VMRS APSR_nzcv, FPSCR instruction should evaluate its condition codes using the correct flags.
<b>Issue</b>	Under certain circumstances, a conditional VMRS APSR_nzcv, FPSCR instruction may evaluate its condition codes using the wrong flags and incorrectly execute or not execute.
<b>Condition</b>	<p>The erratum requires the following sequence of instructions in ARMstate: - VMRS&lt;c&gt; APSR_nzcv, FPSCR (formerly FMSTAT&lt;c&gt;), where the condition on the instruction is not always.</p> <ul style="list-style-type: none"> <li>• A flag-setting integer multiply or multiply and accumulate instruction (e.g. MULS)</li> <li>• A single-precision floating-point multiply-accumulate (FP-MAC) instruction (e.g. VMLA), timed such that the accumulate operation is inserted into the pipeline in the cycle in which the VMRS instruction is first attempted to be issued.</li> </ul> <p>To meet the above timing requirements, the VMRS instruction must be three pipeline stages behind the FPMAC. Depending on the rate in which the instructions are fetched, interlocks within this sequence and dual-issuing, this can be up to three other instructions between this pair, plus the multiply. Out-of-order completion of FP-MAC instructions must be enabled.</p>
<b>Implication(s)</b>	If this erratum occurs, the VMRS instruction will pass or fail its condition codes incorrectly, and this will appear in any trace produced by the ETM. This can corrupt the N, Z, C, V flag values in the CPSR which will typically affect the program flow.
<b>Workaround(s)</b>	This erratum can be avoided by disabling out-of-order single-precision floating point multiply-accumulate (SPMAC) instruction completion. Set DOOFMACS, bit [16] in the Secondary Auxiliary Control Register. This will have the side-effect of reducing the performance of SP-MAC operations, though the impact will depend on how these instructions are used in your code.

**CORTEX-R4#58 (ARM ID-726554) DBGDSCR.AdaDiscard Is Wrong When DBGDSCR.DBGack Set**

**Severity** 3-Medium

**Expected Behavior** When the DBGDSCR.ADAdiscard bit is set, asynchronous data aborts are discarded, except for setting the DBGDSCR.ADAabort sticky flag. The Cortex-R4 processor ensures that all possible outstanding asynchronous data aborts have been recognised before it enters debug halt state. The flag is immediately on entry to debug halt state to indicate that the debugger does not need to take any further action to determine whether all possible outstanding asynchronous aborts have been recognized.

**Issue** Because of this erratum, the Cortex-R4 processor also sets the DBGDSCR.ADAdiscard bit when the DBGDSCR.DBGack bit is set. This can cause the DBGDSCR.ADAabort bit to become set when the processor is not in debug halt state, and it is not cleared when the processor enters debug halt state. However, the processor does not discard the abort. It is pending or generates an exception as normal.

**Condition**

- The processor is not in debug halt state
- The DBGDSCR.DBGack bit is set
- An asynchronous data abort (for example, SLVERR response to a store to Normal-type memory) is recognized

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**NOTE:** it is not expected that DBGDSCR.DBGack will be set in any Cortex-R4 system

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**Implication(s)** If this erratum occurs, and the processor subsequently enters debug halt state, the DBGDSCR.ADAabort bit will be set, when in fact no asynchronous data abort has occurred in debug state. Before exiting debug state, the debugger will check this bit and will typically treat it as an error. If no other asynchronous data abort has occurred in debug state, this is a false error.

**Workaround(s)** None.

**CORTEX-R4#59 (ARM ID-748619) Missing Reset Exception On ETM Interface**

<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	Able to use the ETM to trace through a soft reset.
<b>Issue</b>	When tracing program execution through soft reset using the ETM, a reset exception might not be traced.
<b>Condition</b>	<p>Two sets of conditions cause this erratum to occur:</p> <p>Conditions set 1: The first instruction at the reset vector is a load-store multiple instruction. This must not happen because after reset there is no base register whose contents can be guaranteed.</p> <p>Conditions set 2: The processor enters debug halt state before executing the instruction at the reset vector due to the following conditions being true when the processor comes out of soft reset with nCPUHALTm HIGH or when nCPUHALTm is first de-asserted after soft reset:</p> <ul style="list-style-type: none"> <li>• The Halting mode debug enable bit (bit[14] in the DBGDSCR) is set AND</li> <li>• The DBGENm input pin is asserted AND</li> <li>• A debug event is triggered due to: - The Halt request bit in the DBGDRCR being set OR</li> <li>• The EDBGRQm input pin being asserted</li> </ul>
<b>Implication(s)</b>	<p>For conditions set 1:</p> <p>if the last instruction before the reset was an indirect branch instruction, a branch to the reset vector will be traced but not marked with a reset exception. If the last instruction before the reset was not an indirect branch, a trace analysis tool might incorrectly infer the execution of one or more instructions after the instruction before the reset until the processor executes an indirect branch. Typically, the instruction at the reset vector is an indirect branch and therefore this error is limited to one or two instructions. The address comparators in the ETM are unaffected unless an address comparison was set on the address of the instruction just before the reset occurs and the exact match bit was set for comparison. In this case the instruction is always considered to be executed.</p> <p>For conditions set 2:</p> <p>The reset exception is not traced and only the debug exception is traced. Any ETM address comparator configured to match on the instruction at the reset vector will not match.</p>
<b>Workaround(s)</b>	None.

