

## LMH6572 Triple 2:1 High Speed Video Multiplexer

Check for Samples: [LMH6572](#)

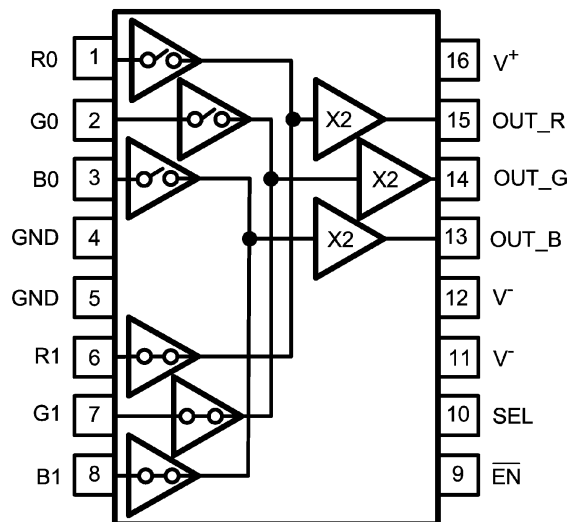
### FEATURES

- 350 MHz, 250 mV –3 dB Bandwidth
- 290 MHz, 2 V<sub>PP</sub> –3 dB Bandwidth
- 10 ns Channel Switching Time
- 90 dB Channel to Channel Isolation @ 5 MHz
- 0.02%, 0.02° Diff. Gain, Phase
- 0.1 dB Gain Flatness to 140 MHz
- 1400 V/μs Slew Rate
- Wide Supply Voltage Range: 6V (±3V) to 12V (±6V)
- –78 dB HD2 @ 10 MHz
- –75 dB HD3 @ 10 MHz

### APPLICATIONS

- RGB Video Router
- Multi Input Video Monitor
- Fault Tolerant Data Switch

### Connection Diagram



**Figure 1. 16-Pin SSOP Package**  
See Package Number DBQ0016A  
Top View

### DESCRIPTION

The LMH™ 6572 is a high performance analog multiplexer optimized for professional grade video and other high fidelity, high bandwidth analog applications. The LMH6572 provides a 290MHz bandwidth at 2 V<sub>PP</sub> output signal levels. The 140 MHz of .1 dB bandwidth and a 1500 V/μs slew rate make this part suitable for High Definition Television (HDTV) and High Resolution Multimedia Video applications.

The LMH6572 supports composite video applications with its 0.02% and 0.02° differential gain and phase errors for NTSC and PAL video signals while driving a single, back terminated 75Ω load. The LM6572 can deliver 80 mA linear output current for driving multiple video load applications.

The LMH6572 has an internal gain of 2 V/V (+6 dBv) for driving back terminated transmission lines at a net gain of 1 V/V (0 dBv).

The LMH6572 is available in the SSOP package.

### Truth Table

SEL	EN	OUT
0	0	CH 1
1	0	CH 0
X	1	Disable

**PRODUCT PREVIEW**



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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### Absolute Maximum Ratings<sup>(1)(2)</sup>

ESD Tolerance <sup>(3)</sup>	Human Body Model	2000V
	Machine Model	200V
Supply Voltage ( $V^+ - V^-$ )		13.2V
$I_{OUT}$ <sup>(4)</sup>		130 mA
Input Voltage Range		$\pm(V_S)$
Maximum Junction Temperature		+150°C <sup>(3)</sup>
Storage Temperature Range		-65°C to +150°C
Soldering Information	Infrared or Convection (20 sec)	235°C
	Wave Soldering (10 sec)	260°C

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not ensured. For ensured specifications, see the Electrical Characteristics tables.
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
- (3) Human Body Model, 1.5 k $\Omega$  in series with 100 pF. Machine Model 0 $\Omega$  in series with 200 pF.
- (4) The maximum output current ( $I_{OUT}$ ) is determined by the device power dissipation limitations. See the [Power Dissipation](#) section of the Application Section for more details. A short circuit condition should be limited to 5 seconds or less.

### Operating Ratings<sup>(1)</sup>

Operating Temperature			-40 °C to 85 °C
Supply Voltage Range			6V to 12V
Thermal Resistance			
Package	16-Pin SSOP	( $\theta_{JA}$ )	125°C/W
		( $\theta_{JC}$ )	36°C/W

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not ensured. For ensured specifications, see the Electrical Characteristics tables.

