

1.8-V MICROPOWER CMOS OPERATIONAL AMPLIFIER ZERO-DRIFT SERIES

Check for Samples: OPA2333-Q1

FEATURES

Qualified for Automotive Applications

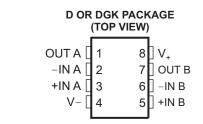
Low Offset Voltage: 23 μV (Max)
0.01-Hz to 10-Hz Noise: 1.1 μV_{PP}

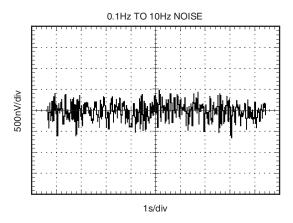
Quiescent Current: 17 μA
Single-Supply Operation
Supply Voltage: 1.8 V to 5.5 V

• Rail-to-Rail Input/Output

APPLICATIONS

- Transducer Applications
- Temperature Measurements
- Electronic Scales
- Medical Instrumentation
- Battery-Powered Instruments
- Handheld Test Equipment





DESCRIPTION/ORDERING INFORMATION

The OPA2333A CMOS operational amplifiers use a proprietary auto-calibration technique to simultaneously provide very low offset voltage (10 μ V max) and near-zero drift over time and temperature. These miniature high-precision low-quiescent-current amplifiers offer high-impedance inputs that have a common-mode range 100 mV beyond the rails and rail-to-rail output that swings within 50 mV of the rails. Single or dual supplies as low as 1.8 V (\pm 0.9 V) and up to 5.5 V (\pm 2.75 V) may be used. They are optimized for low-voltage single-supply operation.

The OPA2333A offers excellent common-mode rejection ratio (CMRR) without the crossover associated with traditional complementary input stages. This design results in superior performance for driving analog-to-digital converters (ADCs) without degradation of differential linearity.

The OPA2333A is specified for operation from -40°C to 125°C.

ORDERING INFORMATION(1)

T _A	PACK	AGE ⁽²⁾	ORDERABLE PART NUMBER	TOP-SIDE MARKING		
40°C to 425°C	SOIC - D	Reel of 2500	OPA2333AQDRQ1	02333Q		
–40°C to 125°C	MSOP - DGK	Reel of 2500	OPA2333AQDGKRQ1	OCOQ		

- (1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.
- (2) Package drawings, thermal data, and symbolization are available at www.ti.com/packaging.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



ABSOLUTE MAXIMUM RATINGS(1)

over operating free-air temperature range (unless otherwise noted)

V_{CC}	Supply voltage	Supply voltage					
VI	Input voltage, signal input terminals (2)	−0.3 V to (V+) + 0.3					
I _{O(SS)}	Output short-circuit circuit ⁽³⁾	Continuous					
T _A	Operating free-air temperature range	-40°C to 125°C					
T_J	Maximum operating virtual-junction temperature	150°C					
T _{stg}	Storage temperature range		−65°C to 150°C				
ECD.	Flootrootatic discharge (FCD) rating	Human-Body Model (HBM)	2000 V				
ESD	Electrostatic discharge (ESD) rating	Charged-Device Model (CDM)	1000 V				

⁽¹⁾ Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS: V_s = 1.8 V to 5.5 V

Boldface limits apply over the specified temperature range, $T_A = -40^{\circ}C$ to 125°C. At $T_A = 25^{\circ}C$, $R_L = 10 \text{ k}\Omega$ connected to $V_S/2$, $V_{CM} = V_S/2$, and $V_{OUT} = V_S/2$ (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
OFFSET VOLTAGE						
Input offset voltage	Vos	$V_S = 5 V$		2	10	μV
over temperature					22	μ V
vs temperature	dV_{OS}/d_T			0.02		μ ۷/°C
vs power supply	PSRR	$V_S = 1.8 \text{ V to } 5.5 \text{ V}$		1	6	μ V/V
Long-term stability ⁽¹⁾				(1)		
Channel separation, dc				0.1		μV/V
INPUT BIAS CURRENT						
Input bias current	Ι _Β			±70	±200	pА
over Temperature				±150		рA
Input offset current	I _{os}			±140	±400	рА
NOISE						
Input voltage noise, f = 0.01 Hz to 1 Hz				0.3		μV_{PP}
Input voltage noise, f = 0.1 Hz to 10 Hz				1.1		μV_{PP}
Input current noise, f = 10 Hz	in			100		fA/\sqrt{Hz}
INPUT VOLTAGE RANGE						
Common mode voltage range	V_{CM}		(V-) - 0.1		(V+) + 0.1	V
Common-Mode Rejection Ratio	CMRR	$(V-) - 0.1 V < V_{CM} < (V+) + 0.1 V$	102	130		dB
INPUT CAPACITANCE						
Differential				2		pF
Common mode				4		pF
OPEN-LOOP GAIN						
Open-loop voltage gain	A OL	$(V-) + 100 \text{ mV} < V_O < (V+) - 100 \text{ mV},$ $R_L = 10 \text{ k}\Omega$	104	130		dB
FREQUENCY RESPONSE						
Gain-bandwidth product	GBW	$C_L = 100 pF$		350		kHz
Slew rate	SR	G = 1		0.16		V/μs