**CORTEX-R4#61 (ARM ID-720270) Latched DTR-Full Flags Not Updated Correctly On DTR Access.**

<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	<p>When the debug Data Transfer Register (DTR) is in non-blocking mode, the latched DTR-full flags (RXfull_I and TXfull_I) record the state of the DTR registers as observed by the debugger and control the flow of data to and from the debugger to prevent race hazards. For example, when the target reads data from DBGDTRRXint, the associated flag RXfull is cleared to indicate that the register has been drained, but the latched value Rxfull_I remains set. Subsequent debugger writes to DBGDTRRXext are ignored because RXfull_I is set. RXfull_I is updated from RXfull when the debugger reads DBGDSCRext such that a debugger write to DBGDTRRXext will only succeed after the debugger has observed that the register is empty. The ARMv7 debug architecture requires that RXfull_I be updated when the debugger reads DBGDSCRext and when it writes DBGDTRRXext. Similarly, TXfull_I must be updated when the debugger reads DBGDSCRext and when it reads DBGDTRTXext.</p>
<b>Issue</b>	<p>Because of this erratum, RXfull_I and TXfull_I are only updated when the debugger reads DBGDSCRext.</p>
<b>Condition</b>	<p>The DTR is in non-blocking mode, that is, DBGDSCR.ExtDCCmode is set to 0b00 and EITHER:</p> <ul style="list-style-type: none"> <li>• The debugger reads DBGDSCRext which shows that RXfull is zero, that is, DBGDTRRX is empty.</li> <li>• The debugger writes data to DBGDTRRXext.</li> <li>• Without first reading the DBGDSCRext, and before the processor has read from DBGDTRRXint, the debugger performs another write to DBGDTRRXext.</li> </ul> <p>OR</p> <ul style="list-style-type: none"> <li>• The debugger reads DBGDSCRext which shows that TXfull is one, that is, DBGDTRTX is full.</li> <li>• The debugger reads data from DBGDTRTXext,</li> <li>• The processor writes new data into DBGDTRTXint,</li> <li>• Without first reading the DBGDSCRext, the debugger performs another read from DBGDTRTXext.</li> </ul>
<b>Implication(s)</b>	<p>The ARMv7 debug architecture requires the debugger to read the DBGDSCRext before attempting to transfer data via the DTR when in non-blocking mode. This erratum only has implications for debuggers that violate this requirement. If the erratum occurs via data transfer, data loss may occur. The architecture requires that data transfer never occur.</p> <p>Texas Instruments has verified that TI's Code Composer Studios IDE is not affected by this issue.</p>
<b>Workaround(s)</b>	None.

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**CORTEX-R4#66 (ARM ID-754269) Register Corruption During A Load-Multiple Instruction At An Exception Vector**

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<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	LDM will execute properly when used as the first instruction of an exception routine.
<b>Issue</b>	Under certain circumstances, a load multiple instruction can cause corruption of a general purpose register.
<b>Condition</b>	<p>All the following conditions are required for this erratum to occur:</p> <ul style="list-style-type: none"> <li>• A UDIV or SDIV instruction is executed with out-of-order completion of divides enabled</li> <li>• A multi-cycle instruction is partially executed before being interrupted by either an IRQ, FIQ or imprecise abort. In this case, a multi-cycle instruction can be any of the following: <ul style="list-style-type: none"> <li>– LDM/STM that transfers 3 or more registers</li> <li>– LDM/STM that transfers 2 registers to an unaligned address without write back</li> <li>– LDM/STM that transfers 2 registers to an aligned address with write back</li> <li>– TBB/TBH</li> </ul> </li> <li>• A load multiple instruction is executed as the first instruction of the exception handler</li> <li>• The load multiple instruction itself is interrupted either by an IRQ, FIQ, imprecise abort or external debug halt request. This erratum is very timing sensitive and requires the UDIV or SDIV to complete when the load multiple is in the Issue stage of the CPU pipeline. The register that is corrupted is not necessarily related to the load-multiple instruction and will depend on the state in the CPU store pipeline when the UDIV or SDIV completes.</li> </ul>
<b>Implication(s)</b>	For practical systems, it is not expected that an interruptible LDM will be executed as the first instruction of an exception handler, because the handler is usually required to save the registers of the interrupted context. Therefore, it is not expected that this erratum has any implications for practical systems. If the situation of the erratum occurs it will result in the corruption of the register bank state and could cause a fatal failure if the corrupted register is subsequently read before being written.
<b>Workaround(s)</b>	To work around this erratum, set bit [7] of the Auxiliary Control Register to disable out-of-order completion for divide instructions. Code performance may be reduced depending on how often divide operations are used.



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**CORTEX-R4#67 (ARM ID-758269) Watchpoint On A Load Or Store Multiple May Be Missed.**


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<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The Cortex-R4 supports synchronous watchpoints. This implies that for load and store multiples, a watchpoint on any memory access will generate a debug event on the instruction itself.
<b>Issue</b>	Due to this erratum, certain watchpoint hits on multiples will not generate a debug event.
<b>Condition</b>	<p>All the following conditions are required for this erratum to occur:</p> <ul style="list-style-type: none"> <li>• A load or store multiple instruction is executed with at least 5 registers in the register list</li> <li>• The address range accessed corresponds to Strongly-Ordered or Device memory</li> <li>• A watchpoint match is generated for an access that does not correspond to either the first two or the last two registers in the list.</li> </ul> <p>Under these conditions the processor will lose the watchpoint. Note that for a "store multiple" instruction, the conditions are also affected by pipeline state making them timing sensitive.</p>
<b>Implication(s)</b>	Due to this erratum, a debugger may not be able to correctly watch accesses made to Device or Strongly-ordered memory. The ARM architecture recommends that watchpoints should not be set on individual Device or Strongly-ordered addresses that can be accessed as part of a load or store multiple. Instead, it recommends the use of the address range masking functionality provided to set watchpoints on an entire region, ensuring that the watchpoint event will be seen on the first access of a load or store multiple to this region. If this recommendation is followed, this erratum will not occur.
<b>Workaround(s)</b>	None.

