

Dual 14-bit 500 MSPS Digital to Analog Converter

FEATURES

- Dual Channel
- 14-Bit Resolution
- Maximum Sample Rate: 500 MSPS
- Pin Compatible with DAC3154/DAC3164 and DAC3151/DAC3161/DAC3171
- Input Interface:
 - 14 LVDS Inputs
 - Single 14-bit wide interface or Dual 7-bit wide interface
 - Single or dual DDR data clock
 - Internal FIFO
- Chip to Chip Synchronization
- Power Dissipation: 460mW
- Spectral Performance at 20 MHz IF
 - SNR: 76 dBFS
 - SFDR: 78 dBc
- Current sourcing DACs
- Compliance Range: –0.5V to 1V
- Package: 64 pin QFN (9x9mm)

APPLICATIONS

- Multi-carrier, Multi-mode Cellular Infrastructure Base Stations
- Radar
- Signal Intelligence
- Software-defined Radio
- Test and Measurement Instrumentation

DESCRIPTION

The DAC3174 is a dual channel 14-bit, 500 MSPS digital-to-analog converter (DAC). The DAC3174 uses a 14-bit wide LVDS digital bus with 1 or 2 independent data clocks for flexibility in providing each channel's data from different data sources. An input FIFO allows independent data and sample clocks. FIFO input and output pointers can be synchronized across multiple devices for precise signal synchronization. The DAC outputs are current sourcing and terminate to GND with a compliance range of –0.5 to 1V. DAC3174 is pin compatible with the dual-channel, 12-/10-bit, 500 MSPS digital-to-analog converter DAC3164/DAC3154 and single-channel, 14-/12-/10-bit, 500 MSPS digital-to-analog converter DAC3171/DAC3161/DAC3151.

The device is available in a QFN-64 PowerPAD™ package is specified over the full industrial temperature range (–40°C to 85°C).



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PowerPAD is a trademark of Texas Instruments.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of the Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

Copyright © 2013, Texas Instruments Incorporated

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

BLOCK DIAGRAMS

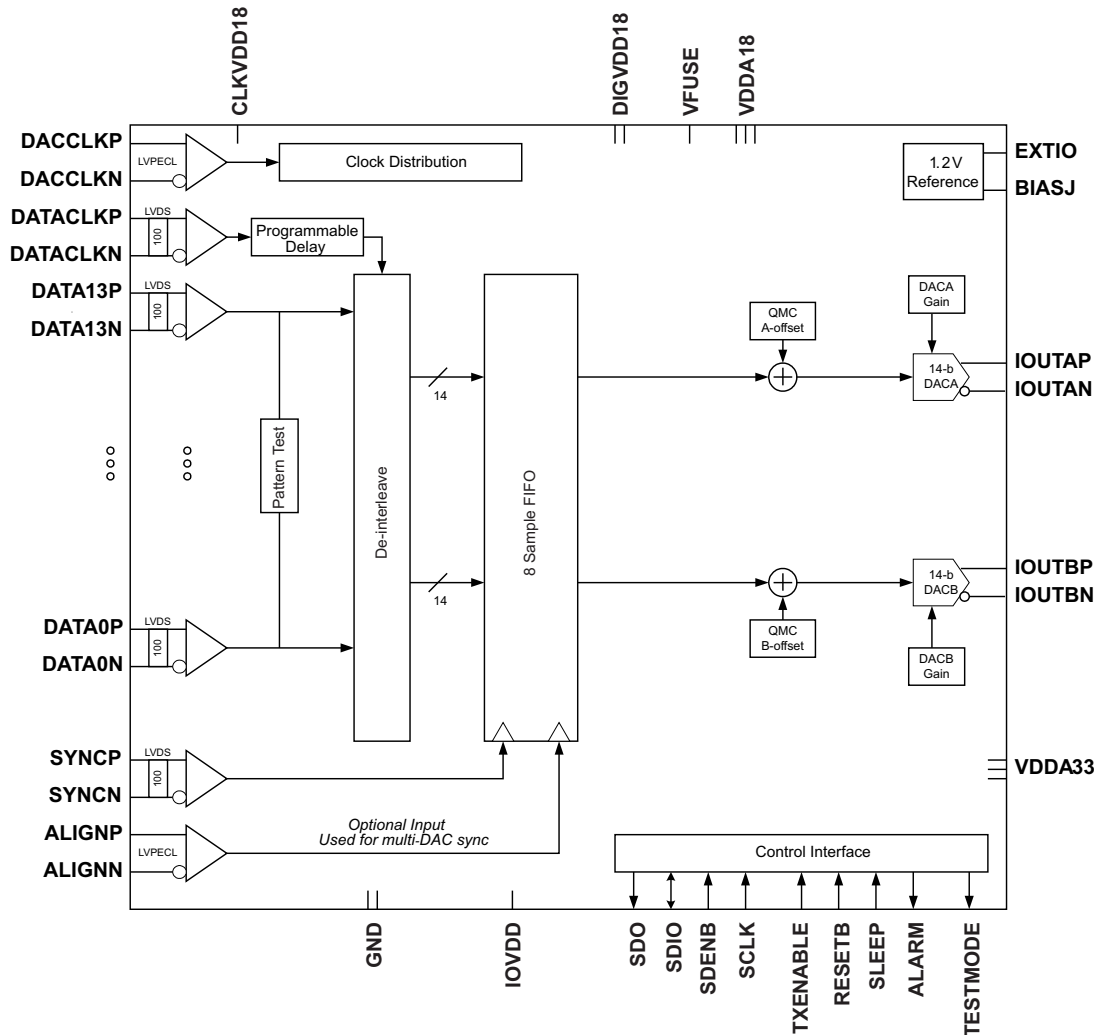


Figure 1. 14-Bit Interface Mode

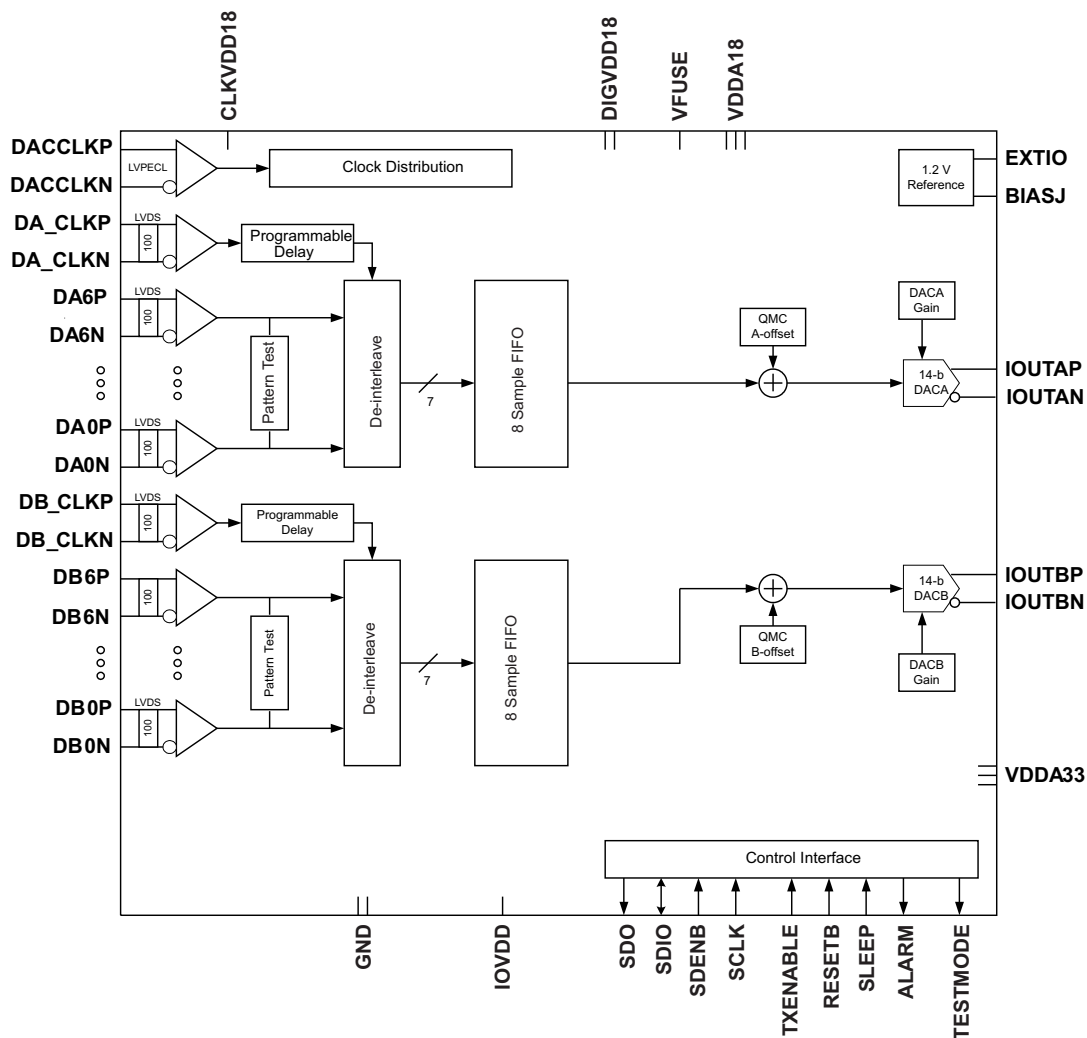


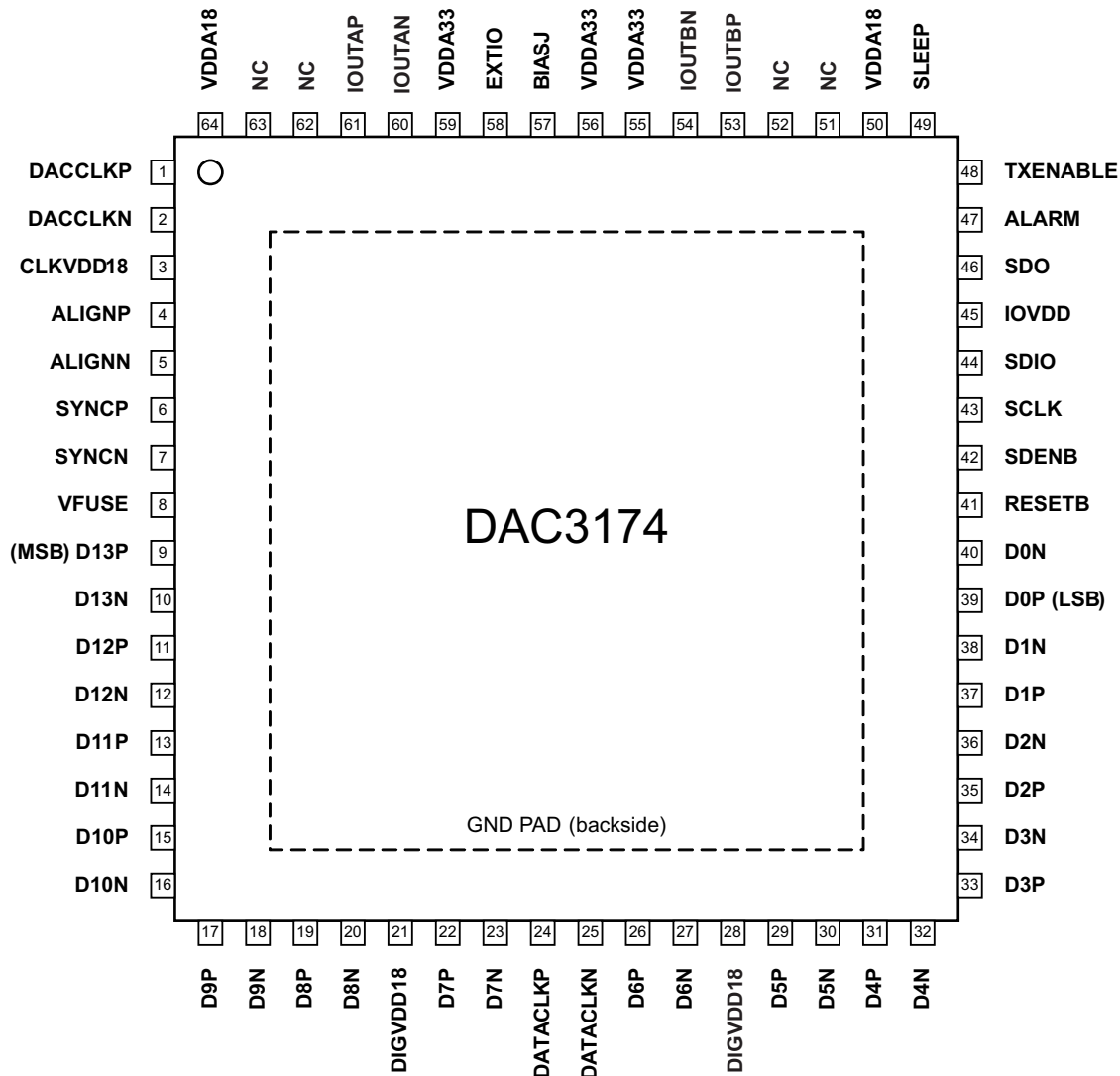
Figure 2. 7-Bit Interface Mode

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

PINOUT – SINGLE BUS MODE



PIN ASSIGNMENT TABLE – SINGLE BUS MODE

PIN		I/O	DESCRIPTION
NAME	NO.		
CONTROL/SERIAL			
SCLK	43	I	Serial interface clock. Internal pull-down.
SDENB	42	I	Serial interface clock. Internal pull-up.
SDIO	44	I/O	Bi-directional serial data in 3 pin mode (default). In 4-pin interface mode (register sif4_ena (config 0, bit 9)), the SDIO pin in an input only. Internal Pull-down.
SDO	46	O	Uni-directional serial interface data in 4 pin mode (register sif4_ena (config 0, bit 9)). The SDO pin is tri-stated in 3-pin interface mode (default). Internal Pulldown.
RESETB	41	I	Serial interface reset input. Active low. Initialized internal registers during high to low transition. Asynchronous. Internal pull-up.
ALARM	47	O	CMOS output for ALARM condition.
TXENABLE	48	I	Transmit enable active high input. TXENABLE must be high for the DATA to the DAC to be enabled. When TXENABLE is low, the digital logic section is forced to all 0, and any input data is ignored. Internal pull-down.
SLEEP	49	I	Puts device in sleep, active high. Internal pull-down.