## ±5V Electrical Characteristics

Unless otherwise specified,  $V_S = \pm 5V$ ,  $R_L = 100\Omega$ .

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Typ	Max	Units
<b>Frequency Domain Performance</b>						
SSBW	–3 dB Bandwidth	$V_{OUT} = 0.25 V_{PP}$		350		MHz
LSBW	–3 dB Bandwidth <sup>(2)</sup>	$V_{OUT} = 2 V_{PP}$	250	290		MHz
.1 dBBW	0.1 dB Bandwidth	$V_{OUT} = 0.25 V_{PP}$		140		MHz
DG	Differential Gain	$R_L = 150\Omega$ , $f = 4.43$ MHz		0.02		%
DP	Differential Phase	$R_L = 150\Omega$ , $f = 4.43$ MHz		0.02		deg
<b>Time Domain Response</b>						
TRS	Channel to Channel Switching Time	Logic Transition to 90% Output		10		ns
	Enable and Disable Times	Logic Transition to 90% or 10% Output		11		ns
TRL	Rise and Fall Time	2V Step		1.5		ns
TSS	Settling Time to 0.05%	2V Step		17		ns
OS	Overshoot	4V Step		5		%
SR	Slew Rate <sup>(2)</sup>	4V Step	1200	1400		V/ $\mu$ s
<b>Distortion</b>						
HD2	2 <sup>nd</sup> Harmonic Distortion	$2 V_{PP}$ , 10 MHz		–78		dBc
HD3	3 <sup>rd</sup> Harmonic Distortion	$2 V_{PP}$ , 10 MHz		–75		dBc
IMD	3 <sup>rd</sup> Order Intermodulation Products	10 MHz, Two tones $2 V_{PP}$ at Output		–80		dBc
<b>Equivalent Input Noise</b>						
VN	Voltage	>1 MHz, Input Referred		5		nV/ $\sqrt{Hz}$
ICN	Current	>1 MHz, Input Referred		5		pA/ $\sqrt{Hz}$
<b>Static, DC Performance</b>						
GAIN	Voltage Gain			2.0		V/V
	Gain Error <sup>(3)</sup>	No load, with respect to nominal gain of 2.00 V/V.		±0.3	±0.5 ±0.7	%
	Gain Error	$R_L = 50\Omega$ , with respect to nominal gain of 2.00 V/V		0.3		%
VIO	Output Offset Voltage <sup>(3)</sup>	$V_{IN} = 0V$		1	±14 ±17.5	mV
DVIO	Average Drift			27		$\mu$ V/ $^{\circ}$ C
IBN	Input Bias Current <sup>(3) (4)</sup>	$V_{IN} = 0V$		–1.4	±5.0 ±5.6	$\mu$ A
DIBN	Average Drift			7		nA/ $^{\circ}$ C
PSRR	Power Supply Rejection Ratio <sup>(3)</sup>	DC, Input referred	50 48	54		dB
ICC	Supply Current <sup>(3)</sup>	No load	20	23	25 28.5	mA
	Supply Current Disabled <sup>(3)</sup>	No load		2.0	2.2 2.3	mA
VIH	Logic High Threshold <sup>(3)</sup>	Select & Enable Pins	2.0			V
VIL	Logic Low Threshold <sup>(3)</sup>	Select & Enable Pins			0.8	V
II <sub>L</sub>	Logic Pin Input Current Low <sup>(4)</sup>	Logic Input = 0V		–1	±5.0 ±15	$\mu$ A
II <sub>H</sub>	Logic Pin Input Current High <sup>(4)</sup>	Logic Input = 2.0V	112 100	150	200 210	$\mu$ A

(1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that  $T_J = T_A$ . No specification of parametric performance is indicated in the electrical tables under conditions of internal self heating where  $T_J > T_A$ . See Applications Section for information on temperature de-rating of this device. Min/Max ratings are based on product testing, characterization and simulation. Individual parameters are tested as noted.

(2) Parameters ensured by design.

(3) Parameters ensured by electrical testing at 25° C.

(4) Positive Value is current into device.