**AHB\_ACCES\_PORT#3 (ARM ID-529470)** — *DAP AHB-AP Access Port Might Fail To Return Slave Error On AHB Reset.*  
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**AHB\_ACCES\_PORT#3 (ARM ID-529470) DAP AHB-AP Access Port Might Fail To Return Slave Error On AHB Reset.**

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<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	If a system reset (nRST goes low) occurs while the AHB-AP is performing an access on the AHB bus, the AHB-AP should return a slave error back to the JTAG-DP.
<b>Issue</b>	Instead, the AHB-AP might indicate that the access completed successfully and return unpredictable data if the access was a read.
<b>Condition</b>	System reset is asserted LOW on a specific cycle while the AHB-AP is completing an access on the AHB bus. This is not an issue for a CPU only reset. This is not an issue during power-on reset (nPORRST) as the DAP will also be reset.
<b>Implication(s)</b>	Data read using the DAP while a system reset occurs may be corrupt, writes may be lost.
<b>Workaround(s)</b>	This is a workaround for users and tools vendors. Avoid performing debug reads and writes while the device might be reset.

<b>DMA#27</b>	<b><i>DMA Requests Lost During Suspend Mode</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	All DMA requests will be held while the device is in suspend mode.
<b>Issue</b>	DMA requests from a peripheral can be lost.
<b>Condition</b>	<p>This only happens when:</p> <ul style="list-style-type: none"> <li>• The device is suspended by a debugger</li> <li>• A peripheral continues to generate requests while the device is suspended</li> <li>• The DMA is setup to continue the current block transfer during suspend mode with DMA DEBUG MODE set to '01'</li> <li>• And the request trigger type TTYPE is set to frame trigger</li> </ul>
<b>Implication(s)</b>	When the DMA comes out of the suspend mode to resume the transfer, the data transfers corresponding to the third and subsequent requests will be lost.
<b>Workaround(s)</b>	Either use TTYPE = Block transfer when DMA DEBUG MODE is '01' (Finish Current Block Transfer) or use DMA DEBUG MODE = '00' (Ignore suspend) when using TTYPE = Frame transfer to complete block transfer even after suspend/halt is asserted.

<b>DMM#16</b>	<b><i>BUSY Flag Not Set When DMM Starts Receiving A Packet</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The BUSY flag in the DMMGLBCTRL register should be set when DMM starts receiving a packet or has data in its internal buffers.
<b>Issue</b>	However, the BUSY flag in the DMMGLBCTRL register is not set when DMM starts receiving a packet.
<b>Condition</b>	The BUSY bit is set only after the packet has been received, deserialized and written to the internal buffers. It stays active while data is still in the DMM internal buffers. If the internal buffers are empty (this means that no data needs to be written to the destination memory) then the BUSY bit will be cleared.
<b>Implication(s)</b>	<p>Care needs to be taken when turning the DMM module off (ON/OFF = 0101). The DMM module will still finish the reception and data transmission to the destination memory if it has been programmed to the off state during an ongoing reception. The BUSY flag will not be set while this reception on the external DMM interface is in progress and all internal buffers are empty.</p> <p>Depending on the module configuration and the packet width it may take a different number of DMMCLK cycles before the BUSY flag is set.</p> <p>For example in Trace Mode the maximum packet size = 88</p> <ul style="list-style-type: none"> <li>• port width = 1, it takes 91 DMMCLK cycles to receive and deserialize the packet</li> <li>• port width = 16, it takes 9 DMMCLK cycles to receive and deserialize the packet</li> </ul>
<b>Workaround(s)</b>	Wait for a number of DMMCLK cycles (e.g. 95 DMMCLK cycles) beyond the longest reception and deserialization time needed for a given packet size and DMM port configuration before checking the status of the busy flag, after the DMM ON/OFF register field has been programmed to OFF.

<b>EMIF#3</b>	<b><i>EMIF Reports Incorrect Time-out Error on Register Read</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The EMIF should not cause an abort when accessing EMIF registers.
<b>Issue</b>	After an abort because an external asynchronous memory failed to respond, a read to an EMIF register will also generate an abort.
<b>Condition</b>	<ol style="list-style-type: none"> <li>1. The EMIF is used for asynchronous memory accesses in Extended Wait mode.</li> <li>2. A time-out error occurs. For example, the memory does not de-assert the EMIF_nWAIT input.</li> <li>3. The asynchronous memory access with time-out error is followed by an EMIF register read.</li> </ol>
<b>Implication(s)</b>	Aborts will be generated on EMIF register reads until the "time-out" status is corrected by a successful EMIF region read.
<b>Workaround(s)</b>	If a timeout error occurs, perform a dummy read from the EMIF region that does not return an error. This can be a synchronous read, a read from another asynchronous chip select that is not configured to be in Extended Wait mode, or to the same asynchronous chip select after disabling the Extended Wait mode on that chip select.

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**ERAY#52 (FLEXRAY#52) WUS Generates Redundant SIR.WUPA/B Events**


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<b>Severity</b>	4-Low
<b>Expected Behavior</b>	If a sequence of wakeup symbols (WUS) is received and all are separated by appropriate idle phases then a valid wakeup pattern (WUP) should be detected after <i>every second WUS</i> .
<b>Issue</b>	Instead, the FlexRay module detects a valid wakeup pattern after the second WUS and then after each following WUS.
<b>Condition</b>	A sequence of wakeup symbols (WUS) is received. All separated by appropriate idle phases.
<b>Implication(s)</b>	More SIR.WUPA/B events are seen than expected especially when an application program frequently resets the appropriate SIR.WUPA/B bits
<b>Workaround(s)</b>	Ignore redundant SIR.WUPA/B events.