PIN ASSIGNMENT TABLE – SINGLE BUS MODE (continued)

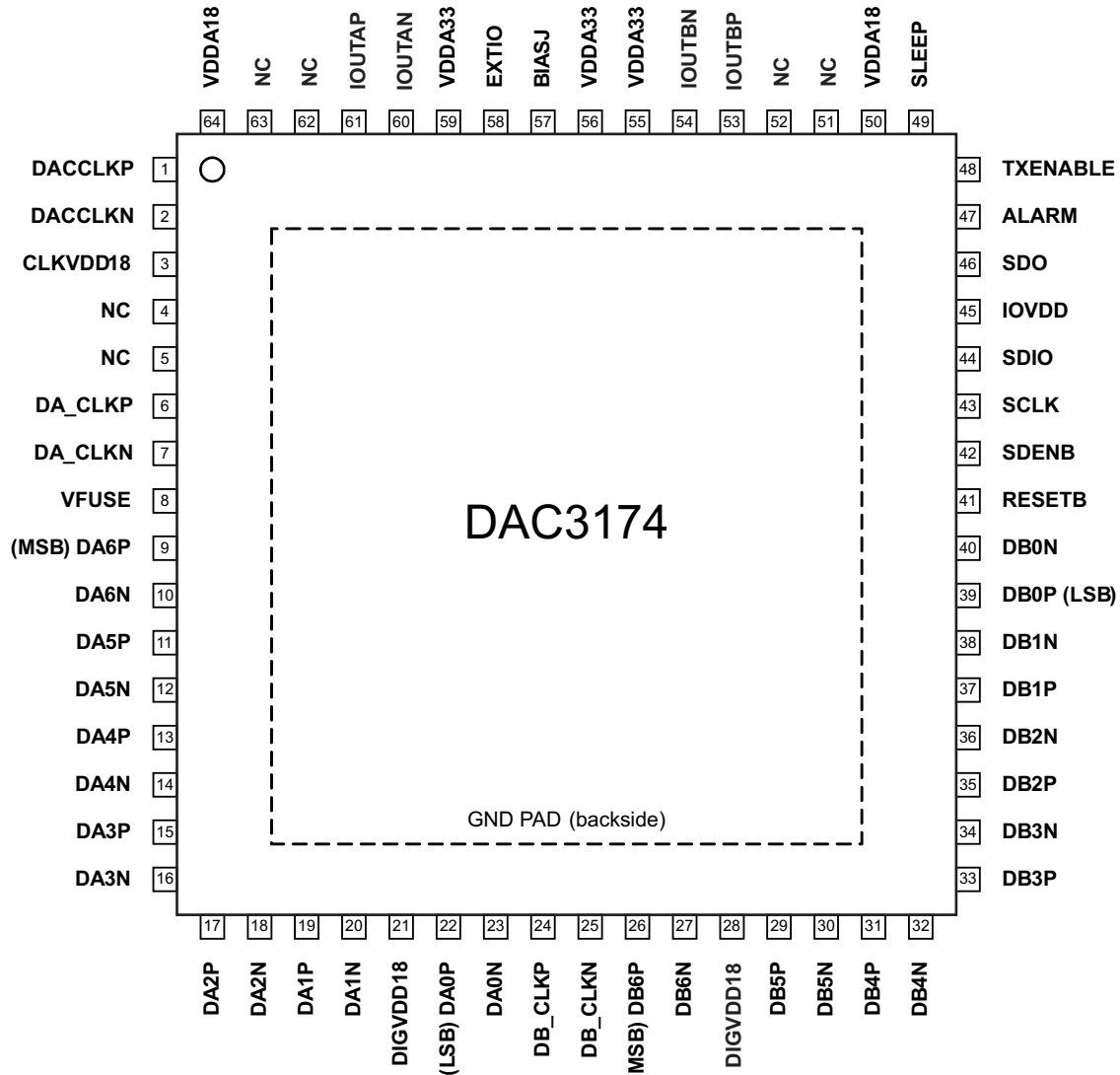
PIN		I/O	DESCRIPTION
NAME	NO.		
DATA INTERFACE			
DATA[13:0]P/N	9/10-19/20 22/23 26/27 29/30-39/40	I	LVDS input data bits for both channels. Each positive/negative LVDS pair has an internal 100 Ω termination resistor. Data format relative to DATACLKP/N clock is Double Data Rate (DDR) with two data transfers per DATAKP/N clock cycle. The data format is interleaved with channel A (rising edge) and channel B falling edge. In the default mode (reverse bus not enabled): DATA13P/N is most significant data bit (MSB) DATA0P/N is most significant data bit (LSB)
DATACLKP/N	24/25	I	DDR differential input data clock. Edge to center nominal timing. Ch A rising edge, Ch B falling edge in multiplexed output mode.
SYNCP/N	6/7	I	Reset the FIFO or to be used as a syncing source. These two functions are captured with the rising edge of DATACLKP/N. The signal captured by the falling edge of DATACLKP/N.
ALIGNP/N	4/5	I	LVPECL FIFO output synchronization. This positive/negative pair is captured with the rising edge of DACCLKP/N. It is used to reset the clock dividers and for multiple DAC synchronization. If unused it can be left unconnected.
OUTPUT/CLOCK			
DACCLKP/N	½	I	LVPECL clock input for DAC core with a self-bias of approximately CLKVDD18/2.
IOUTAP/N	61/60	O	A-Channel DAC current output. An offset binary data pattern of 0x0000 at the DAC input results in a full scale current source and the most positive voltage on the IOUTAP pin. Similarly, a 0xFFFF data input results in a 0 mA current source and the least positive voltage on the IOUTAP pin.
IOUTBP/N	53/54	O	B-Channel DAC current output. An offset binary data pattern of 0x0000 at the DAC input results in a full scale current source and the most positive voltage on the IOUTBP pin. Similarly, a 0xFFFF data input results in a 0 mA current source and the least positive voltage on the IOUTBP pin.
REFERENCE			
EXTIO	58	I/O	Used as external reference input when internal reference is disabled. Requires a 0.1 μF decoupling capacitor to GND when used as reference output.
BIASJ	57	O	Full-scale output current bias. For 20 mA full-scale output current, connect a 960 Ω resistor to GND.
POWER SUPPLY			
IOVDD	45	I	Supply voltage for CMOS IO's. 1.8V – 3.3V.
CLKVDD18	3	I	1.8V clock supply
DIGVDD18	21, 28	I	1.8V digital supply. Also supplies LVDS receivers.
VDDA18	50, 64	I	Analog 1.8V supply
VDDA33	55, 56, 59	I	Analog 3.3V supply
VFUSE	8	I	Digital supply voltage. (1.8V) This supply pin is also used for factory fuse programming. Connect to DVDD pins for normal operation.
NC	51, 52 62, 63		Not used. These pins can be left open or tied to GROUND in actual application use.

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

PIN OUT – DUAL BUS MODE



PIN ASSIGNMENT TABLE – DUAL BUS MODE

PIN		I/O	DESCRIPTION
NAME	NO.		
CONTROL/SERIAL			
SCLK	43	I	Serial interface clock. Internal pull-down.
SDENB	42	I	Serial interface clock. Internal pull-up.
SDIO	44	I/O	Bi-directional serial data in 3 pin mode (default). In 4-pin interface mode (register sif4_ena (config 0, bit 9)), the SDIO pin in an input only. Internal Pull-down.
SDO	46	O	Uni-directional serial interface data in 4 pin mode (register sif4_ena (config 0, bit 9)). The SDO pin is tri-stated in 3-pin interface mode (default). Internal Pulldown.
RESETB	41	I	Serial interface reset input. Active low. Initialized internal registers during high to low transition. Asynchronous. Internal pull-up.
ALARM	47	O	CMOS output for ALARM condition.
TXENABLE	48	I	Transmit enable active high input. TXENABLE must be high for the DATA to the DAC to be enabled. When TXENABLE is low, the digital logic section is forced to all 0, and any input data is ignored. Internal pull-down.
SLEEP	49	I	Puts device in sleep, active high. Internal pull-down.

PIN ASSIGNMENT TABLE – DUAL BUS MODE (continued)

PIN		I/O	DESCRIPTION
NAME	NO.		
DATA INTERFACE			
DA[6:0]P/N	9/10-19/20 22/23	I	LVDS positive input data bits for channel A. Each positive/negative LVDS pair has an internal 100 Ω termination resistor. Data format relative to DA_CLKP/N clock is Double Data Rate (DDR) with two data transfers per DA_CLKP/N clock cycle. The data format is 7 MSBs (rising edge)/7 LSBs falling edge. In the default mode (reverse bus not enabled): D6P/N is most significant data bit (MSB) D0P/N is most significant data bit (LSB)
DB[6:0]P/N	26/27 29/30-39/40	I	LVDS positive input data bits for channel B. Each positive/negative LVDS pair has an internal 100 Ω termination resistor. Data format relative to DB_CLKP/N clock is Double Data Rate (DDR) with two data transfers per DB_CLKP/N clock cycle. The data format is 7 MSBs (rising edge)/7 LSBs falling edge. In the default mode (reverse bus not enabled): D6P/N is most significant data bit (MSB) D0P/N is most significant data bit (LSB)
DA_CLKP/N	6/7	I	DDR differential input data clock for channel A. Edge to center nominal timing.
DB_CLKP/N	24/25	I	DDR differential input data clock for channel B. Edge to center nominal timing.
OUTPUT/CLOCK			
DACCLKP/N	½	I	LVPECL clock input for DAC core with a self-bias of approximately CLKVDD18/2.
IOUTAP/N	61/60	O	A-Channel DAC current output. An offset binary data pattern of 0x0000 at the DAC input results in a full scale current source and the most positive voltage on the IOUTAP pin. Similarly, a 0xFFFF data input results in a 0 mA current source and the least positive voltage on the IOUTAP pin. The IOUTAN pin is the complement of IOUTAP.
IOUTBP/N	53/54	O	B-Channel DAC current output. An offset binary data pattern of 0x0000 at the DAC input results in a full scale current source and the most positive voltage on the IOUTBP pin. Similarly, a 0xFFFF data input results in a 0 mA current source and the least positive voltage on the IOUTBP pin. The IOUTBN pin is the complement of IOUTBP.
REFERENCE			
EXTIO	58	I/O	Used as external reference input when internal reference is disabled. Requires a 0.1 μF decoupling capacitor to GND when used as reference output.
BIASJ	57	O	Full-scale output current bias. For 20 mA full-scale output current, connect a 960 Ω resistor to GND.
POWER SUPPLY			
IOVDD	45	I	Supply voltage for CMOS IO's. 1.8V – 3.3V.
CLKVDD18	3	I	1.8V clock supply
DIGVDD18	21, 28	I	1.8V digital supply. Also supplies LVDS receivers.
VDDA18	50, 64	I	Analog 1.8V supply
VDDA33	55, 56, 59	I	Analog 3.3V supply
VFUSE	8	I	Digital supply voltage. (1.8V) This supply pin is also used for factory fuse programming. Connect to DVDD pins for normal operation.
NC	4, 5 51, 52 62, 63		Not used. In actual application, Pins 51, 52, 62 and 63 can be left open or tied to GROUND. It is recommended to tie Pins 4 and 5 to DIGVDD18 and GROUND , respectively.

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

PACKAGE/ORDERING INFORMATION⁽¹⁾

PRODUCT	PACKAGE-LEAD	PACKAGE DESIGNATOR	SPECIFIED TEMPERATURE RANGE	ECO PLAN	LEAD/BALL FINISH	ORDERING NUMBER	TRANSPORT MEDIA	QUANTITY
DAC3174	QFN-64	RGC	–40°C to 85°C	GREEN (RoHS and no Sb/Br)	NiPdAu	DAC3174IRGC25	Tape and Reel	25
						DAC3174IRGCT		250
						DAC3174IRGCR		2000

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)

		VALUE	UNIT
Supply voltage	VDDA33 to GND	–0.5 to 4	V
	VDDA18 to GND	–0.5 to 2.3	
	CLKVDD18 to GND	–0.5 to 2.3	
	IOVDD to GND	–0.5 to 4	
	DIGVDD18 to GND	–0.5 to 2.3	
Terminal voltage range	CLKVDD18 to DIGVDD18	–0.5 to 0.5	V
	VDDA18 to DIGVDD18	–0.5 to 0.5	
	DA[6..0]P, DA[6..0]N, DB[6..0]P, DB[6..0]N, D[13..0]P, D[13..0]N, DATACLKP, DATACLKN, DA_CLKP, DA_CLKPN, DB_CLKP, DB_CLKN, SYNCN, SYNCN to GND	–0.5 to DIGVDD18 + 0.5	
	DACCLKP, DACCLKN, ALIGNP, ALIGNN	–0.5 to CLKVDD18 + 0.5	
	TXENABLE, ALARM, SDO, SDIO, SCLK, SDENB, RESETB to GND	–0.5 to IOVDD + 0.5	
	IOUTAP, IOUTAN, IOUTBP, IOUTBN to GND	–0.7 to 1.4	
	EXTIO, BIASJ to GND	–0.5 to VDDA33 + 0.5	
Storage temperature range		–65 to 150	°C
ESD, Human Body Model		2	kV

THERMAL INFORMATION

THERMAL METRIC ⁽¹⁾		DAC3174	UNITS
		QFN (64 PIN)	
θ_{JA}	Junction-to-ambient thermal resistance	23.0	°C/W
θ_{JCTop}	Junction-to-case (top) thermal resistance	7.6	
θ_{JB}	Junction-to-board thermal resistance	2.8	
Ψ_{JT}	Junction-to-top characterization parameter	0.1	
Ψ_{JB}	Junction-to-board characterization parameter	2.8	
θ_{JCbott}	Junction-to-case (bottom) thermal resistance	0.2	

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

ELECTRICAL CHARACTERISTICS – DC SPECIFICATIONS

Typical values at $T_A = 25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = 85^\circ\text{C}$, DAC sample rate = 500MSPS, 50% clock duty cycle, $V_{\text{DDA33}}/\text{IOVDD} = 3.3\text{V}$, $V_{\text{DDA18}}/\text{CLKVDD18}/\text{DIGVDD18} = 1.8\text{V}$, $\text{IOUT}_{\text{FS}} = 20\text{mA}$ (unless otherwise noted).