(1) 300-hour life test at 150°C demonstrated randomly distributed variation of approximately 1 μV.

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⁽²⁾ Input terminals are diode clamped to the power-supply rails. Input signals that can swing more than 0.3 V beyond the supply rails should be current limited to 10 mA or less.

⁽³⁾ Short circuit to ground, one amplifier per package



ELECTRICAL CHARACTERISTICS: $V_s = 1.8 \text{ V to } 5.5 \text{ V (continued)}$

Boldface limits apply over the specified temperature range, $T_A = -40^{\circ}C$ to 125°C. At $T_A = 25^{\circ}C$, $R_L = 10 \text{ k}\Omega$ connected to $V_S/2$, $V_{CM} = V_S/2$, and $V_{OUT} = V_S/2$ (unless otherwise noted).

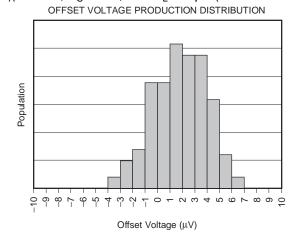
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
OUTPUT							
Voltage output swing from rail		$R_L = 10 \text{ k}\Omega$		30	50	mV	
over temperature		$R_L = 10 \text{ k}\Omega$			85	mV	
Short-circuit current	ISC			±5		mA	
Capacitive load drive	CL						
⁽²⁾ Open-loop output impedance		$f = 350 \text{ kHz}, I_0 = 0$		2		kΩ	
POWER SUPPLY							
Specified voltage range	Vs		1.8		5.5	V	
Quiescent current per amplifier	IQ	I _O = 0		17	25	μΑ	
over temperature					30	μ A	
Turn-on time		$V_S = 5 V$		100		μS	
TEMPERATURE RANGE							
Specified range			-40		125	°C	
Operating range			-40		125	°C	
Storage range			-65		150	°C	
Thermal resistance	θ_{JA}	SO-8 (D)		150		°C/W	
		MSOP-8 (DGK)		172.47		°C/W	

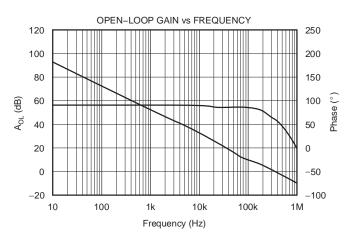
⁽²⁾ See Typical Characteristics.

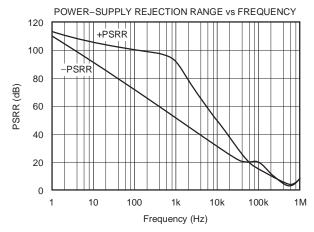


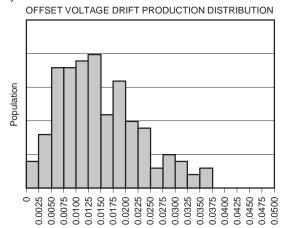
TYPICAL CHARACTERISTICS

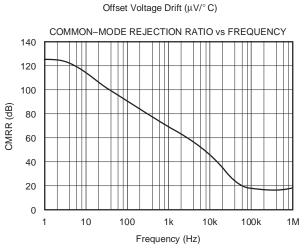
At $T_A = 25$ °C, $V_S = 5$ V, and $C_L = 0$ pF (unless otherwise noted)