**ERAY#58 (FLEXRAY#58) *Erroneous Cycle Offset During Startup after abort of startup or normal operation***


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<b>Severity</b>	4-Low
<b>Expected Behavior</b>	Correct cycle offset inspite of abort of startup or normal operation by a READY command.
<b>Issue</b>	The state INITIALIZE_SCHEDULE may be one macrotick too short during an integration attempt. This leads to an early cycle start in state INTEGRATION_COLDSTART_CHECK or INTEGRATION_CONSISTENCY_CHECK.
<b>Condition</b>	An abort of startup or normal operation by a READY command near the macrotick border. The erratum is limited to applications where READY command is used to leave STARTUP, NORMAL_ACTIVE, or NORMAL_PASSIVE state
<b>Implication(s)</b>	<p>As a result the integrating node calculates a cycle offset of one macrotick at the end of the first even/odd cycle pair in the states INTEGRATION_COLDSTART_CHECK or INTEGRATION_CONSISTENCY_CHECK and tries to correct this offset.</p> <p>If the node is able to correct the offset of one macrotick (<math>pOffsetCorrectionOut &gt;&gt; gdMacrotick</math>), the node enters NORMAL_ACTIVE with the first startup attempt.</p> <p>If the node is not able to correct the offset error because <math>pOffsetCorrectionOut</math> is too small (<math>pOffsetCorrectionOut \leq gdMacrotick</math>), the node enters ABORT_STARTUP and is ready to try startup again. The next (second) startup attempt is not affected by this erratum.</p>
<b>Workaround(s)</b>	With a configuration of $pOffsetCorrectionOut >> gdMacrotick * (1 + cClockDeviationMax)$ the node will be able to correct the offset and therefore also be able to successfully integrate.

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**ERAY#59 (FLEXRAY#59) First WUS Following Received Valid WUP May Be Ignored**


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<b>Severity</b>	4-Low
<b>Expected Behavior</b>	The FlexRay controller protocol engine should recognize all wakeup symbols (WUS).
<b>Issue</b>	However, the protocol engine may ignore the first wakeup symbol (WUS) following the below stated state transition therefore signalling the next SIR.WUPA/B at the third WUS instead of the second WUS.
<b>Condition</b>	The erratum is limited to the reception of redundant wakeup patterns. When the protocol engine is in state WAKEUP_LISTEN and receives a valid wakeup pattern (WUP), it transfers into state READY and updates the wakeup status vector CCSV.WSV[2:0] as well as the status interrupt flags SIR.WST and SIR.WUPA/B.
<b>Implication(s)</b>	Delayed setting of status interrupt flags SIR.WUPA/B for redundant wakeup patterns.
<b>Workaround(s)</b>	None.



**ERAY#60 (FLEXRAY#60) *READY Command Accepted In READY State***

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<b>Severity</b>	4-Low
<b>Expected Behavior</b>	The FlexRay module should ignore a READY command while in READY state.
<b>Issue</b>	However, the FlexRay module does not ignore a READY command while in READY state.
<b>Condition</b>	The erratum is limited to the READY state.
<b>Implication(s)</b>	Flag CCSV.CSI is set. Cold starting needs to be enabled by the Protocol Operation Controller (POC) command ALLOW_COLDSTART (SUCC1.CMD = "1001").
<b>Workaround(s)</b>	None.

**ERAY#61 (FLEXRAY#61) Slot Status vPOC!SlotMode Is Reset Immediately When Entering HALT State**


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<b>Severity</b>	4-Low
<b>Expected Behavior</b>	According to the FlexRay protocol specification, the slot mode should not be reset to SINGLE slot mode before the following state transition from HALT to DEFAULT_CONFIG state.
<b>Issue</b>	However, when the protocol engine is in NORMAL_ACTIVE or NORMAL_PASSIVE state, a HALT or FREEZE command issued by the CPU resets vPOC!SlotMode immediately to SINGLE slot mode (CCSV.SLM[1:0] = "00").
<b>Condition</b>	The erratum is limited to the HALT state.
<b>Implication(s)</b>	The slot status vPOC!SlotMode is reset to SINGLE when entering HALT state.
<b>Workaround(s)</b>	None.

<b>ETM_R4#16</b>	<b><i>ETM-R4 Fails To Trace VNT Packet For The Second Half Of SWP Instruction</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	When tracing SWP or SWPB instructions, the load and store parts of the SWP or SWPB instruction should be traced with separate data packets to the same address. If the load transfer is traced and the store transfer is not traced, the store transfer should be traced with a "Value Not Traced" packet.
<b>Issue</b>	ETM-R4 Fails To Trace Value Not Traced (VNT) Packet For The Second Half Of SWP Instruction. No trace is generated for the store transfer of the SWP or SWPB.
<b>Condition</b>	<p>The following conditions must occur:</p> <ul style="list-style-type: none"> <li>• The ETM is enabled and is tracing</li> <li>• A SWP or SWPB instruction is executed</li> <li>• ViewData is configured to only trace the load part of the SWP or SWPB instruction</li> </ul>
<b>Implication(s)</b>	If the ETM traces any data transfer, a data packet must be traced for every subsequent data transfer for that same instruction. This allows trace analysis tools to determine which registers were used or updated by the traced data items. When this erratum occurs, only the load part of the SWP or SWPB instruction is traced and therefore analysis tools cannot determine if the transfer is the load or store part of the SWP or SWPB instruction. This might cause misinterpretation of the execution of the processor by the analysis tool. The trace stream is not corrupted.
<b>Workaround(s)</b>	<p>The following workarounds are for users or tool vendors:</p> <ul style="list-style-type: none"> <li>• Ensure that for all SWP and SWPB instructions in your code ViewData is not configured to trace load data only</li> <li>• If ViewData has been configured to trace only the load transfer of a SWP or SWPB instruction and a single transfer has been traced, the trace analyser can assume that this corresponds to the load part of the instruction</li> </ul>

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**FMC#67 (FLASH WRAPPER#67) Error Status Register Bit B2\_COR\_ERR Set Erroneously**