PARAMETER		NOTES	MIN	TYP	MAX	UNITS
Resolution			14			Bits
DC ACCURACY						
DNL Differential nonlinearity		1 LSB = IOUT _{FS} /2 ¹⁴	±1			LSB
INL Integral nonlinearity			±2			
ANALOG OUTPUTS						
Coarse gain linearity			±0.4			LSB
Offset error		Mid code offset	0.01			%FSR
Gain error		With external reference	±2			%FSR
		With internal reference	±2			
Gain mismatch		With internal reference	-2		2	%FSR
Minimum full scale output current		Nominal full-scale current, IOUT _{FS} = 16xIBAIS current	2			mA
Maximum full scale output current			20			
Output compliance range		IOUTFS = 20 mA	-0.5		1	V
Output resistance			300			kΩ
Output capacitance			5			pF
REFERENCE OUTPUT						
V _{REF}	Reference output voltage		1.14	1.2	1.26	V
	Reference output current		100			nA
REFERENCE INPUT						
VEXTIO Input voltage range		External reference mode	0.1	1.2	1.25	V
Input resistance			1			MΩ
Small signal bandwidth			500			kHz
Input capacitance			100			pF
TEMPERATURE COEFFICIENTS						
Offset drift			±1			ppm of FSR/°C
Gain drift		With external reference	±15			
		With internal reference	±30			
Reference voltage drift			±8			ppm /°C
POWER SUPPLY						
DIGVDD18, VFUSE, VDDA18, CLKVDD18			1.71	1.8	1.89	V
VDDA33			3.15	3.3	3.45	V
IOVDD		Sets CMOS IO voltage levels. Nominal 1.8V, 2.5V or 3.3V	1.71		3.45	V

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

ELECTRICAL CHARACTERISTICS – DC SPECIFICATIONS (continued)

Typical values at $T_A = 25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = 85^\circ\text{C}$, DAC sample rate = 500MSPS, 50% clock duty cycle, $V_{\text{DDA33}}/\text{IOVDD} = 3.3\text{V}$, $V_{\text{DDA18}}/\text{CLKVDD18}/\text{DIGVDD18} = 1.8\text{V}$, $\text{IOUT}_{\text{FS}} = 20\text{mA}$ (unless otherwise noted).

PARAMETER		NOTES	MIN	TYP	MAX	UNITS
POWER CONSUMPTION						
I_{VDDA33}	3.3V Analog supply current	MODE 1 $f_{\text{DAC}} = 491.52\text{ MSPS}$, QMC on, IF = 20 MHz		52	59	mA
$\text{I}_{\text{CLKVDD18}}$	1.8V Clock supply current			49	57	mA
$\text{I}_{\text{DIGVDD18}}$	1.8V Digital supply current (DIGVDD18 and VFUSE)			115	130	mA
I_{IOVDD}	1.8V IO Supply current			0.002	0.015	mA
P_{dis}	Total power dissipation			464	530	mW
I_{VDDA33}	3.3V Analog supply current	MODE 2 $f_{\text{DAC}} = 320\text{ MSPS}$, QMC on, IF = 20 MHz		51		mA
$\text{I}_{\text{CLKVDD18}}$	1.8V Clock supply current			38		mA
$\text{I}_{\text{DIGVDD18}}$	1.8V Digital supply current (DIGVDD18 and VFUSE)			87		mA
I_{IOVDD}	1.8V IO Supply current			0.002		mA
P_{dis}	Total power dissipation			396		mW
I_{VDDA33}	3.3V Analog supply current	MODE 3 Sleep mode, $f_{\text{DAC}} = 491.52\text{ MSPS}$, DAC in sleep mode		2.6		mA
$\text{I}_{\text{CLKVDD18}}$	1.8V Clock supply current			43		mA
$\text{I}_{\text{DIGVDD18}}$	1.8V Digital supply current (DIGVDD18 and VFUSE)			110		mA
I_{IOVDD}	1.8V IO Supply current			0.003		mA
P_{dis}	Total power dissipation			284		mW
I_{VDDA33}	3.3V Analog supply current	MODE 4 Power-down mode, no clock, DAC in sleep mode		1.6	4	mA
$\text{I}_{\text{CLKVDD18}}$	1.8V Clock supply current			1.8	4	mA
$\text{I}_{\text{DIGVDD18}}$	1.8V Digital supply current (DIGVDD18 and VFUSE)			0.7	3	mA
I_{IOVDD}	1.8V IO Supply current			0.003	0.015	mA
P_{dis}	Total power dissipation			10	26	mW
PSRR	Power supply rejection ratio	DC tested	-0.4		0.4	%/FSR/V
T	Operating temperature		-40		85	$^\circ\text{C}$

ELECTRICAL CHARACTERISTICS – AC SPECIFICATIONS

Typical values at $T_A = 25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = 85^\circ\text{C}$, DAC sample rate = 500MSPS, 50% clock duty cycle, $V_{\text{DDA33}}/\text{IOVDD} = 3.3\text{V}$, $V_{\text{DDA18}}/\text{CLKVDD18}/\text{DIGVDD18} = 1.8\text{V}$, $\text{IOUT}_{\text{FS}} = 20\text{mA}$ (unless otherwise noted).

PARAMETER		NOTES	MIN	TYP	MAX	UNITS
ANALOG OUTPUT						
f_{DAC}	Maximum sample rate		500			MSPS
$t_{\text{s(DAC)}}$	Output settling time to 0.1%	Transition: Code 0x0000 to 0x3FFF		11		ns
t_{PD}	Output propagation delay	Does not include digital latency		2		ns
$t_{\text{r(IOUT)}}$	Output rise time 10% to 90%			200		ps
$t_{\text{f(IOUT)}}$	Output fall time 90% to 10%			200		ps
	Digital Latency	Length of delay from DAC pin inputs to DATA at output pins. In normal operation mode including the latency of FIFO.		26		DACCLK
AC PERFORMANCE						
SFDR	Spurious free dynamic range	$f_{\text{DAC}} = 500 \text{ MSPS}, f_{\text{out}} = 10.1 \text{ MHz}$		82		dBc
		$f_{\text{DAC}} = 500 \text{ MSPS}, f_{\text{out}} = 20.1 \text{ MHz}$		78		
		$f_{\text{DAC}} = 500 \text{ MSPS}, f_{\text{out}} = 70.1 \text{ MHz}$		74		
IMD3	Intermodulation distortion	$f_{\text{DAC}} = 500 \text{ MSPS}, f_{\text{out}} = 10.1 \pm 0.5 \text{ MHz}$		84		dBc
		$f_{\text{DAC}} = 500 \text{ MSPS}, f_{\text{out}} = 20.1 \pm 0.5 \text{ MHz}$		84		
		$f_{\text{DAC}} = 500 \text{ MSPS}, f_{\text{out}} = 70.1 \pm 0.5 \text{ MHz}$		75		
		$f_{\text{DAC}} = 500 \text{ MSPS}, f_{\text{out}} = 150.1 \pm 0.5 \text{ MHz}$		63		
NSD	Noise spectral density	$f_{\text{DAC}} = 500 \text{ MSPS}, f_{\text{out}} = 10.1 \text{ MHz}$		160		dBc/Hz
		$f_{\text{DAC}} = 500 \text{ MSPS}, f_{\text{out}} = 20.1 \text{ MHz}$		157		
		$f_{\text{DAC}} = 500 \text{ MSPS}, f_{\text{out}} = 70.1 \text{ MHz}$		155		
ACLR	Adjacent channel leakage ratio	$f_{\text{DAC}} = 491.52 \text{ MSPS}, f_{\text{out}} = 30.72 \text{ MHz}, \text{WCDMA TM1}$		78		dBc
		$f_{\text{AC}} = 491.52 \text{ MSPS}, f_{\text{out}} = 153.6 \text{ MHz WCDMA TM1}$		74		

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

ELECTRICAL CHARACTERISTICS – DIGITAL SPECIFICATIONS

Typical values at $T_A = 25^{\circ}\text{C}$, full temperature range is $T_{\text{MIN}} = -40^{\circ}\text{C}$ to $T_{\text{MAX}} = 85^{\circ}\text{C}$, DAC sample rate = 500MSPS, 50% clock duty cycle, VDDA33/IOVDD = 3.3V, VDDA18/CLKVDD18/DIGVDD18 = 1.8V, $\text{IOUT}_{\text{FS}} = 20\text{mA}$ (unless otherwise noted).

PARAMETERS		NOTES	MIN	TYP	MAX	UNITS
CMOS DIGITAL INPUTS (RESETB, SDENB, SCLK, SDIO, TXENABLE)						
V _{IH}	High-level input voltage	IOVDD = 3.3 V, 2.5 V or 1.8 V	IOVDDx 0.6			V
V _{IL}	Low-level input voltage		0.25xIO VDD			V
I _{IH}	High-level input current		-40		40	μA
I _{IL}	Low-level input current		-40		40	μA
DIGITAL OUTPUTS – CMOS INTERFACE (SDOUT, SDIO)						
V _{OH}	High-level output voltage	IOVDD = 3.3 V, 2.5 V, 1.8 V	0.85xIOV DD			V
V _{OL}	Low-level output voltage		0.125xI OVDD			V
SERIAL PORT TIMING						
t _s (SENDB)	Setup time, SDENB to rising edge of SCLK		20			ns
t _s (SDIO)	Setup time, SDIO to rising edge of SCLK		10			ns
t _h (SDIO)	Hold time, SDIO from rising edge of SCLK		5			ns
t _i (SCLK)	Period of SCLK		100			ns
t _i (SCLKH)	High time of SCLK		40			ns
t _i (SCLKL)	Low time of SCLK		40			ns
t _d (DATA)	Data output delay after falling edge of SCLK		10			ns
T _{RESET}	Minimum RESTB pulse width		25			ns
LVDS INTERFACE (D[x..0]P/N, DA[x..0]P/N, DB[x..0]P/N, DA_CLKP/N, DB_CLKP/N, DATACLKP/N, SYNCN/P, ALIGNN/P)						
V _{A,B+}	Logic high differential input voltage threshold		175			mV
V _{A,B-}	Logic low differential input voltage threshold				-175	mV
V _{COM}	Input Common Mode Range		1.0	1.2	2.0	V
Z _T	Internal termination		85	110	135	Ω
C _L	LVDS input capacitance		2			pF

ELECTRICAL CHARACTERISTICS – DIGITAL SPECIFICATIONS (continued)

Typical values at $T_A = 25^{\circ}\text{C}$, full temperature range is $T_{\text{MIN}} = -40^{\circ}\text{C}$ to $T_{\text{MAX}} = 85^{\circ}\text{C}$, DAC sample rate = 500MSPS, 50% clock duty cycle, VDDA33/IOVDD = 3.3V, VDDA18/CLKVDD18/DIGVDD18 = 1.8V, $\text{IOUT}_{\text{FS}} = 20\text{mA}$ (unless otherwise noted).

PARAMETERS		NOTES		MIN	TYP	MAX	UNITS
LVDS INPUT TIMING							
t _s (DATA)	Setup time	D[x..0] valid to DATACLK rising or falling edge for single bus single clock mode ; DA/DB[x...0] valid to DB_CLK rising or falling edge for dual bus single clock mode; DA[x..0] valid to DA_CLK rising or falling edge, and DB[x...0] valid for DB_CLK rising or falling edge for dual bus dual clock mode	config3 Setting				ps
			datadly	clkdly			
			0	0	-20		
			0	1	-120		
			0	2	-220		
			0	3	-310		
			0	4	-390		
			0	5	-480		
			0	6	-560		
			0	7	-630		
			1	0	70		
			2	0	150		
			3	0	230		
			4	0	330		
			5	0	430		
			6	0	530		
			7	0	620		
t _h (DATA)	Hold time	D[x..0] valid to DATACLK rising or falling edge for single bus single clock mode; DA/DB[x...0] valid to DB_CLK rising or falling edge for dual bus single clock mode; DA[x..0] valid to DA_CLK rising or falling edge, and DB[x...0] valid for DB_CLK rising or falling edge for dual bus dual clock mode.	congfig3 Setting				ps
			datadly	clkdly			
			0	0	310		
			0	1	390		
			0	2	480		
			0	3	560		
			0	4	650		
			0	5	740		
			0	6	850		
			0	7	930		
			1	0	200		
			2	0	100		
			3	0	20		
			4	0	-60		
			5	0	-140		
			6	0	-220		
			7	0	-290		

TYPICAL CHARACTERISTICS

All plots are at 25°C, nominal supply voltages, $f_{DAC} = 500\text{MSPS}$, 50% clock duty cycle, 0-dBFS input signal and 20mA full-scale output current (unless otherwise noted).