## ±5V Electrical Characteristics (continued)

Unless otherwise specified,  $V_S = \pm 5V$ ,  $R_L = 100\Omega$ .

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Typ	Max	Units
<b>Miscellaneous Performance</b>						
RF	Internal Feedback and Gain Set Resistor Values		650 <b>620</b>	800	940 <b>1010</b>	$\Omega$
RODIS	Disabled Output Resistance	Internal Feedback and Gain Set Resistors in Series to Ground	1.3	1.6	1.88	k $\Omega$
RIN+	Input Resistance			100		k $\Omega$
CIN	Input Capacitance			0.9		pF
ROUT	Output Resistance			0.26		$\Omega$
VO	Output Voltage Range	No Load	$\pm 3.83$ <b><math>\pm 3.80</math></b>	$\pm 3.9$		V
VOL		$R_L = 100\Omega$	$\pm 3.52$ <b><math>\pm 3.5</math></b>	$\pm 3.53$		V
CMIR	Input Voltage Range		$\pm 2$	$\pm 2.5$		V
IO	Linear Output Current <sup>(3)(4)</sup>	$V_{IN} = 0V$	+70 40	$\pm 80$		mA
ISC	Short Circuit Current <sup>(5)</sup>	$V_{IN} = \pm 2V$ , Output Shorted to Ground		$\pm 230$		mA
XTLK	Channel to Channel Crosstalk	$V_{IN} = 2 V_{PP}$ @ 5 MHz		-90		dBc
XTLK	Channel to Channel Crosstalk	$V_{IN} = 2 V_{PP}$ @ 100 MHz		-54		dBc
XTLK	All Hostile Crosstalk	In A, C. Out B, $V_{IN} = 2 V_{PP}$ @ 5 MHz		-95		dBc

(5) The maximum output current ( $I_{OUT}$ ) is determined by the device power dissipation limitations. See the [Power Dissipation](#) section of the Application Section for more details. A short circuit condition should be limited to 5 seconds or less.

### ±3.3V Electrical Characteristics

Unless otherwise specified,  $V_S = \pm 3.3V$ ,  $R_L = 100\Omega$ .

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Typ	Max	Units
<b>Frequency Domain Performance</b>						
SSBW	–3 dB Bandwidth	$V_{OUT} = 0.25 V_{PP}$		360		MHz
LSBW	–3 dB Bandwidth	$V_{OUT} = 2.0 V_{PP}$		270		MHz
0.1 dBBW	0.1 dB Bandwidth	$V_{OUT} = 0.5 V_{PP}$		80		MHz
GFP	Peaking	DC to 200 MHz		0.3		dB
DG	Differential Gain	$R_L = 150\Omega$ , $f=4.43$ MHz		0.02		%
DP	Differential Phase	$R_L = 150\Omega$ , $f=4.43$ MHz		0.03		deg
<b>Time Domain Response</b>						
TRL	Rise and Fall Time	2V Step		2.0		ns
TSS	Settling Time to 0.05%	2V Step		15		ns
OS	Overshoot	2V Step		5		%
SR	Slew Rate	2V Step		1000		V/ $\mu$ s
<b>Distortion</b>						
HD2	2 <sup>nd</sup> Harmonic Distortion	2 $V_{PP}$ , 10 MHz		–70		dBc
HD3	3 <sup>rd</sup> Harmonic Distortion	2 $V_{PP}$ , 10 MHz		–74		dBc
IMD	3 <sup>rd</sup> Order Intermodulation Products	10 MHz, Two tones 2 $V_{PP}$ at Output		–79		dBc
<b>Static, DC Performance</b>						
GAIN	Voltage Gain			2.0		V/V
VIO	Output Offset Voltage	$V_{IN} = 0V$		1		mV
DVIO	Average Drift			36		$\mu$ V/ $^{\circ}$ C
IBN	Input Bias Current <sup>(2)</sup>	$V_{IN} = 0V$		2		$\mu$ A
DIBN	Average Drift			24		nA/ $^{\circ}$ C
PSRR	Power Supply Rejection Ratio	DC, Input Referred		54		dB
ICC	Supply Current	$R_L = \infty$		20		mA
VIH	Logic High Threshold	Select & Enable Pins			1.3	V
VIL	Logic Low Threshold	Select & Enable Pins	0.4			V
<b>Miscellaneous Performance</b>						
RIN+	Input Resistance			100		k $\Omega$
CIN	Input Capacitance			0.9		pF
ROUT	Output Resistance			0.27		$\Omega$
VO	Output Voltage Range	No Load		$\pm 2.5$		V
VOL		$R_L = 100\Omega$		$\pm 2.2$		V
CMIR	Input Voltage Range			$\pm 1.2$		V
IO	Linear Output Current	$V_{IN} = 0V$		$\pm 60$		mA
ISC	Short Circuit Current	$V_{IN} = \pm 1V$ , Output Shorted to Ground		$\pm 150$		mA
XTLK	Channel to Channel Crosstalk	5 MHz		–90		dBc