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<b>Severity</b>	4-Low
<b>Expected Behavior</b>	The Error status register bit B2_COR_ERR is normally set when a correctable error occurs on Bus 2 access to OTP, TI-OTP or ECC. It should not be set during correctable error profiling.
<b>Issue</b>	The Error status register bit B2_COR_ERR bit also gets set along with the the ERR_PRF_FLG bit when doing correctable error profiling.
<b>Condition</b>	<ol style="list-style-type: none"> <li>1. In correctable error profiling mode</li> <li>2. Either the EOFEN (error "one" fail enable) or EZFEN (error "zero" fail enable) bit is set</li> <li>3. The wrapper has already found a bus 1 (CPU) correctable error</li> <li>4. A correctable error occurs on a bus 2 which causes the correctable error count to reach the threshold.</li> </ol>
<b>Implication(s)</b>	It is not expected that the customer would enable error profiling and EOFEN or EZFEN. However, if they do, the CPU reads the status register after the bus 2 error and it looks like the bus 2 caused the error, but the error address and position register point to a bus 1 location, creating confusion.
<b>Workaround(s)</b>	If profiling errors, do not enable EOFEN or EZFEN.

<b>FMC#79</b>	<b><i>Abort on Unaligned Access at End of Bank</i></b>
<b>Severity</b>	4-Low
<b>Expected Behavior</b>	The Cortex R4 can make unaligned accesses to any memory within the ATCM space (program flash space).
<b>Issue</b>	The CortexR4 CPU will sometimes get an abort when making an unaligned access near the physical end of the bank boundary (in the range from 0xn timer FFF1 through 0xn timer FFFF). The unaligned access can be from a data read such as a LDR or an LDRH instruction, or from fetching a 32-bit thumb2 instruction which is not aligned on a four byte boundary.
<b>Condition</b>	This only occurs in the program flash on the ATCM. It only occurs when the flash is in single cycle mode and operating above 20MHz speed. This is more likely to occur with high Vcc-core and at high temperature.
<b>Implication(s)</b>	The CPU could take either an instruction abort or a data abort.
<b>Workaround(s)</b>	<p>Use an option to keep the compiler from generating unaligned data or instructions. For the TI compiler use --unaligned_access=off. Also ensure that hand generated assembly language routines do not create an unaligned access to these locations.</p> <p>OR</p> <p>Do not use single cycle mode (RWAIT=0) at frequencies above 20MHz.</p> <p>OR</p> <p>Reserve the last fifteen bytes of flash in each bank on the ATCM with either a dummy structure that is not accessed, or with a structure that will not create an unaligned access.</p>

<b>FMC#80</b>	<b><i>Abort on acceses switching between two banks</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	No aborts occur when accessing flash from either bank within the device's valid memory space.
<b>Issue</b>	An abort is sometimes generated when a memory access (either instruction fetch or data read) switches from one bank to the other.
<b>Condition</b>	<p>This only occurs in single cycle mode at HCLK frequenies greater than 20MHz.</p> <p>This only occurs on devices with at least two banks of flash in the main memory (TCM) and at least one bank is not a size that can be expressed by <math>2^n</math>.</p> <p>Examples:</p> <p>A 3MB device that is comprised of two 1.5MB flash banks may generate an abort if an access to bank 0 is followed by an access between 1.5MB and 2MB (in bank 1). Likewise, the device may generate an abort if an access to bank 1 is followed by an access between 1MB and 1.5MB (in bank 0).</p> <p>A 2MB device that is comprised of bank 0 with 1.5MB of flash and bank 1 with 0.5MB flash may generate an abort if an access to bank 0 is followed by an access to bank 1.</p> <p>A 2MB device that is comprised of bank 0 with 0.5MB of flash and bank 1 with 1.5MB flash may generate an abort if an access to bank 1 is followed by an access to bank 0.</p>
<b>Implication(s)</b>	An abort exception is taken
<b>Workaround(s)</b>	Do not operate in single cycle mode above 20MHz in speed.

<b>FTU#8</b>	<b><i>FlexRay Transfer Unit Not Disabled On Memory Protection Violation (MPV) Error</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	On memory protection violation (MPV) errors the FTU should get disabled.
<b>Issue</b>	The FTU does not get disabled under some conditions.
<b>Condition</b>	<p>If an MPV error occurs during the following transfer scenarios:</p> <ul style="list-style-type: none"> <li>• During header transfer from system memory to FlexRay RAM, when FTU is configured to transfer header and payload</li> <li>• During payload transfer from FlexRay RAM to system memory, when FTU is configured to transfer header and payload</li> <li>• During a transfer from FlexRay RAM to system memory, when FTU is configured to transfer payload only</li> </ul>
<b>Implication(s)</b>	The MPV error flag in the Transfer Error Interrupt Flag (TIEF) register is set, but the TUE flag in the Global Control Set/Reset (GCS/R) register does not get cleared. As a result, the FTU does not get disabled.
<b>Workaround(s)</b>	<p>This errata can be avoided in the following ways:</p> <ul style="list-style-type: none"> <li>• For transfers from system memory to FlexRay RAM, transfer the payload only</li> <li>• Generate an MPV interrupt and clear the TUE flag in the Global Control Set/Reset (GCS/R) register in the interrupt service routine</li> </ul>

<b>FTU#19</b>	<b><i>TCCOx Flag Clearing Masked</i></b>
<b>Severity</b>	4-Low
<b>Expected Behavior</b>	According to the documentation, when TOOFF (Transfer Occurred Offset) is read, the corresponding message flag in TCCOx must be cleared.
<b>Issue</b>	In some conditions, the read of TOOFF register would not be up to date and would not reflect the last buffer completed.q
<b>Condition</b>	There may be a timing condition when TCCOx flag clearing could be masked due to the state machine clearing of TTCCx (Trigger transfer to communication controller) within the same cycle as software reading TOOFF.
<b>Implication(s)</b>	The TCCOx flag is not being cleared.
<b>Workaround(s)</b>	After reading the TOOFF to determine the highest buffer completed, clear the corresponding flag in TCCOx.