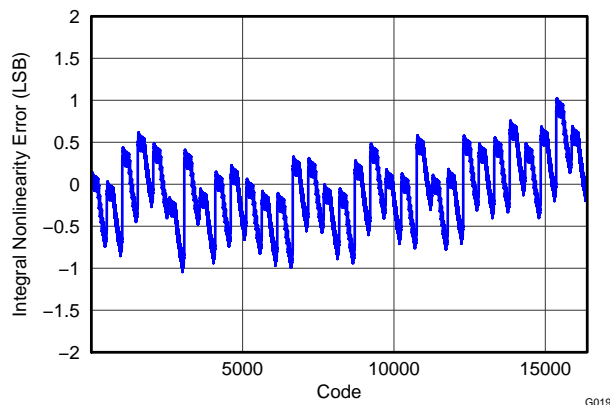


Figure 3. Integral Nonlinearity

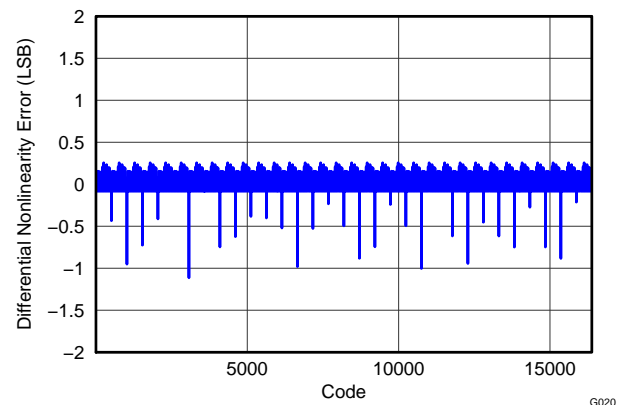


Figure 4. Differential Nonlinearity

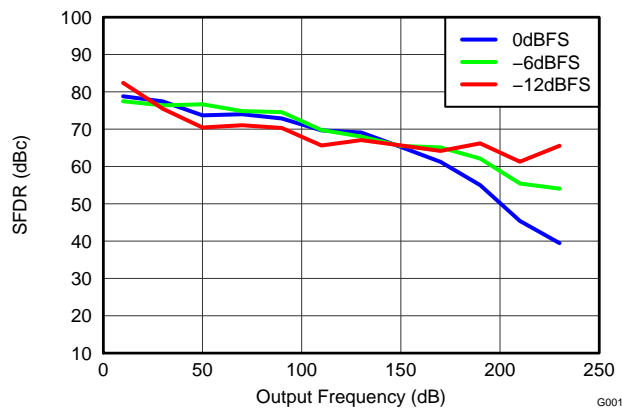


Figure 5. SFDR vs Output Frequency Over Input Scale

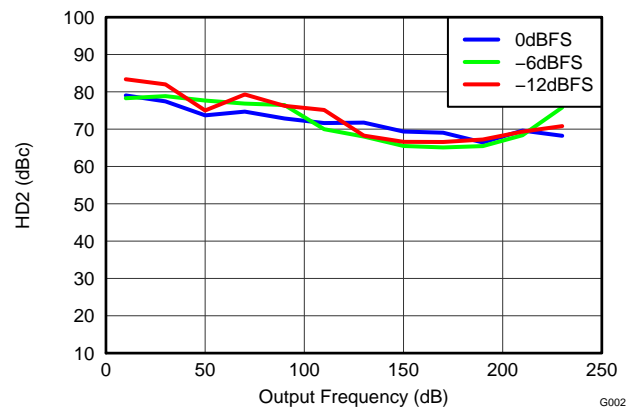


Figure 6. Second-Order Harmonic Distortion vs Output Frequency Over Input Scale

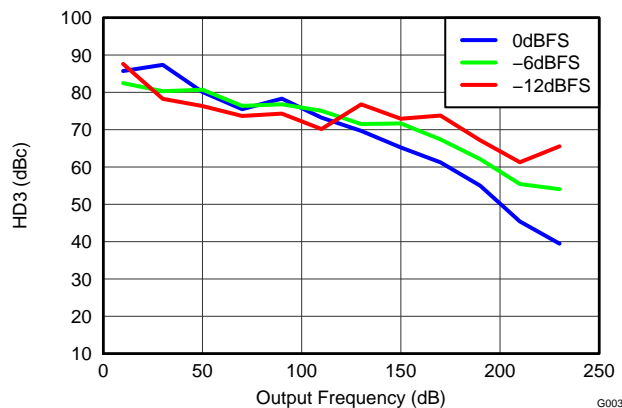
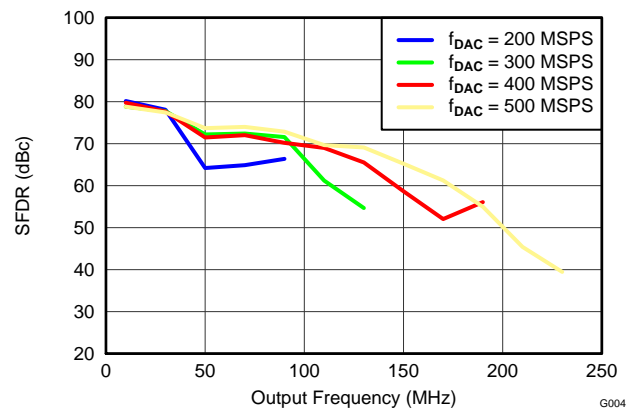


Figure 7. Third-Order Harmonic Distortion vs Output Frequency Over Input Scale


Figure 8. SFDR vs Output Frequency Over f_{DAC}

TYPICAL CHARACTERISTICS (continued)

All plots are at 25°C, nominal supply voltages, $f_{DAC} = 500$ MSPS, 50% clock duty cycle, 0-dBFS input signal and 20mA full-scale output current (unless otherwise noted).

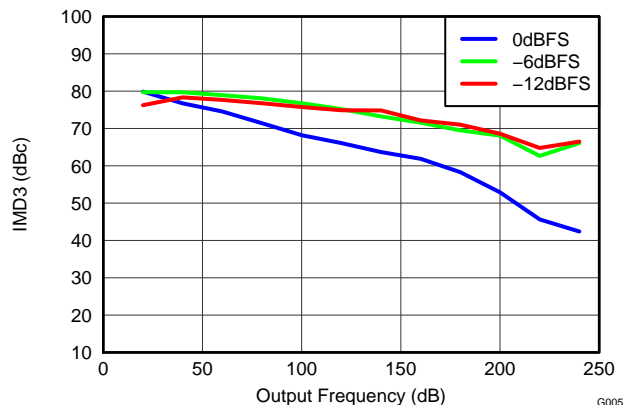


Figure 9. IMD3 vs Output Frequency Over Input Scale

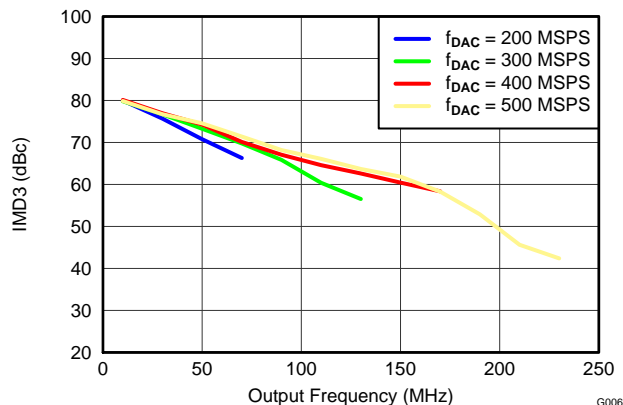


Figure 10. IMD3 vs Output Frequency Over f_{DAC}

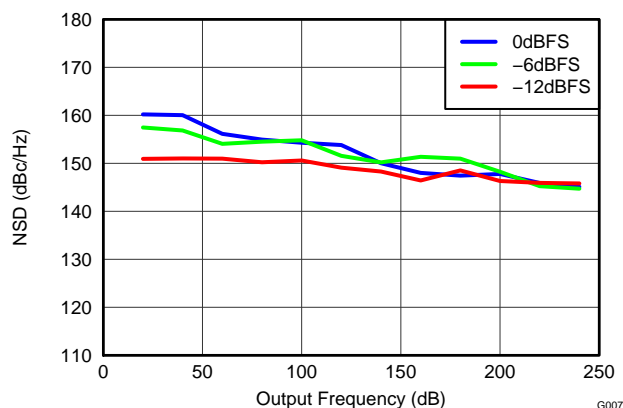


Figure 11. NSD vs Output Frequency Over Input Scale

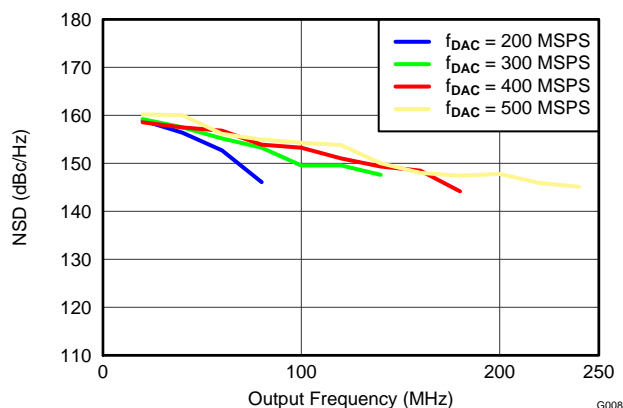


Figure 12. NSD vs Output Frequency Over f_{DAC}

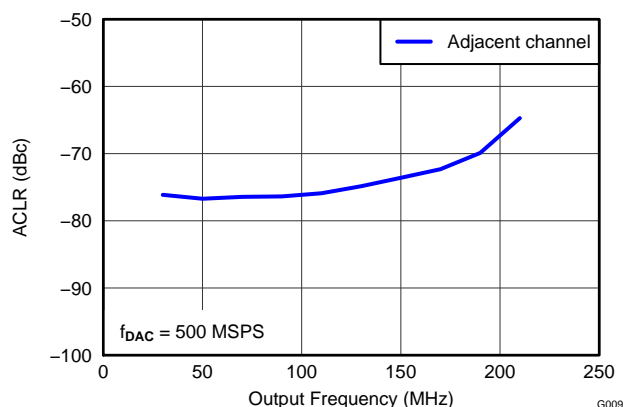


Figure 13. ACLR (Adjacent Channel) vs Output Frequency

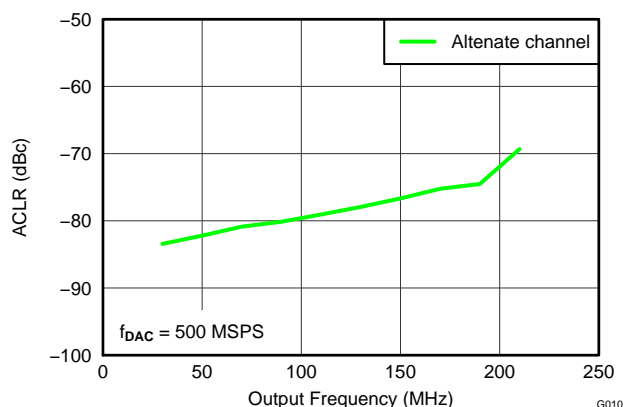


Figure 14. ACLR (Alternate Channel) vs Output Frequency

TYPICAL CHARACTERISTICS (continued)

All plots are at 25°C, nominal supply voltages, $f_{DAC} = 500\text{MSPS}$, 50% clock duty cycle, 0-dBFS input signal and 20mA full-scale output current (unless otherwise noted).

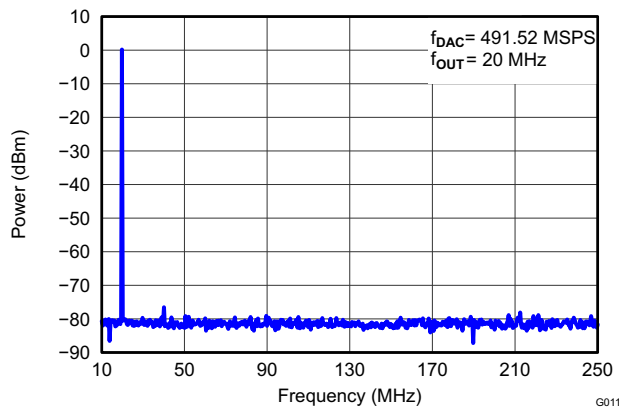


Figure 15. Single-Tone Spectral Plot (IF = 20MHz)

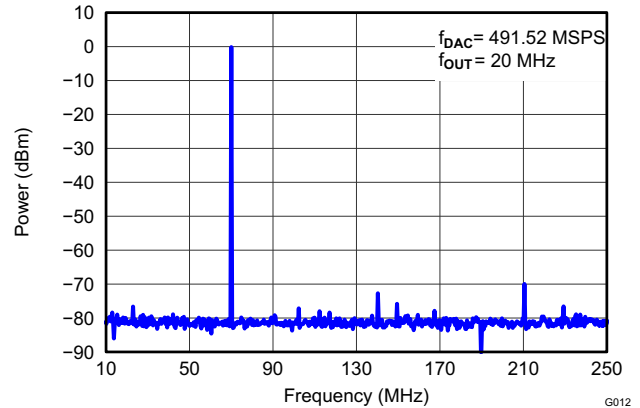


Figure 16. Single-Tone Spectral Plot (IF = 70MHz)

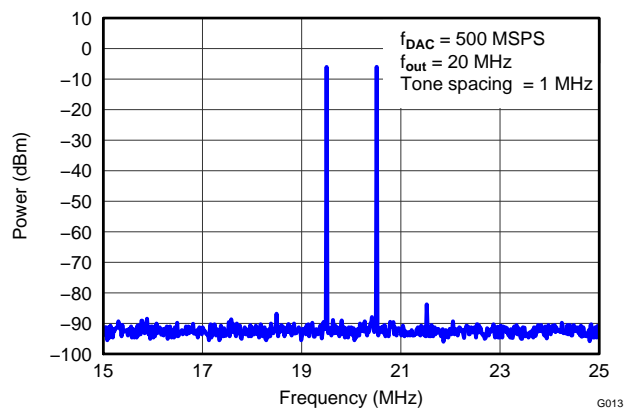


Figure 17. Two-tone Spectral Plot (IF = 20MHz)

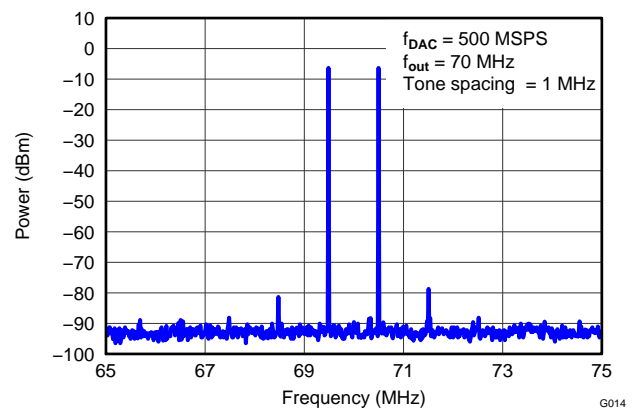


Figure 18. Two-tone Spectral Plot (IF = 70MHz)

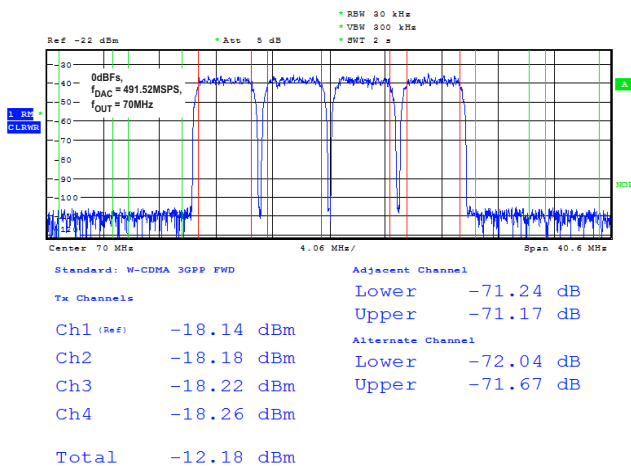


Figure 19. Four-carrier WCDMA Test Mode 1

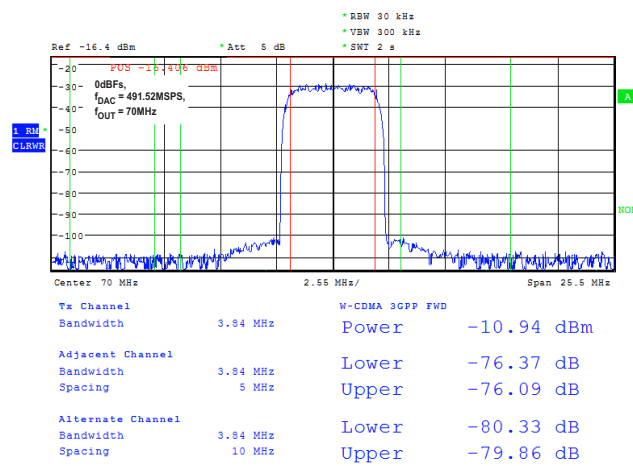


Figure 20. Single-carrier WCDMA Test Mode 1

TYPICAL CHARACTERISTICS (continued)

All plots are at 25°C, nominal supply voltages, $f_{DAC} = 500\text{MSPS}$, 50% clock duty cycle, 0-dBFS input signal and 20mA full-scale output current (unless otherwise noted).