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that  $T_J = T_A$ . No specification of parametric performance is indicated in the electrical tables under conditions of internal self heating where  $T_J > T_A$ . See Applications Section for information on temperature de-rating of this device. Min/Max ratings are based on product testing, characterization and simulation. Individual parameters are tested as noted.
- (2) Positive Value is current into device.

## Typical Performance Characteristics

Unless otherwise specified,  $V_S = \pm 5V$ ,  $R_L = 100\Omega$ .

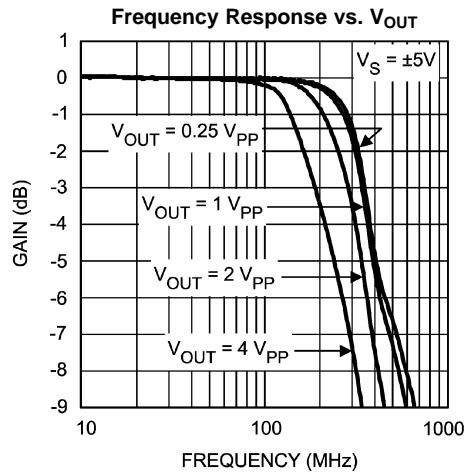


Figure 2.

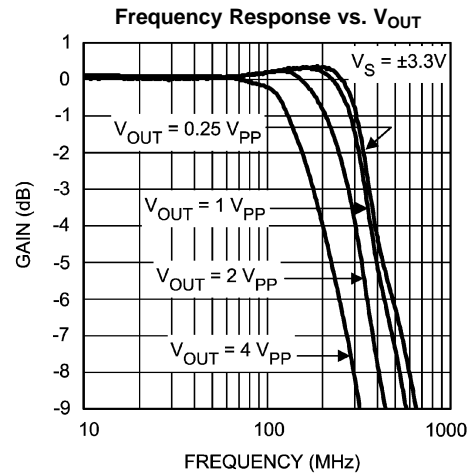


Figure 3.

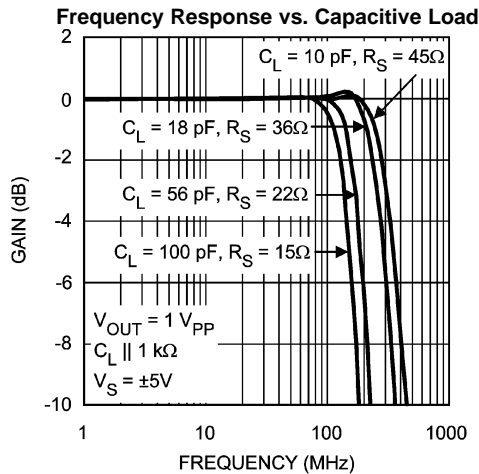


Figure 4.

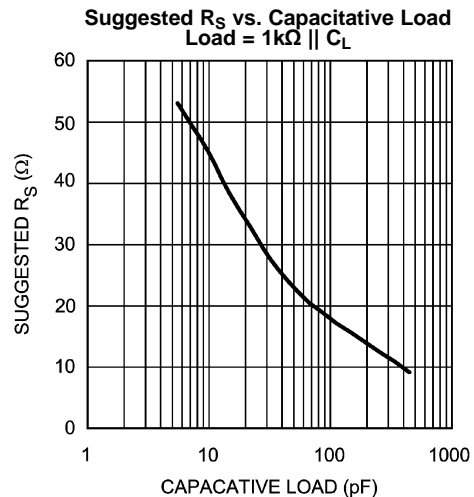


Figure 5.