<b>LPO#16</b>	<b><i>Oscillator Fault Detection Starts too Soon</i></b>
<b>Severity</b>	4-Low
<b>Expected Behavior</b>	The oscillator fault detection circuit allows approximately 200ms after the release of nPORRST for the oscillator to become valid before monitoring for oscillator faults.
<b>Issue</b>	Sometimes the oscillator fault detection starts monitoring the oscillator shortly (approximately 100us) after the release of nPORRST.
<b>Condition</b>	This occurs only after power on.
<b>Implication(s)</b>	If the oscillator is slow to start, an oscillator fault condition may be detected and the device will switch to using the HFLPO as a clock source.
<b>Workaround(s)</b>	<p>This condition is avoided if nPORRST is held low long enough for the oscillator to stabilize.</p> <p>This condition can be corrected by software. The software should check for oscillator fault after detecting a power-on reset. If an oscillator fault has occurred, the software can attempt to restart the oscillator. Refer to the device Technical Reference Manual for information on how to recover from an oscillator failure.</p>

**MCRC#18** — *CPU Abort Generated on Write to Implemented MCRC Space After Write to Unimplemented MCRC Space*  
[www.ti.com](http://www.ti.com)

<b>MCRC#18</b>	<b><i>CPU Abort Generated on Write to Implemented MCRC Space After Write to Unimplemented MCRC Space</i></b>
<b>Severity</b>	4-Low
<b>Expected Behavior</b>	A write to the legal address region of the MCRC module should not generate an abort
<b>Issue</b>	An Unexpected CPU Data Abort is generated in the following condition:
<b>Condition</b>	When a normal mode write to an illegal address region of MCRC register space is followed by a write to a legal address region of the MCRC register space.
<b>Implication(s)</b>	A write to an illegal address region of the MCRC register space generates a data abort as expected. The next write to a legal address region of the MCRC register space generates an unexpected second data abort.
<b>Workaround(s)</b>	None.

<b>MIBSPI#110</b>	<b><i>Multibuffer Slave In 3 or 4 Pin Mode Transmits Data Incorrectly for Slow SPICLK Frequencies and for Clock Phase = 1</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The SPI must be able to transmit and receive data correctly in slave mode as long as the SPICLK is slower than the maximum frequency specified in the device datasheet.
<b>Issue</b>	The MibSPI module, when configured in multi-buffered slave mode with 3 functional pins (CLK, SIMO, SOMI) or 4 functional pins (CLK, SIMO, SOMI, nENA), could cause incorrect data to be transmitted.
<b>Condition</b>	<p>This issue can occur under the following condition:</p> <ul style="list-style-type: none"> <li>• Module is configured to be in multi-buffered mode, AND</li> <li>• Module is configured to be a slave in the SPI communication, AND</li> <li>• SPI communication is configured to be in 3-pin mode or 4-pin mode with nENA, AND</li> <li>• Clock phase for SPICLK is 1, AND</li> <li>• SPICLK frequency is VCLK frequency / 12 or slower</li> </ul>
<b>Implication(s)</b>	Under the above described condition, the slave MibSPI module can transmit incorrect data.
<b>Workaround(s)</b>	<p>The issue can be avoided by setting the CSHOLD bit in the control field of the TX RAM. The nCS is not used as a functional signal in this communication, hence setting the CSHOLD bit does not cause any other effect on the SPI communication.</p>

<b>MIBSPI#111</b>	<b><i>Data Length Error Is Generated Repeatedly In Slave Mode when I/O Loopback is Enabled</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	After a data length (DLEN) error is generated and the interrupt is serviced the SPI should abort the ongoing transfer and stop.
<b>Issue</b>	When a DLEN error is created in Slave mode of the SPI using nSCS pins in IO LoopBack Test mode, the SPI module re-transmits the data with the DLEN error instead of aborting the ongoing transfer and stopping.
<b>Condition</b>	This is only an issue for a IOLPBK mode Slave in Analog Loopback configuration, when the intentional error generation feature is triggered using CTRL_DLENERR(IOLPBKTSTCR.16).
<b>Implication(s)</b>	The SPI will repeatedly transmit the data with the DLEN error when configured in the above configuration.
<b>Workaround(s)</b>	After the DLEN_ERR interrupt is detected in IOLPBK mode, disable the transfers by clearing the SPIEN bit of SPIGCR1 register (bit 24) and then re-enable the transfers by setting SPIEN.

<b>MIBSPI#137</b>	<b><i>Spurious RX DMA REQ from a Slave mode MIBSPI</i></b>
<b>Severity</b>	4-Low
<b>Expected Behavior</b>	The MIBSPI should not generate unexpected DMA requests when it did not receive data from the master
<b>Issue</b>	A spurious DMA request is generated even when the SPI slave is not transferring data.
<b>Condition</b>	<p>This errata is only valid when all below conditions are true:</p> <ul style="list-style-type: none"> <li>• The MIBSPI is configured in compatible mode (SPI) as a slave.</li> <li>• SPIINT0.16 (DMA_REQ_EN) bit is set to enable DMA requests.</li> <li>• The nSCS (Chip Select) pin is in active state, but no transfers are active.</li> <li>• The SPI is disabled by clearing SPIGCR1.24 (SPIEN) bit from '1' to '0'.</li> </ul> <p>The above sequence triggers a false request pulse on the Receive DMA Request as soon as SPIEN bit is cleared from '1' to '0'.</p>
<b>Implication(s)</b>	The SPI generates a false DMA request to the DMA module when the data is not yet available for the DMA module to retrieve.
<b>Workaround(s)</b>	Whenever the SPI is to be disabled by clearing SPIEN bit, clear the DMA_REQ_EN bit to '0' first and then clear the SPIEN bit.