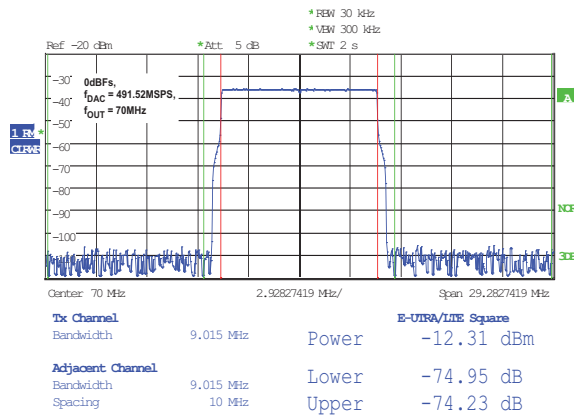


Figure 21. ACPR - LTE 10MHz FDD E-TM 1.1

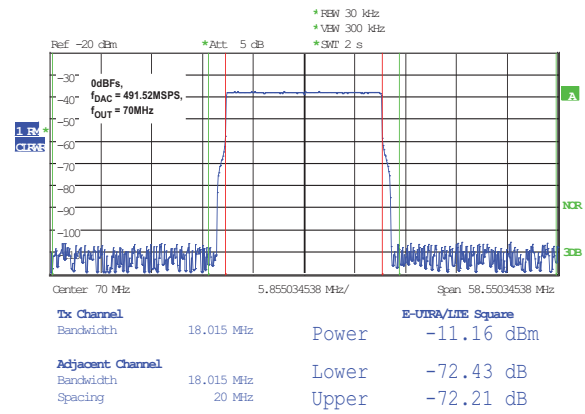


Figure 22. ACPR - LTE 20MHz FDD E-TM 1.1

DEFINITION OF SPECIFICATIONS

Adjacent Carrier Leakage Ratio (ACLR): Defined as the ratio in decibels relative to the carrier (dBc) between the measured power within a channel and that of its adjacent channel.

Analog and Digital Power Supply Rejection Ratio (APSSR, DPSSR): Defined as the percentage error in the ratio of the delta IOUT and delta supply voltage normalized with respect to the ideal IOUT current.

Differential Nonlinearity (DNL): Defined as the variation in analog output associated with an ideal 1 LSB change in the digital input code.

Gain Drift: Defined as the maximum change in gain, in terms of ppm of full-scale range (FSR) per °C, from the value at ambient (25°C) to values over the full operating temperature range.

Gain Error: Defined as the percentage error (in FSR%) for the ratio between the measured full-scale output current and the ideal full-scale output current.

Integral Nonlinearity (INL): Defined as the maximum deviation of the actual analog output from the ideal output, determined by a straight line drawn from zero scale to full scale.

Intermodulation Distortion (IMD3): The two-tone IMD3 is defined as the ratio (in dBc) of the 3rd-order intermodulation distortion product to either fundamental output tone.

Offset Drift: Defined as the maximum change in DC offset, in terms of ppm of full-scale range (FSR) per °C, from the value at ambient (25°C) to values over the full operating temperature range.

Offset Error: Defined as the percentage error (in FSR%) for the ratio between the measured mid-scale output current and the ideal mid-scale output current.

Output Compliance Range: Defined as the minimum and maximum allowable voltage at the output of the current-output DAC. Exceeding this limit may result reduced reliability of the device or adversely affecting distortion performance.

Reference Voltage Drift: Defined as the maximum change of the reference voltage in ppm per degree Celsius from value at ambient (25°C) to values over the full operating temperature range.

Spurious Free Dynamic Range (SFDR): Defined as the difference (in dBc) between the peak amplitude of the output signal and the peak spurious signal.

Signal to Noise Ratio (SNR): Defined as the ratio of the RMS value of the fundamental output signal to the RMS sum of all other spectral components below the Nyquist frequency, including noise, but excluding the first six harmonics and dc.

TIMING DIAGRAMS

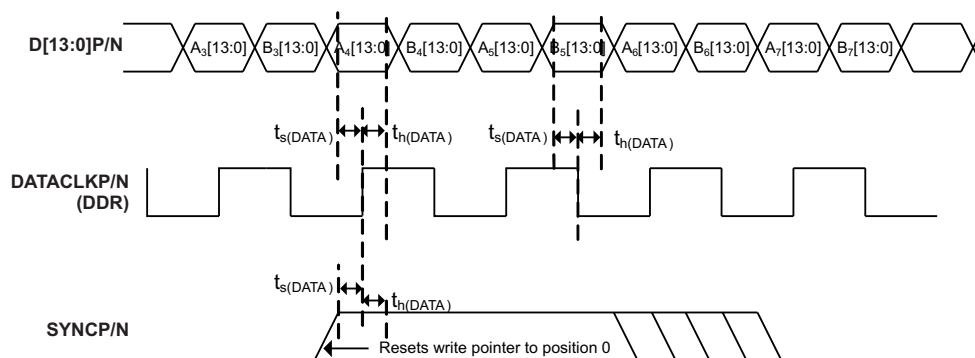


Figure 23. Input Data Timing for Single Bus Single Clock Mode

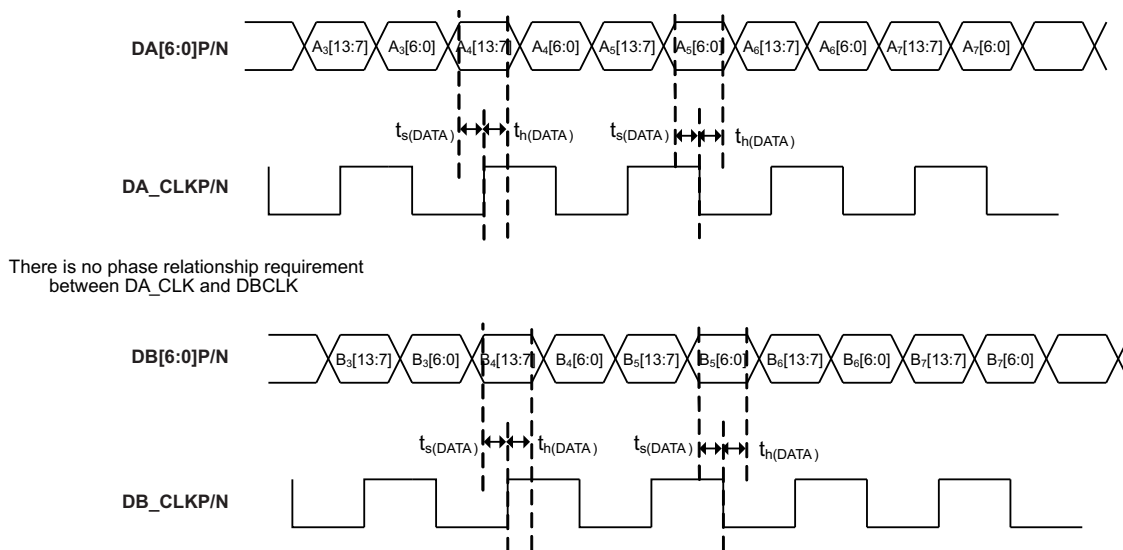


Figure 24. Input Data Timing Diagram for Dual Bus Dual Clock Mode

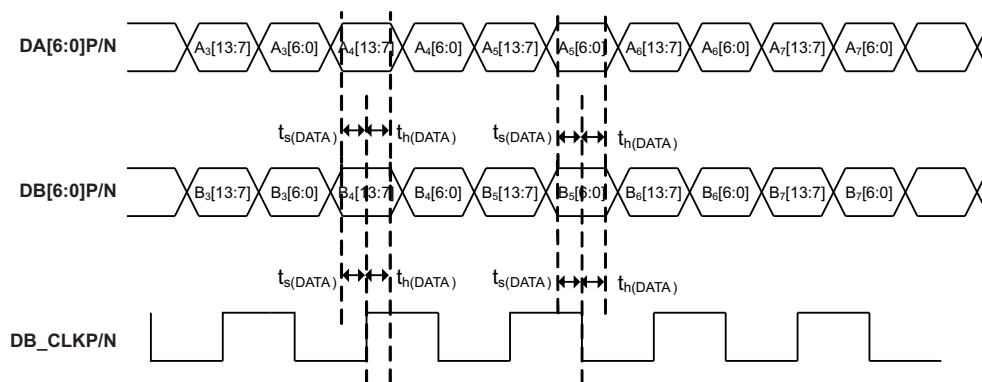


Figure 25. Input Data Timing Diagram Dual Bus Single Clock Mode

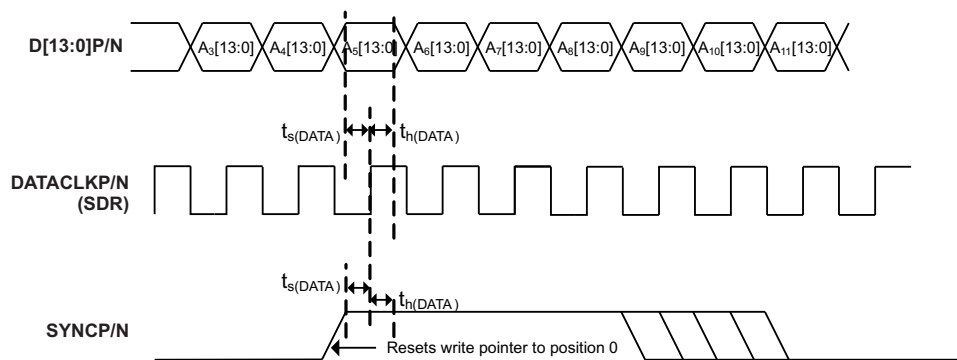


Figure 26. Input Data Timing Diagram Single Channel SDR Mode

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

DATA INPUT FORMATS

Table 1. Mode: Single Bus Single Clock Mode

DIFFERENTIAL PAIR (P/N)	BITS	
	DATACLK RISING EDGE	DATACLK FALLING EDGE
D13	A13	B13
D12	A12	B12
D11	A11	B11
D10	A10	B10
D9	A9	B9
D8	A8	B8
D7	A7	B7
D6	A6	B6
D5	A5	B5
D4	A4	B4
D3	A3	B3
D2	A2	B2
D1	A1	B1
D0	A0	B0
SYNC	FIFO Write Reset	–

Table 2. Mode: Single Channel SDR Mode

DIFFERENTIAL PAIR (P/N)	BITS	
	DATACLK RISING EDGE	DATACLK FALLING EDGE
D13	A13	–
D12	A12	–
D11	A11	–
D10	A10	–
D9	A9	–
D8	A8	–
D7	A7	–
D6	A6	–
D5	A5	–
D4	A4	–
D3	A3	–
D2	A2	–
D1	A1	–
D0	A0	–
SYNC	FIFO Write Reset	–

Table 3. Mode: Dual Bus Single Clock Mode

DIFFERENTIAL PAIR (P/N)	DB_CLK RISING EDGE	DB_CLK FALLING EDGE
DA6	A13	A6
DA5	A12	A5
DA4	A11	A4
DA3	A10	A3
DA2	A9	A2
DA1	A8	A1
DA0	A7	A0
DB6	B13	B6
DB5	B12	B5
DB4	B11	B4
DB3	B10	B3
DB2	B9	B2
DB1	B8	B1
DB0	B7	B0
SYNC	FIFO Write Reset	-

Table 4. Mode: Dual Bus Dual Clock Mode

DIFFERENTIAL PAIR (P/N)	DA_CLK RISING EDGE	DA_CLK FALLING EDGE
DA6	A13	A6
DA5	A12	A5
DA4	A11	A4
DA3	A10	A3
DA2	A9	A2
DA1	A8	A1
DA0	A7	A0
	DB_CLK RISING EDGE	DB_CLK FALLING EDGE
DB6	B13	B6
DB5	B12	B5
DB4	B11	B4
DB3	B10	B3
DB2	B9	B2
DB1	B8	B1
DB0	B7	B0

NOTE

When the register rev (config0, bit 11) is asserted, the MSB/LSB of the input bits are reversed. When using the 14-bit interface, all 14 bits are reversed as one word; when using the 7-bit interface, each 7-bit is reversed.

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

SERIAL INTERFACE DESCRIPTION

The serial port of the DAC3174 is a flexible serial interface which communicates with industry standard microprocessors and microcontrollers. The interface provides read/write access to all registers used to define the operating modes of DAC3174. It is compatible with most synchronous transfer formats and can be configured as a 3 or 4 pin interface by *sif4_ena* in register config 0, bit 9. In both configurations, SCLK is the serial interface input clock and SDENB is serial interface enable. For 3 pin configuration, SDIO is a bidirectional pin for both data in and data out. For 4 pin configuration, SDIO is data in only and SDO is data out only. Data is input into the device with the rising edge of SCLK. Data is output from the device on the falling edge of SCLK.

Each read/write operation is framed by signal SDENB (Serial Data Enable Bar) asserted low. The first frame byte is the instruction cycle which identifies the following data transfer cycle as read or write as well as the 7-bit address to be accessed. [Table 5](#) indicates the function of each bit in the instruction cycle and is followed by a detailed description of each bit. The data transfer cycle consists of two bytes.

Table 5. Instruction byte of the Serial interface

	MSB							LSB
Bit	7	6	5	4	3	2	1	0
Description	R/W	A6	A5	A4	A3	A2	A1	A0

R/W Identifies the following data transfer cycle as a read or write operation. A high indicates a read operation from DAC3174 and a low indicates a write operation to DAC3174.

[A6 : A0] Identifies the address of the register to be accessed during the read or write operation.

[Figure 27](#) shows the serial interface timing diagram for a DAC3174 write operation. SCLK is the serial interface clock input to DAC3174. Serial data enable SDENB is an active low input to DAC3174. SDIO is serial data in. Input data to DAC3174 is clocked on the rising edges of SCLK.