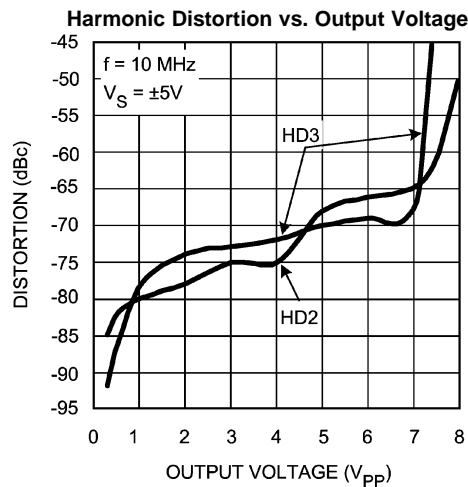


Figure 6.

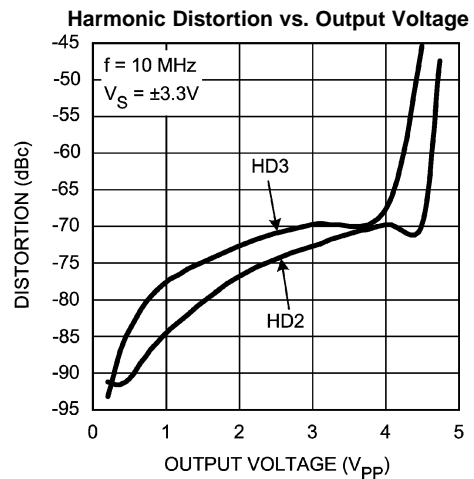
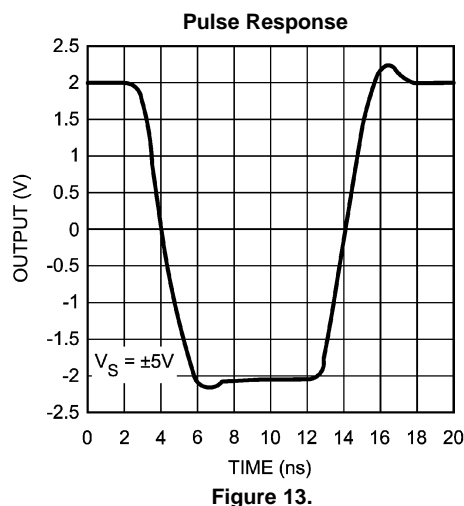
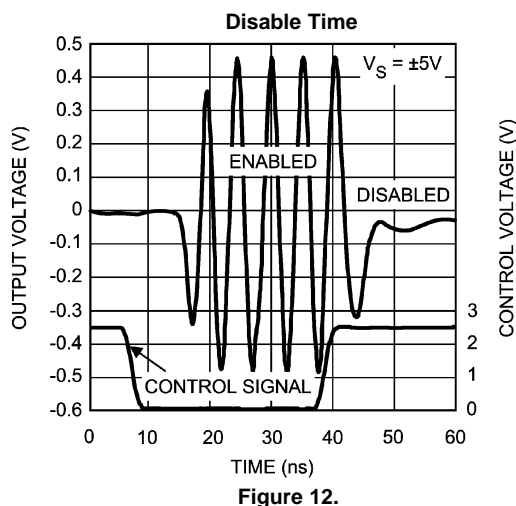
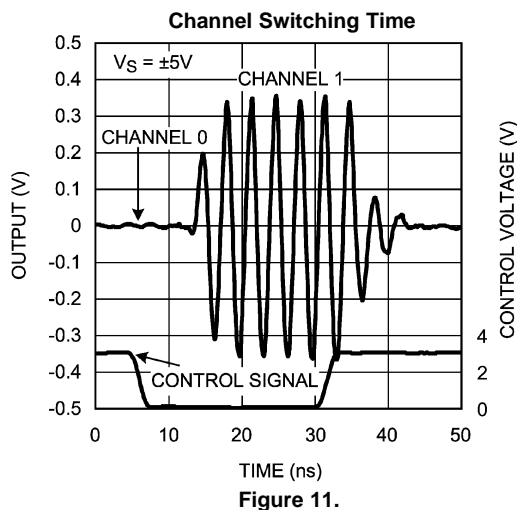
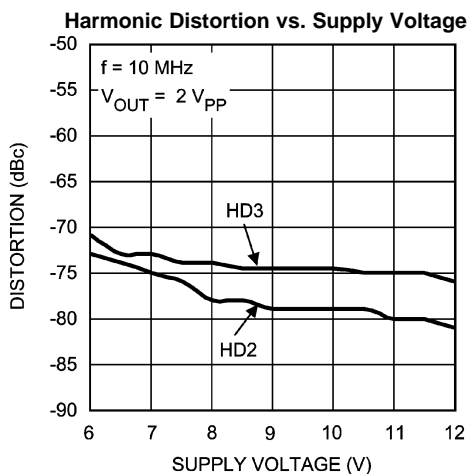
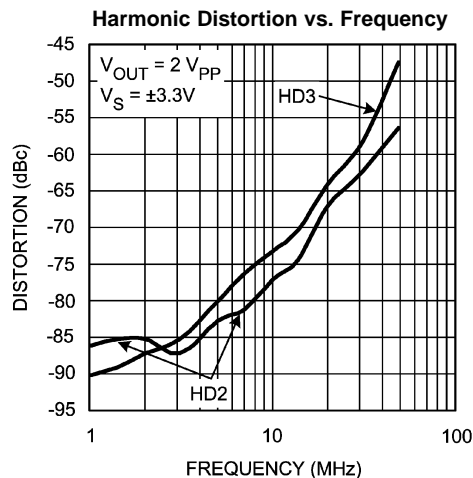
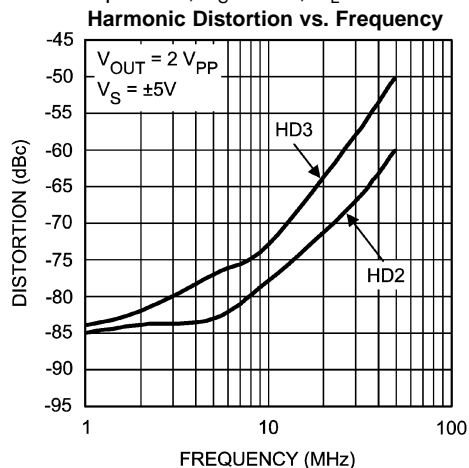