<b>MIBSPI#139</b>	<b><i>Mibspi RX RAM RXEMPTY bit does not get cleared after reading</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The MibSPI RXEMPTY flag is auto-cleared after a CPU or DMA read.
<b>Issue</b>	Under a certain circumstance, the RXEMPTY flag is not auto-cleared after a DMA/CPU read.
<b>Condition</b>	<p>The TXFULL flag of the latest buffer that the sequencer read out of TXRAM for the currently active TG is 0; And,</p> <p>A higher priority TG interrupts the current TG and the sequencer starts to read the first buffer of the new TG from the TXRAM; And,</p> <p>Simultaneously, the host (CPU/DMA) is reading out an RXRAM location that contains valid RX data from the previous transfers.</p>
<b>Implication(s)</b>	<p>The fake RXEMPTY '1' suspends the next Mibspi transfer with BUFMODE 6 or 7.</p> <p>With other BUFMODEs, a false "Receive data buffer overrun" will be reported for the next Mibspi transfer.</p>
<b>Workaround(s)</b>	<ol style="list-style-type: none"> <li>1. Use one transfer group, or,</li> <li>2. Keep the Tx buffer full</li> </ol>

<b>NHET#52</b>	<b><i>WCAP May not Capture the High-Resolution Offset Correctly</i></b>
<b>Severity</b>	2-High
<b>Expected Behavior</b>	The N2HET is a complex timer that operates by executing a set of timing instructions each loop resolution period. Normally counters and capture registers implemented by such a scheme would be limited in resolution to the timer's loop period; but the N2HET also includes dedicated hardware timer extensions that enable operation with resolution as high as 1/128 of the timer loop period. The WCAP instruction is designed to support time-stamping with high resolution. When an event occurs, the WCAP instruction is supposed to capture both the loop-resolution counter value and the high-resolution offset measuring the interval between the start of the loop and actual event occurrence, and return the combined result. This result can be interpreted as the number of timer loops represented as fixed point number with 25 integral bits and 7 fractional bits.
<b>Issue</b>	WCAP may not capture the high-resolution offset correctly.
<b>Condition</b>	If the edge event that the WCAP instruction is time-stamping happens to align with the start of the internal loop resolution period, then the WCAP does not properly capture the 7-bit fractional offset correctly
<b>Implication(s)</b>	Under the above described condition, the fractional offset of the previous event captured by WCAP is what is stored in the WCAP data field. Because of this boundary condition, the WCAP instruction should not be used in high-resolution mode
<b>Workaround(s)</b>	<p>A similar instruction, PCNT, is known to operate correctly at high resolution. While WCAP returns the absolute time-stamp of an event, PCNT returns the interval between an event and the previous event on the same pin.</p> <p>Instead of WCAP, if an absolute time-stamp is required it is recommended to use a PCNT instruction followed by an ADD instruction that accumulates the time between successive edge measurements to produce a result that is a WCAP equivalent.</p> <p>An example for using PCNT as a work around for WCAP:</p> <p>WCAP {next = ERRPIN, cond_addr = JMP, hr_lr=high, reg=T, event=RISE, pin=16, data=0}</p> <p>Work around:</p> <p>SAVE: MOV32 {remote=TRISE, type=REMTOREG, reg=R}</p> <p>TRISE: PCNT {next=TIMESTAMP, hr_lr=high, type=RISE2RISE, pin=16, data=0}</p> <p>TIMESTAMP: ADD {src1=REM, src2=R, rdest=REM, remote=TRISE, dest=NONE, next=CHK, data=0, hr_data=0}</p> <p>CHK: BR {cond_addr=JMP, event=RISE, pin=20,next=ERRPIN}</p>

<b>NHET#53</b>	<b><i>Enhanced Input Capture Pins May not Capture Small Pulses Correctly</i></b>
<b>Severity</b>	2-High
<b>Expected Behavior</b>	The PCNT and WCAP instructions can capture pulse length or time stamp of small pulses that have two edges within a single loop resolution.
<b>Issue</b>	The high resolution value may be captured incorrectly.
<b>Condition</b>	If the second edge event that the PCNT or WCAP instruction is using happens to align with the start of the internal loop resolution period, then the instruction does not properly capture the high resolution value correctly.
<b>Implication(s)</b>	Because of this boundary condition, the PCNT and WCAP instructions should not be used on small pulses.
<b>Workaround(s)</b>	None



<b>PBIST#4</b>	<b><i>PBIST ROM Algorithm Doesn't Execute</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	PBIST controller downloads algorithms from PBIST ROM to PBIST RAM and executes.
<b>Issue</b>	It is possible that the PBIST algo will not download from PBIST ROM to PBIST RAM.
<b>Condition</b>	The possibility that PBIST algo will not download only occurs the first time PBIST is executed after power on reset.
<b>Implication(s)</b>	The PBIST test may return with a pass status, even though the algo was not properly executed.
<b>Workaround(s)</b>	Ignore the first PBIST result after power up and re-run the PBIST Test. The second execution of the PBIST test will provide correct results

<b>SSWF021#35</b>	<b><i>Potential clock glitch when switching PLL clock divider from divide by 1.</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	User should be able to switch the clock divider without generating an internal clock glitch.
<b>Issue</b>	When switching the PLL output clock divider from divide by 1 to any other ratio a clock glitch may be generated. Unpredictable results can occur as a result of switching the PLL clock divider from divide by 1.
<b>Condition</b>	Directly switching the PLL clock divider from divide by 1 while that PLL is the clock source.
<b>Implication(s)</b>	The result of generating a clock glitch while the PLL is the clock source is unpredictable.
<b>Workaround(s)</b>	Switch all clock domains to a different source (such as oscillator) before switching the PLL output clock divider from divide by 1 to a larger divider, then switch back to using the PLL as source.