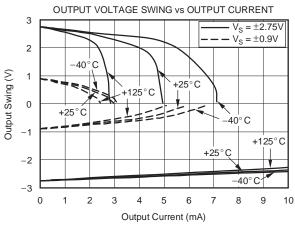






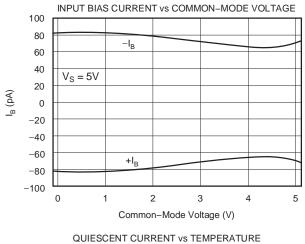


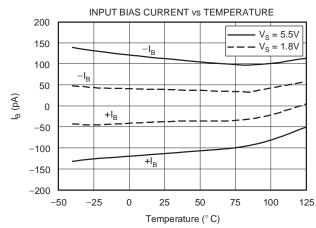


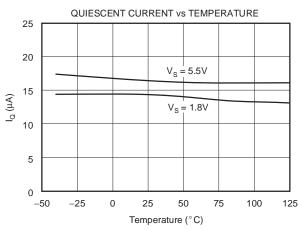


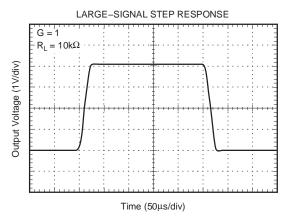


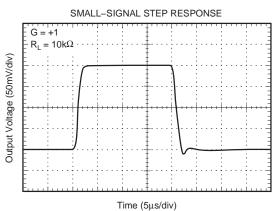
TYPICAL CHARACTERISTICS (continued)

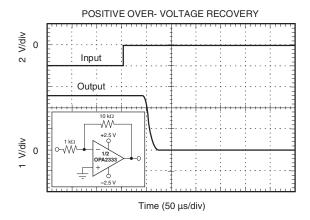






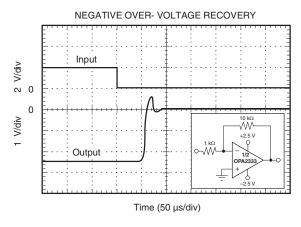


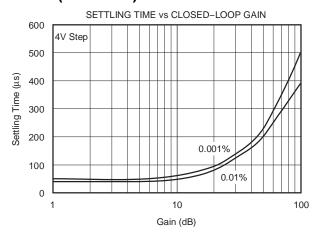


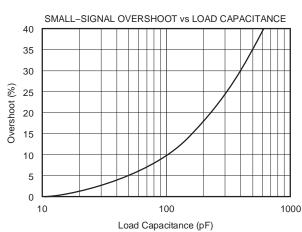


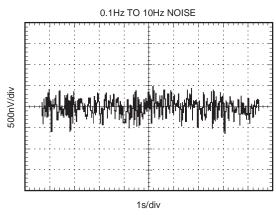


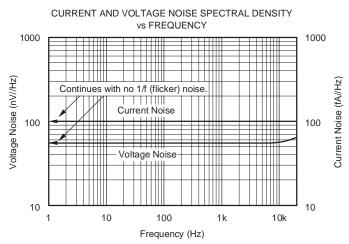
TYPICAL CHARACTERISTICS (continued)













APPLICATION INFORMATION

The OPA2333A op amps are unity-gain stable and free from unexpected output phase reversal. They use a proprietary auto-calibration technique to provide low offset voltage and very low drift over time and temperature. For lowest offset voltage and precision performance, circuit layout and mechanical conditions should be optimized. Avoid temperature gradients that create thermoelectric (Seebeck) effects in the thermocouple junctions formed from connecting dissimilar conductors. These thermally-generated potentials can be made to cancel by ensuring they are equal on both input terminals. Other layout and design considerations include:

- Use low thermoelectric-coefficient conditions (avoid dissimilar metals)
- · Thermally isolate components from power supplies or other heat sources
- Shield op amp and input circuitry from air currents, such as cooling fans

Following these guidelines will reduce the likelihood of junctions being at different temperatures, which can cause thermoelectric voltages of 0.1 μ V/°C or higher, depending on materials used.

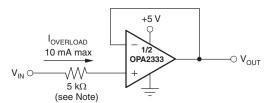
Operating Voltage

The OPA2333A op amps operate over a power-supply range of 1.8 V to 5.5 V (±0.9 V to ±2.75 V). Supply voltages higher than 7 V (absolute maximum) can permanently damage the device. Parameters that vary over supply voltage or temperature are shown in the Typical Characteristics section of this data sheet.