Figure 7.

## Typical Performance Characteristics (continued)

Unless otherwise specified,  $V_S = \pm 5V$ ,  $R_L = 100\Omega$ .



## Typical Performance Characteristics (continued)

Unless otherwise specified,  $V_S = \pm 5V$ ,  $R_L = 100\Omega$ .

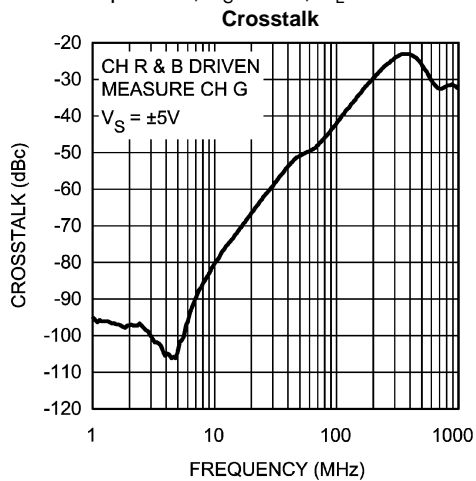


Figure 13.

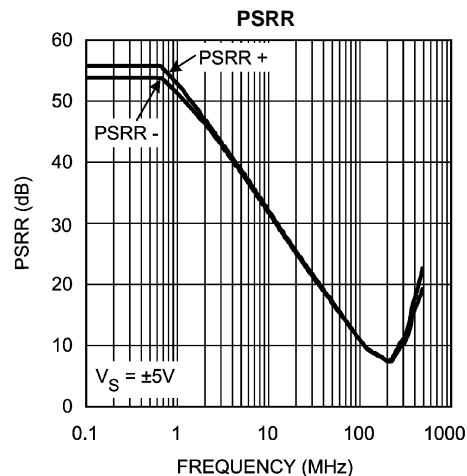


Figure 14.