<b>STC#26</b>	<b><i>STCTPR is Reset at the End of Each Self Test</i></b>
<b>Severity</b>	4-Low
<b>Expected Behavior</b>	Once the STC Timeout-counter Preload Register (STCTPR) is written to, this value will be used to preload the timeout-counter for each self test run.
<b>Issue</b>	The STCTPR is reset at the end of each CPU self test run.
<b>Condition</b>	None
<b>Implication(s)</b>	Subsequent self test runs will use a maximum timeout value of 0xFFFFFFFF
<b>Workaround(s)</b>	The Timeout preload value in STCTPR register needs to be programmed to the required time out value before starting a self test every time if a timeout count of other than 0xFFFFFFFF is desired.

<b>SYS#46</b>	<b><i>Clock Source Switching Not Qualified With Clock Source Enable And Clock Source Valid</i></b>
<b>Severity</b>	4-Low
<b>Expected Behavior</b>	An attempt to switch to a clock source which is not valid yet should be discarded.
<b>Issue</b>	Switching a clock source by simply writing to the GHVSRC bits of the GHVSRC register may cause unexpected behavior. The clock will switch to a source even if the clock source was not ready.
<b>Condition</b>	The clock source is enabled (CSDIS register) but the clock source is not yet valid (CSVSTAT register).
<b>Implication(s)</b>	Unexpected behavior stated above.
<b>Workaround(s)</b>	Always check the CSDIS register to make sure the clock source is turned on and check the CSVSTAT register to make sure the clock source is valid. Then write to GHVSRC to switch the clock.

<b>SYS#102</b>	<b><i>Bit field EFUSE_Abort[4:0] in SYSTASR register is read-clear</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	The documentation states that EFUSE_Abort[4:0] of the SYSTASR register should be write-clear in privilege mode.
<b>Issue</b>	However, these bits are implemented as read-clear.
<b>Condition</b>	Always.
<b>Implication(s)</b>	Software implementation for error handling needs to take care of this.
<b>Workaround(s)</b>	None

<b>SYS#111</b>	<b><i>The Device may Generate Multiple nRST Pulses.</i></b>
<b>Severity</b>	3-Medium
<b>Expected Behavior</b>	When an event on the device causes a system reset to be asserted, the device drives the system reset (nRST) terminal low for 8 VCLK periods. Then the nRST terminal is released and gets pulled High, allowing the device to start its warm boot sequence.
<b>Issue</b>	It is possible that the device may generate multiple system resets instead of releasing the system reset (nRST) at the appropriate time and starting the warm boot sequence.
<b>Condition</b>	This issue occurs when either the external reset pulse or the 8 VCLK internally generated reset pulse is the same length as the delay caused by the glitch filter on the nRST signal. The delay time of the glitch filter is a function of device process, the temperature and the core voltage. This has a range from 500ns to 2s. The length of the 8 VCLK reset extension is a function of the crystal speed, the I/O voltage, and the resistance and capacitance on the nRST pin. There is no problem when reset is properly asserted with the PORRST pin for 1ms or more as the nRST pin will be held low much longer than 2us.
<b>Implication(s)</b>	Usually only one or two extra reset pulses are generated. However, it is possible, in rare cases, that a long string of reset pulses could be generated. This would prevent the device from booting up in a timely manner.
<b>Workaround(s)</b>	In the case of an externally generated system reset, this issue can be avoided by asserting the system reset signal for more than 2s. If the issue is seen when there is an internal system reset condition then the issue can be avoided by using an 8MHz or slower crystal. Another workaround is to avoid the use of internal system resets such as software reset, oscillator fault reset, watchdog reset, or debug reset. This erratum has been fixed on newer devices by extending the time of the internally generated nRSTpulse to 32 VCLK cycles. This extends the reset pulse to a minimum time of 3.2us (with a 20MHz crystal), which exceeds the maximum delay time of the glitch filter.

<b>VIM#27</b>	<b><i>Unexpected phantom interrupt</i></b>
<b>Severity</b>	2-High
<b>Expected Behavior</b>	When responding to an interrupt and a subsequent interrupt is received, the corresponding VIM request should be flagged as pending in the VIM status registers. When the CPU is ready to service the subsequent interrupt, the correct service routine address should be fetched by the CPU.
<b>Issue</b>	On rare occasions the VIM may return the phantom interrupt vector instead of the real interrupt vector.
<b>Condition</b>	This condition is specific to software and hardware vectored modes. This is not applicable for legacy interrupt servicing mode. This condition occurs when the ratio of GCLK to VCLK is 3:1 or greater for hardware vectored mode, or the ratio of GCLK to VCLK is 5:1 or greater for software vectored mode. A subsequent interrupt request must occur when the VIM is finishing acknowledging a previous interrupt.
<b>Implication(s)</b>	The subsequent interrupt request vectors to the phantom interrupt routine instead of the correct service routine.
<b>Workaround(s)</b>	This issue can be avoided if the clock ratio GCLK:VCLK is 1:1 or 2:1. Otherwise the VIM status register can be checked within the phantom interrupt routine to determine the source of the subsequent interrupt.

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No TI components are authorized for use in FDA Class III (or similar life-critical medical equipment) unless authorized officers of the parties have executed a special agreement specifically governing such use.

Only those TI components which TI has specifically designated as military grade or "enhanced plastic" are designed and intended for use in military/aerospace applications or environments. Buyer acknowledges and agrees that any military or aerospace use of TI components which have **not** been so designated is solely at the Buyer's risk, and that Buyer is solely responsible for compliance with all legal and regulatory requirements in connection with such use.

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