Input Voltage

The OPA2333A input common-mode voltage range extends 0.1 V beyond the supply rails. The device is designed to cover the full range without the troublesome transition region found in some other rail-to-rail amplifiers.

Normally, input bias current is about 70 pA; however, input voltages exceeding the power supplies can cause excessive current to flow into or out of the input pins. Momentary voltages greater than the power supply can be tolerated if the input current is limited to 10 mA. This limitation is easily accomplished with an input resistor(see Figure 1).



NOTE: Current-limiting resistor required if input voltage exceeds supply rails by ≥ 0.5 V.

Figure 1. Input Current Protection

Internal Offset Correction

The OPA2333A op amps use an auto-calibration technique with a time-continuous 350-kHz op amp in the signal path. This amplifier is zero corrected every 8 μ s using a proprietary technique. Upon power up, the amplifier requires approximately 100 μ s to achieve specified V_{OS} accuracy. This design has no aliasing or flicker noise.

Achieving Output Swing to the Op Amp Negative Rail

Some applications require output voltage swings from 0 V to a positive full-scale voltage (such as 2.5 V) with excellent accuracy. With most single-supply op amps, problems arise when the output signal approaches 0 V, near the lower output swing limit of a single-supply op amp. A good single-supply op amp may swing close to single-supply ground, but will not reach ground. The output of the OPA2333A can be made to swing to ground or slightly below on a single-supply power source. To do so requires the use of another resistor and an additional, more negative, power supply than the op amp negative supply. A pulldown resistor may be connected between the output and the additional negative supply to pull the output down below the value that the output would otherwise achieve (see Figure 2).

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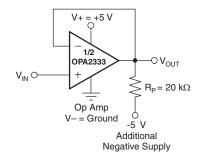


Figure 2. V_{OUT} Range to Ground

The OPA2333A has an output stage that allows the output voltage to be pulled to its negative supply rail, or slightly below, using the technique previously described. This technique only works with some types of output stages. The OPA2333A has been characterized to perform with this technique; however, the recommended resistor value is approximately 20 k Ω . Note that this configuration increases the current consumption by several hundreds of microamps. Accuracy is excellent down to 0 V and as low as -2 mV. Limiting and nonlinearity occurs below -2 mV, but excellent accuracy returns as the output is again driven above -2 mV. Lowering the resistance of the pulldown resistor allows the op amp to swing even further below the negative rail. Resistances as low as 10 k Ω can be used to achieve excellent accuracy down to -10 mV.

General Layout Guidelines

Attention to good layout practices is always recommended. Keep traces short and, when possible, use a printed circuit board (PCB) ground plane with surface-mount components placed as close to the device pins as possible. Place a 0.1-µF capacitor closely across the supply pins. These guidelines should be applied throughout the analog circuit to improve performance and provide benefits, such as reducing the electromagnetic interference (EMI) susceptibility.

Operational amplifiers vary in their susceptibility to radio frequency interference (RFI). RFI can generally be identified as a variation in offset voltage or dc signal levels with changes in the interfering RF signal. The OPA2333A has been specifically designed to minimize susceptibility to RFI and demonstrates remarkably low sensitivity compared to previous-generation devices. Strong RF fields may still cause varying offset levels.

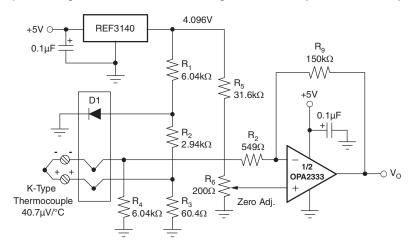


Figure 3. Temperature Measurement

Figure 4 shows the basic configuration for a bridge amplifier.