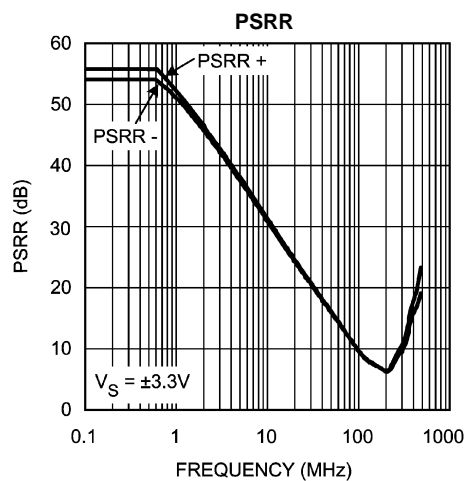


Figure 15.

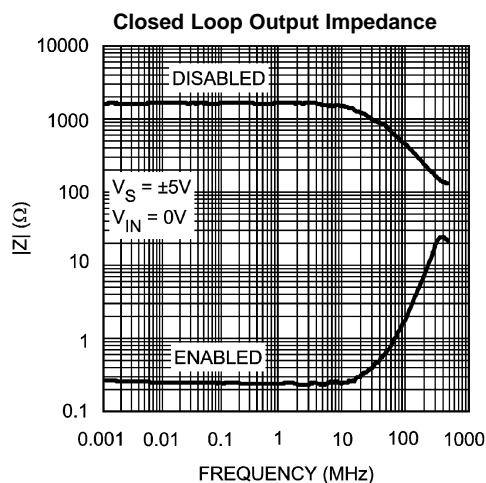


Figure 16.

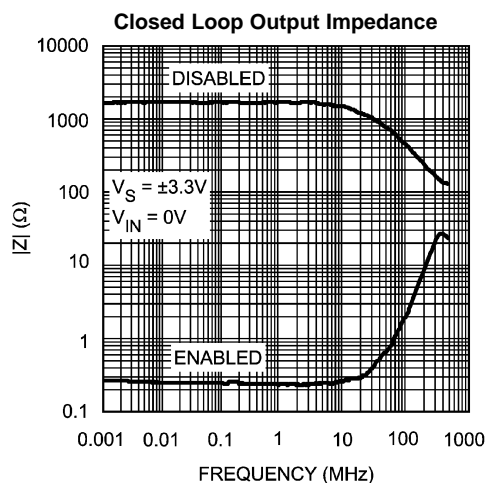


Figure 17.



## APPLICATION NOTES

### General Information

The LMH6572 is a high-speed triple 2:1 analog multiplexer, optimized for very high speed and low distortion. With a fixed gain of 2 and excellent AC performance, the LMH6572 is ideally suited for switching high resolution, presentation grade video signals. The LMH6572 has no internal ground reference. Single or split supply configurations are both possible. The LMH6572 features very high speed channel switching and disable times. When disabled the LMH6572 output is high impedance, making multiplexer expansion possible by combining multiple devices.

### Single Supply Operation

The LMH6572 uses mid-supply referenced circuits for the select and disable pins. In order to use the LMH6572 in single supply configuration, it is necessary to use a circuit similar to Figure 19. In this configuration the logical inputs are compatible with high breakdown open collector TTL, or open drain CMOS logic. In addition, the default logic state is reversed since there is a pull-up resistor on those pins. Single supply operation also requires the input to be biased to within the common mode input range of roughly  $\pm 2V$  from the mid-supply point.