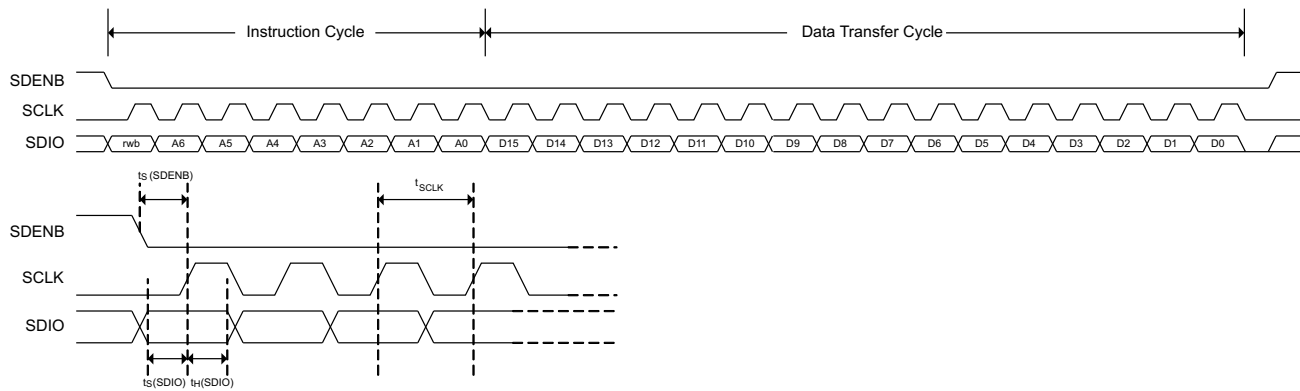


Figure 27. Serial Interface Write Timing Diagram

[Figure 28](#) shows the serial interface timing diagram for a DAC3174 read operation. SCLK is the serial interface clock input to DAC3174. Serial data enable SDENB is an active low input to DAC3174. SDIO is serial data in during the instruction cycle. In 3 pin configuration, SDIO is data out from the DAC3174 during the data transfer cycle, while SDO is in a high-impedance state. In 4 pin configuration, both SDIO and SDO are data out from the DAC3174 during the data transfer cycle. At the end of the data transfer, SDIO and SDO will output low on the final falling edge of SCLK until the rising edge of SDENB when they will 3-state.

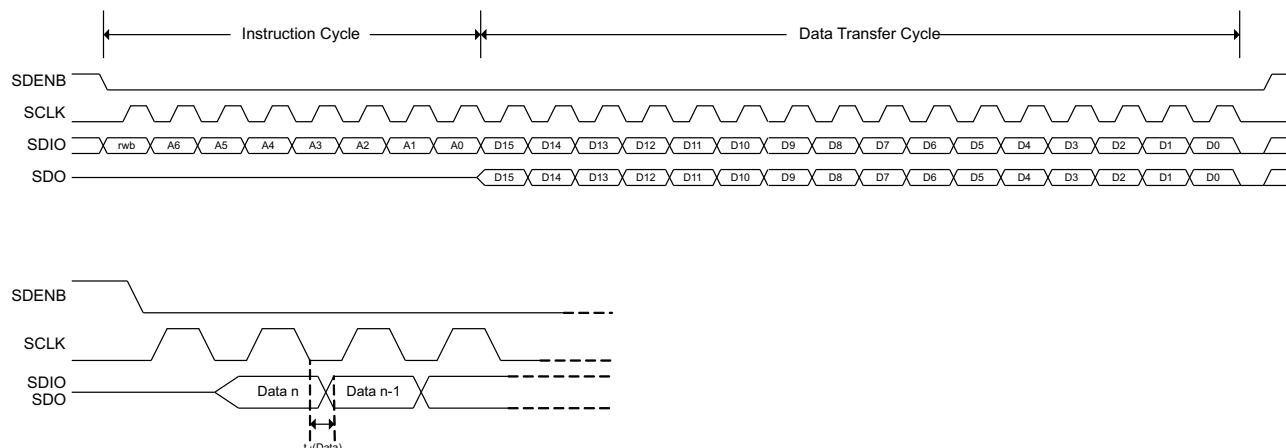


Figure 28. Serial Interface Read Timing Diagram

REGISTER DESCRIPTIONS

In the SIF interface there are four types of registers:

NORMAL: The NORMAL register type allows data to be written and read from. All 16-bits of the data are registered at the same time. There is no synchronizing with an internal clock thus all register writes are asynchronous with respect to internal clocks. There are three subtypes of NORMAL:

AUTOSYNC: A NORMAL register that causes a sync to be generated after the write is finished. These are most commonly used in things like offsets and phaseadd where there is a word or block setup that extends across multiple registers and all of the registers need to be programmed before any take effect on the circuit. For example, the phaseadd is two registers long. It wouldn't serve the user to have the first write 16 of the 32 bits cause a change in the frequency, so the design allows all the registers to be written and then when that last one for this block is finished, an autosync is generated for the mixer telling it to grab all the new SIF values. This will occur on a mixer clock cycle so that no meta-stability errors occur.

No RESET Value: These are NORMAL registers, but for one reason or another reset value can not be guaranteed. This could be because the register has some read_only bits or some internal logic partially controls the bit values. An example is the SIF_CONFIG6 register. The bits come from the temperature sensor and the fuses. Depending on which fuses are blown and what the die temp is the reset value will be different.

FUSE controlled: While this isn't a type of register, you may see this description in the area describing the default value for the register. What it means is that fuses will change the default value and the value shown in the document is for when no fuses are blown.

READ_ONLY: Registers that are internal wires ANDed with the address bus then connected to the SIF output data bus.

WRITE_TO_CLEAR: These registers are just like NORMAL registers with one exception. They can be written and read, however, when the internal logic asynchronously sets a bit high in one of these registers, that bit stays high until it is written to '0'. This way interrupts will be captured and stay constant until cleared by the user.

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com
Table 6. Register Map

Name	Address	Default	(MSB) Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	(LSB) Bit 0	
config0	0x00	0x44FC	qmc_offset_ena	dual_ena	chipwidth (1:0)		rev	twos	sif4_ena	reserved	fifo_ena	alarm_out_ena	alarm_out_pol	alignrx_ena	syncrx_ena	lvdsdataclk_ena	reserved	synconly_ena	
config1	0x01	0x600E	iotest_ena	bsideclk_ena	fullword_interface_ena	64cnt_ena	daccclk_gone_ena	dataclkgone_ena	collision_ena	reserved	daca_compliment	dacb_compliment	sif_sync	sif_sync_ena	alarm_2away_ena	alarm_1away_ena	alarm_collision_ena	reserved	
config2	0x02	0x3FFF	reserved		lvdsdata_ena (13:0)														
config3	0x03	0x0000	datadlya (2:0)			clkdiya (2:0)			datadlyb (2:0)			clkdiyb (2:0)			extref_ena	reserved		dual_ena	
config4	0x04	0x0000	reserved		iotest_results (13:0)														
config5	0x05	0x0000	alarm_from_zerochka	alarm_from_zerochkb	alarms_from_fifo (2:0)			alarms_from_fifob (2:0)			alarm_daccclk_gone	alarm_dataclk_gone	clock_gone	alarm_from_iotesta	alarm_from_iotestb	reserved			
config6	0x06	0x0000	tempdata (7:0)								fuse_cntl (5:0)							reserved	
config7	0x07	0xFFFF	alarms_mask (15:0)																
config8	0x08	0x4000	reserved			qmc_offseta (12:0)													
config9	0x09	0x8000	fifo_offset (2:0)			qmc_offsetb (12:0)													
config10	0x0A	0xF080	coarse_dac (3:0)				fuse_sleep	reserved	reserved	tsense_sleep	clkrecv_ena	sleepa	sleepb	reserved					
config11	0x0B	0x1111	reserved				reserved				reserved				reserved				
config12	0x0C	0x3A7A	reserved		iotest_pattern0 (13:0)														
config13	0x0D	0x36B6	reserved		iotest_pattern1 (13:0)														
config14	0x0E	0x2AEA	reserved		iotest_pattern2 (13:0)														
config15	0x0F	0x0545	reserved		iotest_pattern3 (13:0)														
config16	0x10	0x1A1A	reserved		iotest_pattern4 (13:0)														
config17	0x11	0x1616	reserved		iotest_pattern5 (13:0)														
config18	0x12	0x2AAA	reserved		iotest_pattern6 (13:0)														
config19	0x13	0x06C6	reserved		iotest_pattern7 (13:0)														
config20	0x14	0x0000	sifdac_ena	reserved	sifdac (13:0)														
config21	0x15	0xFFFF	sleepcntl (15:0)																
config22	0x16	0x0000	fa002_data(15:0)																
config23	0x17	0x0000	fa002_data(31:16)																
config24	0x18	0x0000	fa002_data(47:32)																
config25	0x19	0x0000	fa002_data(63:48)																
config127	0x7F	0x0044	reserved		reserved		reserved		reserved		reserved	titest_voh	titest_vol	vendorid (1:0)		versionid (2:0)			

Register name: config0 – Address: 0x00, Default: 0x44FC

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config0	0x00	15	qmc_offset_ena	Enable the offset function when asserted.	0
		14	dual_ena	Utilizes both DACs when asserted.	1 FUSE controlled
		13:12	chipwidth	Programmable bits for setting the input interface width. 00: all 14 bits are used. 01: upper 12 bits are used 10: upper 10 bits are used 11: upper 10 bits are used	00
		11	rev	Reverses the input bits. When using the 7bit interface, this reverse each 7-bit input, however when using the 14-bit interface, all 14-bits are reversed as one word.	0
		10	twos	When asserted, this bit tells the chip to presume 2's complement data is arriving at the input. Otherwise offset binary is presumed.	1
		9	sif4_ena	When asserted the SIF interface becomes a 4 pin interface. This bit has a lower priority than the dieid_ena bit.	0
		8	reserved	reserved	0
		7	fifo_ena	When asserted, the FIFO is absorbing the difference between INPUT clock and DAC clock. If it is not asserted then the FIFO buffering is bypassed but the reversing of bits and handling of offset binary input is still available. NOTE: When the FIFO is bypassed the DACCCLK and DATACLK must be aligned or there may be timing errors; and, it is not recommended for actual application use.	1
		6	alarm_out_ena	When asserted the pin alarm becomes an output instead of a tri-stated pin.	1
		5	alarm_out_pol	This bit changes the polarity of the ALARM signal. (0=negative logic, 1=positive logic)	1
		4	alignrx_ena	When asserted the ALIGN pin receiver is powered up. NOTE: It is recommended to clear this bit when ALIGNP/N are not used (dual bus mode, and SYNC ONLY and SIF_SYNC modes in single bus mode).	1
		3	syncrx_ena	When asserted the SYNC pin receiver is powered up NOTE: It is recommended to clear this bit when SYNC P/N are not used (dual bus mode, and SIF_SYNC mode in single bus mode.)	1
		2	lvdsdataclk_ena	When asserted the DATACLK pin receiver is powered up.	1
		1	reserved	reserved	0
		0	synconly_ena	When asserted the chip is put into the SYNC ONLY mode where the SYNC ONLY pin is used as the sync input for both the front and back of the FIFO.	0

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

Register name: config1 – Address: 0x01, Default: 0x600E

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config1	0x01	15	iotest_ena	Turns on the io-testing circuitry when asserted. This is the circuitry that will compare a 8 sample input pattern to SIF programmed registers to make sure the data coming into the chip meets setup/hold requirements. If this bit is a '0' then the clock to this circuitry is turned off for power savings. NOTE: Sample 0 should be aligned with the rising edge of SYNC.	0
		14	bsideclk_ena	When asserted the input clock for the B side datapath is enabled. Otherwise the IOTEST and the FIFO on the B side of the design will not get a clock.	1
		13	fullwordinterface_ena	When asserted the input interface is changed to use the full 14-bits for each word, instead of dual 7-bit buses for two half words.	1
		12	64cnt_ena	This enables the resetting of the alarms after 64 good samples with the goal of removing unnecessary errors. For instance on a lab board, when checking the setup/hold through IO TEST, there may initially be errors, but once the test is up and running everything works. Setting this bit removes the need for a SIF write to clear the alarm register.	0
		11	dacclkgone_ena	This allows the DACCLK gone signal from the clock monitor to be used to shut the output off.	0
		10	dataclkgone_ena	This allows the DATACLK gone signal from the clock monitor to be used to shut the output off.	0
		9	collision_ena	This allows the collision alarm from the FIFO to shut the output off	0
		8	reserved	Reserved.	0
		7	daca_compliment	When asserted the output to the DACA is complimented. This allows the user of the chip to effectively change the + and – designations of the DAC output pins.	0
		6	dacb_compliment	When asserted the output to the DACB is complimented. This allows the user of the chip to effectively change the + and – designations of the DAC output pins.	0
		5	sif_sync	This is the SIF_SYNC signal. Whatever is programmed into this bit will be used as the chip sync when SIF_SYNC mode is enabled.Design is sensitive to rising edges so programming from 0->1 is when the sync pulse is generated. 1->0 has no effect.	0
		4	sif_sync_ena	When asserted enable SIF_SYNC mode.	0
		3	alarm_2away_ena	When asserted alarms from the FIFO that represent the pointers being 2 away are enabled	1
		2	alarm_1away_ena	When asserted alarms from the FIFO that represent the pointers being 1 away are enabled	1
		1	alarm_collision_ena	When asserted the collision of FIFO pointers causes an alarm to be generated	1
		0	reserved	reserved	0

Register name: config2 – Address: 0x02, Default: 0x3FFF

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config2	0x02	15	reserved	reserved	0
		14	reserved	reserved	0
		13:0	lvdsdata_ena	These 14 bits are individual enables for the 14 input pin receivers. bits(13:7) turn on Da(6:0) where as bits(6:0) enable Db(6:0).	0x3FFF

Register name: config3 – Address: 0x03, Default: 0x0000

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config3	0x03	15:13	datadlya	Controls the delay of the A data inputs through the LVDS receivers. 0= no additional delay and each LSB adds a nominal 80ps.	000
		12:10	clkdlya	Controls the delay of the A data clock input through the LVDS receivers. 0= no additional delay and each LSB adds a nominal 80ps.	000
		9:7	datadlyb	Controls the delay of the B data inputs through the LVDS receivers. 0= no additional delay and each LSB adds a nominal 80ps.	000
		6:4	clkdlyb	Controls the delay of the B data clock input through the LVDS receivers. 0= no additional delay and each LSB adds a nominal 80ps.	000
		3	extref_ena	Enables external reference for the DAC when set.	0
		2:1	reserved	reserved	00
		0	dual_clock_ena	When asserted it tells the LVDS input circuit that there are two individual data clocks. NOTE: must be in SIF_SYNC mode.	0