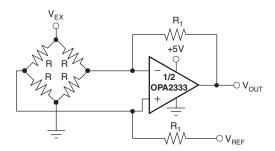
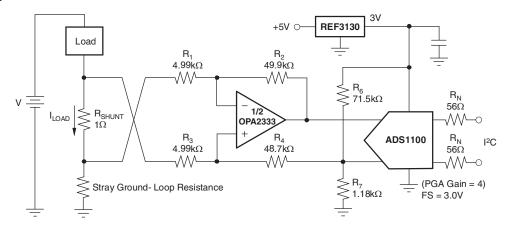


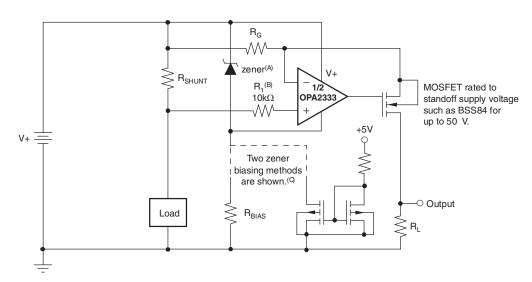
Figure 4. Single Op-Amp Bridge Amplifier

A low-side current shunt monitor is shown in Figure 5. R_N are operational resistors used to isolate the ADS1100 from the noise of the digital I^2C bus. Since the ADS1100 is a 16-bit converter, a precise reference is essential for maximum accuracy. If absolute accuracy is not required, and the 5-V power supply is sufficiently stable, the REF3130 may be omitted.



NOTE: 1% resistors provide adequate common-mode rejection at small ground-loop errors.

Figure 5. Low-Side Current Monitor



- A. Zener rated for op amp supply capability (that is, 5.1 V for OPA2333).
- B. Current-limiting resistor
- C. Choose Zener biasing resistor or dual NMOSFETs (FDG6301N, NTJD4001N, or Si1034).

Figure 6. High-Side Current Monitor

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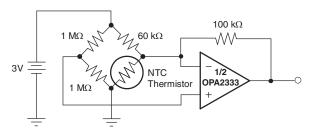


Figure 7. Thermistor Measurement

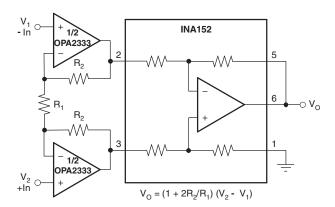
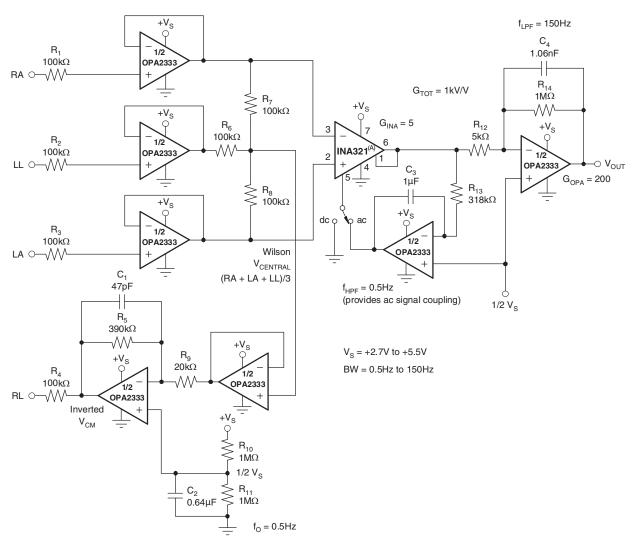


Figure 8. Precision Instrumentation Amplifier





A. Other instrumentation amplifiers can be used, such as the INA326, which has lower noise, but higher quiescent current.

Figure 9. Single-Supply, Very-Low-Power ECG Circuit





11-Apr-2013

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
OPA2333AQDGKRQ1	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR		OCOQ	Samples
OPA2333AQDRQ1	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	02333Q	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

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OTHER QUALIFIED VERSIONS OF OPA2333-Q1:







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• Catalog: OPA2333

NOTE: Qualified Version Definitions:

• Catalog - TI's standard catalog product

PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA2333AQDGKRQ1	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.3	1.3	8.0	12.0	Q1
OPA2333AQDRQ1	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
OPA2333AQDGKRQ1	VSSOP	DGK	8	2500	370.0	355.0	55.0
OPA2333AQDRQ1	SOIC	D	8	2500	367.0	367.0	35.0

DGK (S-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
- E. Falls within JEDEC MO-187 variation AA, except interlead flash.



DGK (S-PDSO-G8)

PLASTIC SMALL OUTLINE PACKAGE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.



D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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