Register name: config4 – Address: 0x04, Default: 0x0000

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config4 WRITE TO CLEAR/ No RESET value	0x04	15:14	reserved	reserved	00
		13:0	iotest_results	The values of these bits tell which bit in the input word failed during the io-test pattern comparison. 13:7 match up with the 7 bits from port A and 6:0 match up with bits from port B.	0x0000

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

Register name: config5 – Address: 0x05, Default: 0x0000

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config5 WRITE TO CLEAR	0x05	15	alarm_from_ zerochka	When this bit is asserted the FIFO A write pointer has an all zeros pattern in it. Since this pointer is a shift register, all zeros will cause the input point to be stuck until the next sync. The result could be a repeated 8T pattern at the output if the mixer is off and no syncs occur. Check for this error will tell the user that another sync is necessary to restart the FIFO write pointer.	0
		14	alarm_from_ zerochkb	When this bit is asserted the FIFO B write pointer has an all zeros pattern in it. Since this pointer is a shift register, all zeros will cause the input point to be stuck until the next sync. The result could be a repeated 8T pattern at the output if the mixer is off and no syncs occur. Check for this error will tell the user that another sync is necessary to restart the FIFO write pointer.	0
		13:11	alarms_from_ fifoa	These bits report the FIFO A pointer status. 000: All fine 001: Pointers are 2 away 01X: Pointers are 1 away 1XX: FIFO Pointer collision	000
		10:8	alarms_from_ fifob	These bits report the FIFO B pointer status. 000: All fine 001: Pointers are 2 away 01X: Pointers are 1 away 1XX: FIFO Pointer collision	0
		7	alarm_dacclk_ gone	Bit gets asserted when the DACCLK has been stopped long for enough cycles to be caught. The number of cycles varies with interpolation.	0
		6	alarm_dataclk_ gone	Bit gets asserted when the DATACLK has been stopped long for enough cycles to be caught. The number of cycles varies with interpolation.	0
		5	clock_gone	This bit gets set when either alarm_dacclk_gone or alarm_dataclk_gone are asserted. It controls the output of the CDRV_SER block. When high, the CDRV_SER block will output "0x8000" for each output connected to a DAC. The bit must be written to '0' for CDRV_SER outputs to resume normal operation.	0
		4	alarm_from_ iotesta	This is asserted when the input data pattern does not match the pattern in the iotest_pattern registers.	0
		3	alarm_from_ iotestb	This is asserted when the input data pattern does not match the pattern in the iotest_pattern registers.	0
		2	reserved	reserved	0
		1	reserved	reserved	0
		0	reserved	reserved	0

Register name: config6 – Address: 0x06, Default: 0x0000

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config6 No RESET Value	0x06	15:8	tempdata	This the output from the chip temperature sensor. NOTE: when reading these bits the SIF interface must be extremely slow, 1MHz range.	0x00
		7:2	fuse_cntl	These are the values of the blown fuses and are used to determine the available functionality in the chip. (** NOTE **) These bits are READ_ONLY and allow the user to check what features have been disabled in the device. bit5 = 1: Forces Full Word interface bit4 = 1: reserved bit3 = 1: reserved bit2 = 1: Forces Single DAC Mode. Note: This does not force the channel B in sleep mode. In order to do so, user needs to program the sleepb SPI bit (config10, bit 5) to "1". bit1:0 : Forces a different bits size. "00" 14bit "01" 12bit "10" 10bit "11" 10bit	0x00
		1	reserved	reserved	0
		0	reserved	reserved	0

Register name: config7 – Address: 0x07, Default: 0xFFFF

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config7	0x07	15:0	alarms_mask	Each bit is used to mask an alarm. Assertion masks the alarm: bit15 = alarm_mask_zerocka bit14 = alarm_mask_zerockb bit13 = alarm_mask_fifoa_collision bit12 = alarm_mask_fifoa_1away bit11 = alarm_mask_fifoa_2away bit10 = alarm_mask_fifob_collision bit9 = alarm_mask_fifob_1away bit8 = alarm_mask_fifob_2away bit7 = alarm_mask_dacclk_gone bit6 = alarm_mask_dataclk_gone bit5 = Masks the signal which turns off the DAC output when a clock or collision occurs. This bit has no effect on the PAD_ALARM output. bit4 = alarm_mask_iotesta bit3 = alarm_mask_iotestb bit2 = bit1 = bit0 =	0xFFFF

Register name: config8 – Address: 0x08, Default: 0x4000

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config8	0x08	15:13	reserved	reserved	010
		12:0	qmc_offseta	The DAC A offset correction. The offset is measured in DAC LSBs.	0x0000

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

Register name: config9 – Address: 0x09, Default: 0x8000

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config9 AUTO SYNC	0x09	15:13	fifo_offset	This is the starting point for the READ_POINTER in the FIFO block. The READ_POINTER is set to this location when a sync occurs on the DACCLK side of the FIFO.	100
		12:0	qmc_offsetb	The DAC B offset correction. The offset is measured in DAC LSBs. NOTE: Writing this register causes an autosync to be generated in the QMOFFSET block.	0x0000

Register name: config10 – Address: 0x0A, Default: 0xF080

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config10	0x0A	15:12	coarse_dac	Scales the output current is 16 equal steps. $\frac{V_{refIO}}{R_{bias}} \times (\text{mem_coarse_daca} + 1)$	1111
		11	fuse_sleep	Put the fuses to sleep when set high.	0
		10	reserved	reserved	0
		9	reserved	reserved	0
		8	tsense_sleep	When asserted the temperature sensor is put to sleep.	0
		7	clkrecv_ena	Turn on the DAC CLOCK receiver block when asserted.	1
		6	sleepa	When asserted DACA is put to sleep.	0
		5	sleepb	When asserted DACB is put to sleep. Note: This bit needs to be programmed to "1" for single DAC mode.	0
		4:0	reserved	reserved	00000

Register name: config11 – Address: 0x0B, Default: 0x1111

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config11	0x0B	15:12	reserved	reserved	0001
		11:8	reserved	reserved	0001
		7:4	reserved	reserved	0001
		3:0	reserved	reserved	0001

Register name: config12 – Address: 0x0C, Default: 0x3A7A

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config12	0x0C	15:14	reserved	reserved	00
		13:0	iotest_pattern0	This is dataword0 in the IO test pattern. It is used with the seven other words to test the input data. NOTE: This word should be aligned with the rising edge of SYNC when testing the IO interface.	0x3A7A

Register name: config13 – Address: 0x0D, Default: 0x36B6

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config13	0x0D	15:14	reserved	reserved	00
		13:0	iotest_pattern1	This is dataword1 in the IO test pattern. It is used with the seven other words to test the input data.	0x36B6

Register name: config14 – Address: 0x0E, Default: 0x2AEA

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config14	0x0E	15:14	reserved	reserved	00
		13:0	iotest_pattern2	This is dataword2 in the IO test pattern. It is used with the seven other words to test the input data.	0x2AEA

Register name: config15 – Address: 0x0F, Default: 0x0545

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config15	0x0F	15:14	reserved	reserved	00
		13:0	iotest_pattern3	This is dataword3 in the IO test pattern. It is used with the seven other words to test the input data.	0x0545

Register name: config16 – Address: 0x10, Default: 0x1A1A

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config16	0x10	15:14	reserved	reserved	00
		13:0	iotest_pattern4	This is dataword4 in the IO test pattern. It is used with the seven other words to test the input data.	0x1A1A

Register name: config17 – Address: 0x11, Default: 0x1616

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config17	0x11	15:14	reserved	reserved	00
		13:0	iotest_pattern5	This is dataword5 in the IO test pattern. It is used with the seven other words to test the input data.	0x1616

Register name: config18 – Address: 0x12, Default: 0x2AAA

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config18	0x12	15:14	reserved	reserved	00
		13:0	iotest_pattern5	This is dataword6 in the IO test pattern. It is used with the seven other words to test the input data.	0x2AAA

Register name: config19 – Address: 0x13, Default: 0x06C6

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config19	0x13	15:14	reserved	reserved	00
		13:0	iotest_pattern7	This is dataword7 in the IO test pattern. It is used with the seven other words to test the input data.	0x06C6

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

Register name: config20– Address: 0x14, Default: 0x0000

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config20	0x14	15	sifdac_ena	When asserted the DAC output is set to the value in sifdac. This can be used for trim setting and other static tests.	0
		14	reserved	reserved	0
		13:0	sifdac	This is the value that is sent to the DACs when sifdac_ena is asserted.	0x0000

Register name: config21– Address: 0x15, Default: 0xFFFF

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config21	0x15	15:0	sleepcntl	<p>This controls what blocks get sent a SLEEP signal when the PAD_SLEEP pin is asserted. Programming a '1' in a bit will pass the SLEEP signal to the appropriate block.</p> <p>bit15 = DAC A bit14 = DAC B bit13 = FUSE Sleep bit12 = Temperature Sensor bit11 = Clock Receiver bit10 = LVDS DATA Receivers bit9 = LVDS SYNC Receiver bit8 = PECL ALIGN Receiver bit7 = LVDS DATACLK Receiver bit6 = bit5 = bit4 = bit3 = bit2 = bit1 = bit0 =</p>	0xFFFF

Register name: config22– Address: 0x16

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config22 READ ONLY	0x16	15:0	fa002_data(15:0)	Lower 16bits of the DIE ID word	

Register name: config23– Address: 0x17

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config23 READ ONLY	0x17	15:0	fa002_data(31:16)	Lower middle 16bits of the DIE ID word	

Register name: config24– Address: 0x18

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config24 READ ONLY	0x18	15:0	fa002_data(47:32)	Upper middle 16bits of the DIE ID word	

Register name: config25– Address: 0x19

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config25 READ ONLY	0x19	15:0	fa002_ data(63:48)	Upper 16bits of the DIE ID word	

Register name: config127– Address: 0x7F, Default: 0x0045

Register Name	Addr (Hex)	Bit	Name	Function	Default Value
config127 READ ONLY/No RESET Value	0x7F	15:14	reserved	reserved	00
		13:12	reserved	reserved	00
		11:10	reserved	reserved	00
		9:8	reserved	reserved	00
		7	reserved	reserved	0
		6	titest_voh	A fixed '1' that can be used to test the Voh at the SIF output.	1
		5	titest_vol	A fixed '0' that can be used to test the Vol at the SIF output.	0
		4:3	vendorid	Fixed at 01	01
		2:0	versionid	Chip version	001

Synchronization Modes

There are three modes of syncing included in the DAC3174.

- **NORMAL Dual Sync** – The SYNC pin is used to align the input side of the FIFO (write pointers) with the A(0) sample. The ALIGN pin is used to reset the output side of the FIFO (read pointers) to the offset value. Multiple chip alignment can be accomplished with this kind of syncing.
- **SYNC ONLY** – In this mode only the SYNC pin is used to sync both the read and write pointers of the FIFO. There is an asynchronized handoff between the DATACLK and DACCLK when using this mode, therefore it is impossible to accurately align multiple chips closer than 2 or 3T.
- **SIF_SYNC** – When neither SYNC nor ALIGN are used, a programmable SYNC pulse can be used to sync the design. However, the same issues as SYNC ONLY apply. There is an asynchronized handoff between the serial clock domain and the two sides of the FIFO. Because of the asynchronous nature of the SIF_SYNC it is impossible to align the sync up with any sample at the input. **NOTE: SIF_SYNC mode is the only synchronization mode supported in dual bus mode.**

NOTE

When ALIGNP/N are not used, it is recommended to clear the alignrx_ena register (config1, bit 4), and tie ALIGNP to DIGVDD18 and ALIGNN to GROUND. When SYNC P/N are not used, it is recommended to clear register syncrx_ena (config0, bit3), and the unused SYNC P/N pins can be left open or tied to GROUND.

DAC3174

SLAS837A – APRIL 2013 – REVISED MAY 2013

www.ti.com

Alarm Monitoring

DAC3174 includes flexible alarm monitoring that can be used to alert a possible malfunction scenario. All alarm events can be accessed either through the SIP registers and/or through the ALARM pin. Once an alarm is set, the corresponding alarm bit in register config5 must be reset through the serial interface to allow further testing. The set of alarms includes the following conditions:

Zero check alarm

- `Alarm_from_zerochk`. Occurs when the FIFO write pointer has an all zeros pattern. Since the write pointer is a shift register, all zeros will cause the input point to be stuck until the next sync event. When this happens a sync to the FIFO block is required.

FIFO alarms

- `alarm_from_fifo`. Occurs when there is a collision in the FIFO pointers or a collision event is close.
- `alarm_fifo_2away`. Pointers are within two addresses of each other.
- `alarm_fifo_1away`. Pointers are within one address of each other.
- `alarm_fifo_collision`. Pointers are equal to each other.

Clock alarms

- `clock_gone`. Occurs when either the DACCLK or DATALOCK have been stopped.
- `alarm_dacclk_gone`. Occurs when the DACCLK has been stopped.
- `alarm_dataclk_gone`. Occurs when the DATACLK has been stopped.

Pattern checker alarm

- `alarm_from_iotest`. Occurs when the input data pattern does not match the pattern key.

To prevent unexpected DAC outputs from propagating into the transmit channel chain, DAC3174 includes a feature that disables the outputs when a catastrophic alarm occurs. The catastrophic alarms include FIFO pointer collision, the loss DACCLK or the loss of DATACLK. When any of these alarms occur the internal TXenable signal is driven low, causing a zeroing of the data going to the DAC in <10T. One caveat is if both clocks stop, the circuit cannot determine clock loss so no alarms are generated; therefore, no zeroing of output data occurs.

REVISION HISTORY

Changes from Original (April 2013) to Revision A	Page
• Deleted PRODUCT PREVIEW banner for device release	1

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
DAC3174IRGC25	ACTIVE	VQFN	RGC	64	25	Green (RoHS & no Sb/Br)	CU NIPDAU Call TI	Level-3-260C-168 HR	-40 to 85	DAC3174I	Samples
DAC3174IRGCR	ACTIVE	VQFN	RGC	64	2000	Green (RoHS & no Sb/Br)	CU NIPDAU Call TI	Level-3-260C-168 HR	-40 to 85	DAC3174I	Samples
DAC3174IRGCT	ACTIVE	VQFN	RGC	64	250	Green (RoHS & no Sb/Br)	CU NIPDAU Call TI	Level-3-260C-168 HR	-40 to 85	DAC3174I	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBsolete: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

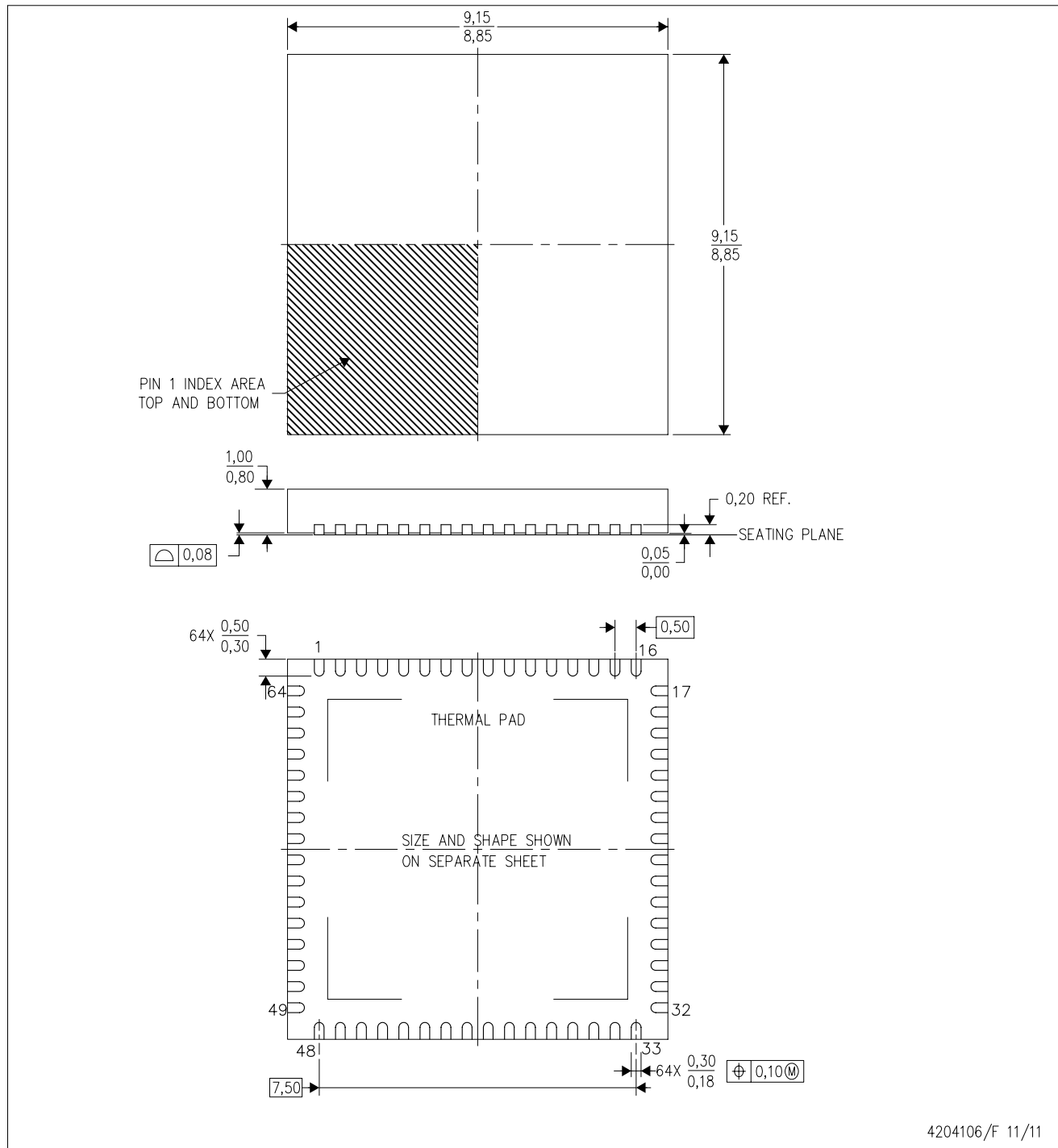
(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and

continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

RGC(S-PVQFN-N64) CUSTOM DEVICE PLASTIC QUAD FLATPACK NO-LEAD



- NOTES:
- All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5-1994.
 - This drawing is subject to change without notice.
 - Quad Flatpack, No-leads (QFN) package configuration.
 - The package thermal pad must be soldered to the board for thermal and mechanical performance.
 - See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.

THERMAL PAD MECHANICAL DATA

RGC (S-PVQFN-N64)

PLASTIC QUAD FLATPACK NO-LEAD

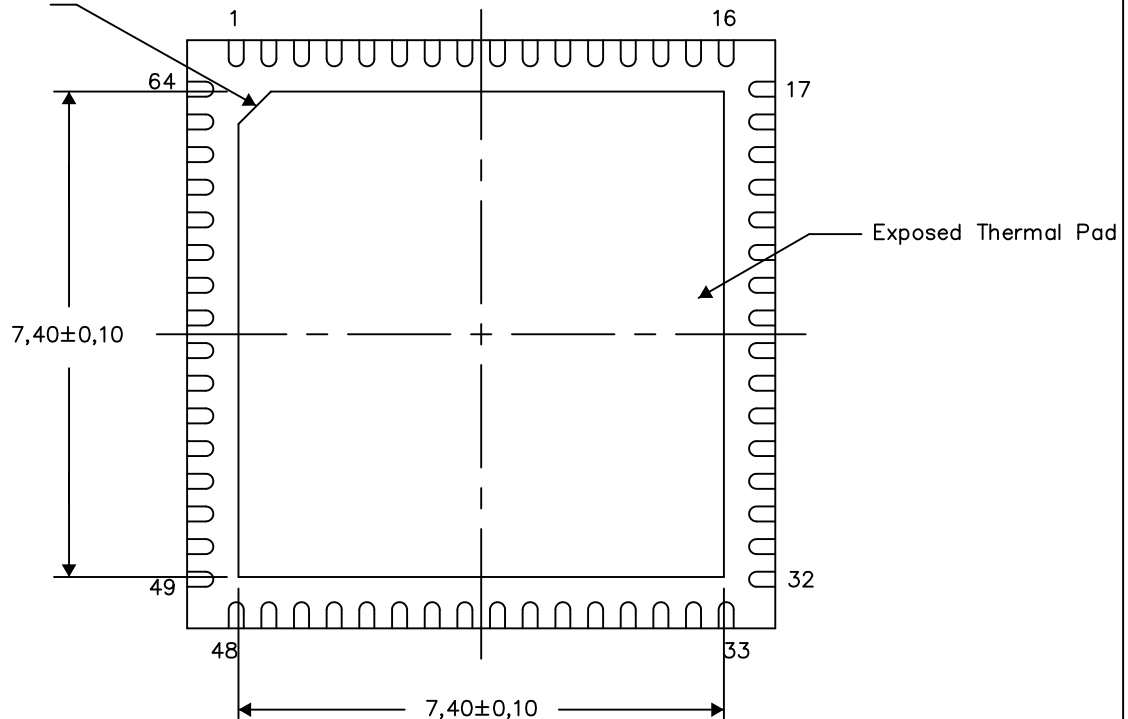
THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

PIN 1 INDICATOR
CO,35



Bottom View

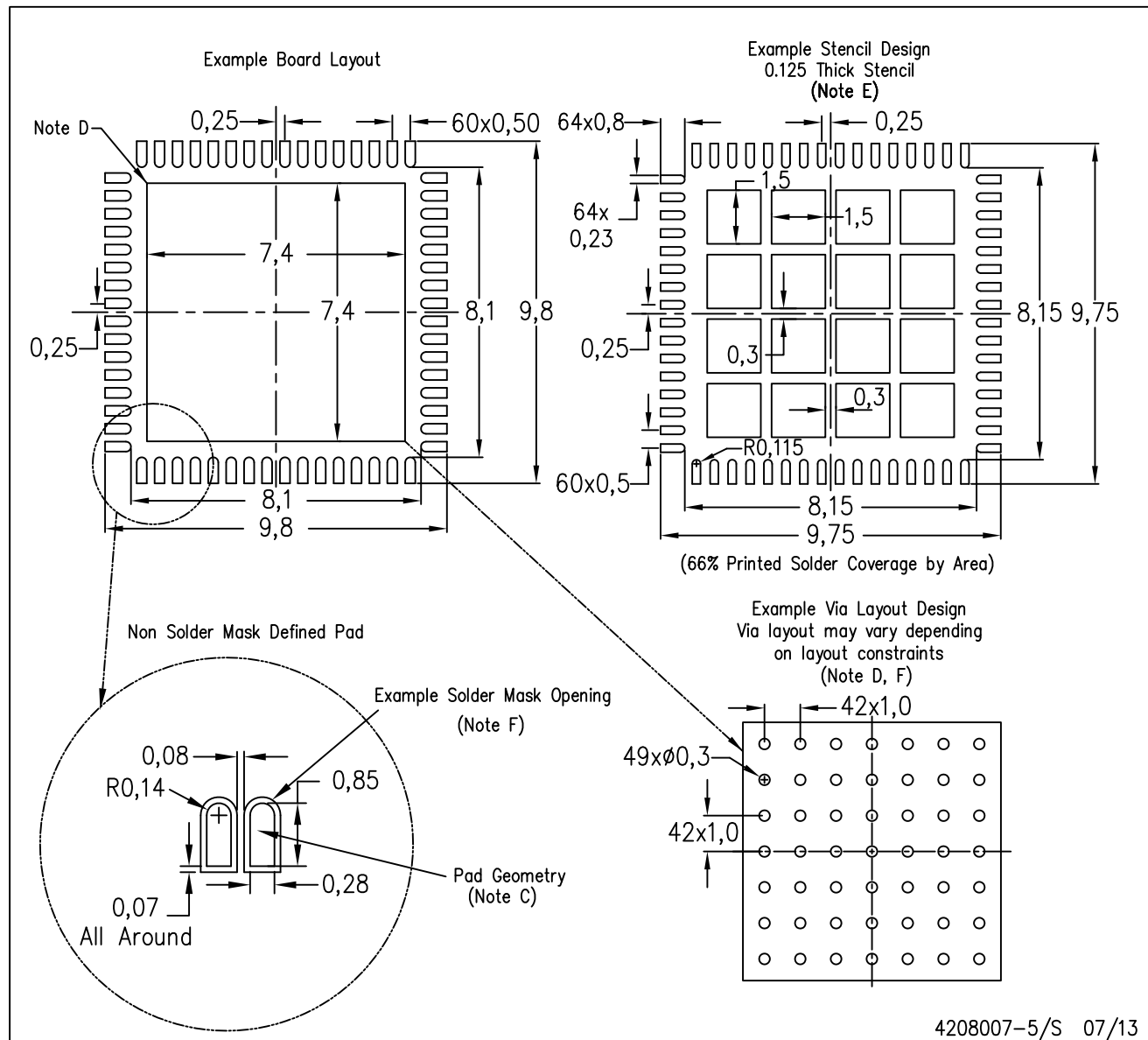
Exposed Thermal Pad Dimensions

4206192-4/AB 10/13

NOTE: A. All linear dimensions are in millimeters

RGC (S-PVQFN-N64)

PLASTIC QUAD FLATPACK NO-LEAD



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Publication IPC-7351 is recommended for alternate designs.
 - D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>.
 - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
 - F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in thermal pad.

IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All semiconductor products (also referred to herein as "components") are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its components to the specifications applicable at the time of sale, in accordance with the warranty in TI's terms and conditions of sale of semiconductor products. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by applicable law, testing of all parameters of each component is not necessarily performed.

TI assumes no liability for applications assistance or the design of Buyers' products. Buyers are responsible for their products and applications using TI components. To minimize the risks associated with Buyers' products and applications, Buyers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI components or services are used. Information published by TI regarding third-party products or services does not constitute a license to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of significant portions of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI components or services with statements different from or beyond the parameters stated by TI for that component or service voids all express and any implied warranties for the associated TI component or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyer acknowledges and agrees that it is solely responsible for compliance with all legal, regulatory and safety-related requirements concerning its products, and any use of TI components in its applications, notwithstanding any applications-related information or support that may be provided by TI. Buyer represents and agrees that it has all the necessary expertise to create and implement safeguards which anticipate dangerous consequences of failures, monitor failures and their consequences, lessen the likelihood of failures that might cause harm and take appropriate remedial actions. Buyer will fully indemnify TI and its representatives against any damages arising out of the use of any TI components in safety-critical applications.

In some cases, TI components may be promoted specifically to facilitate safety-related applications. With such components, TI's goal is to help enable customers to design and create their own end-product solutions that meet applicable functional safety standards and requirements. Nonetheless, such components are subject to these terms.

No TI components are authorized for use in FDA Class III (or similar life-critical medical equipment) unless authorized officers of the parties have executed a special agreement specifically governing such use.

Only those TI components which TI has specifically designated as military grade or "enhanced plastic" are designed and intended for use in military/aerospace applications or environments. Buyer acknowledges and agrees that any military or aerospace use of TI components which have **not** been so designated is solely at the Buyer's risk, and that Buyer is solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI has specifically designated certain components as meeting ISO/TS16949 requirements, mainly for automotive use. In any case of use of non-designated products, TI will not be responsible for any failure to meet ISO/TS16949.

Products

Audio	www.ti.com/audio
Amplifiers	amplifier.ti.com
Data Converters	dataconverter.ti.com
DLP® Products	www.dlp.com
DSP	dsp.ti.com
Clocks and Timers	www.ti.com/clocks
Interface	interface.ti.com
Logic	logic.ti.com
Power Mgmt	power.ti.com
Microcontrollers	microcontroller.ti.com
RFID	www.ti-rfid.com
OMAP Applications Processors	www.ti.com/omap
Wireless Connectivity	www.ti.com/wirelessconnectivity

Applications

Automotive and Transportation	www.ti.com/automotive
Communications and Telecom	www.ti.com/communications
Computers and Peripherals	www.ti.com/computers
Consumer Electronics	www.ti.com/consumer-apps
Energy and Lighting	www.ti.com/energy
Industrial	www.ti.com/industrial
Medical	www.ti.com/medical
Security	www.ti.com/security
Space, Avionics and Defense	www.ti.com/space-avionics-defense
Video and Imaging	www.ti.com/video

TI E2E Community

e2e.ti.com