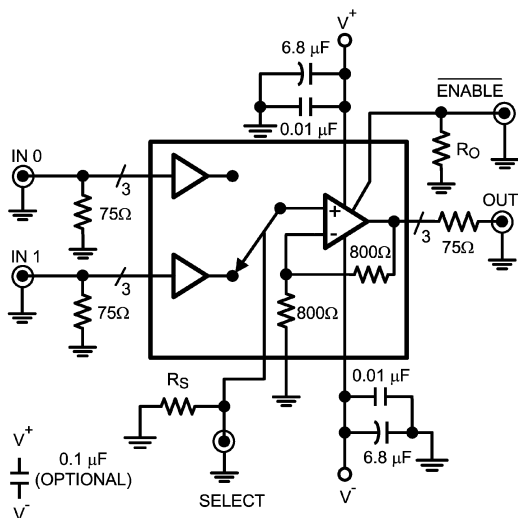


Figure 18. Typical Application

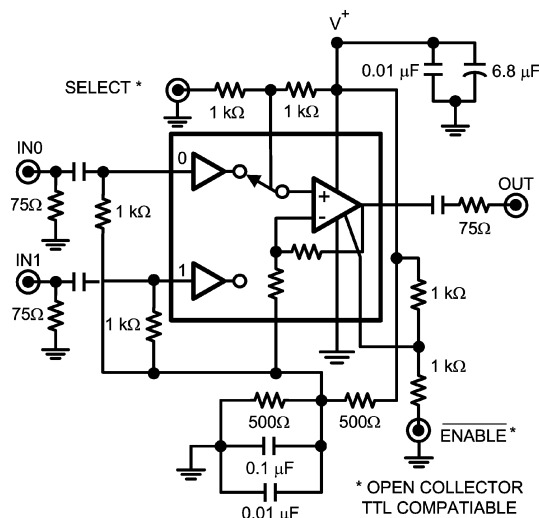


Figure 19. Single Supply Application

### Video Performance

The LMH6572 has been designed to provide excellent performance with production quality video signals in a wide variety of formats such as HDTV and High Resolution VGA. Best performance will be obtained with back-terminated loads. The back termination reduces reflections from the transmission line and effectively masks transmission line and other parasitic capacitances from the amplifier output stage. Figure 18 shows a typical configuration for driving a 75Ω cable. The output buffer is configured for a gain of 2, so using back terminated loads will give a net gain of 1.

### Gain Accuracy

The gain accuracy of the LMH6572 is accurate to  $\pm 0.5\%$  (0.3% typical) and stable over temperature. The internal gain setting resistors, RF and RG, match very well; however, over process and temperature their absolute value will change.

## Expanding the Multiplexer

It is possible to build higher density multiplexers by paralleling several LMH6572s. Figure 20 shows a 4:1 RGB MUX using two LMH6572s:

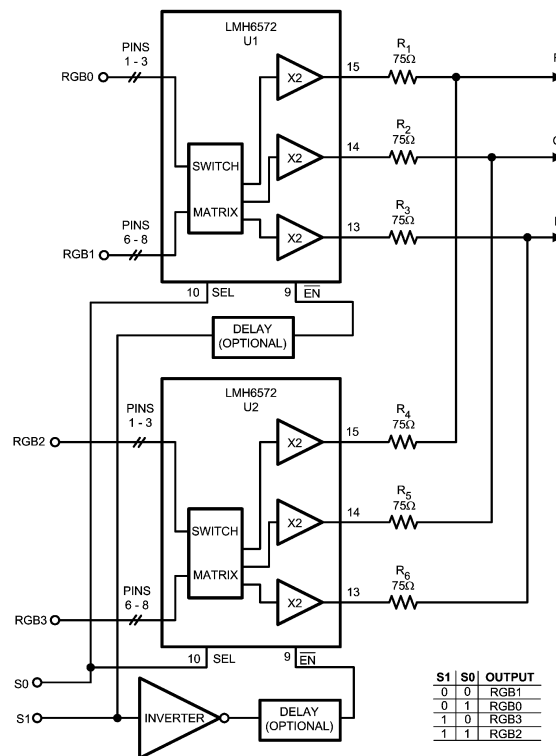


Figure 20. RGB MUX Using Two LMH6572's

If it is important in the end application to make sure that no two inputs are presented to the output at the same time, an optional delay block can be added prior to the ENABLE(EN) pin of each device, as shown. Figure 21 shows one possible approach to this delay circuit. The delay circuit shown will delay ENABLE's H to L transitions ( $R_1$  and  $C_1$  decay) but will not delay its L to H transition.

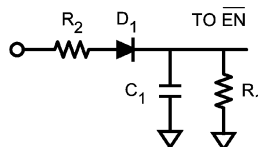


Figure 21. Delay Circuit Implementation

$R_2$  should be kept small compared to  $R_1$  in order to not reduce the ENABLE voltage and to produce little or no delay to the ENABLE L to H transition.

With the ENABLE pin putting the output stage into a high impedance state, several LMH6572's can be tied together to form a larger input MUX. However, there is a slight loading effect on the active output caused by the off-channel feedback and gain set resistors, as shown in Figure 21. Figure 22 is assuming there are four LMH6572 devices tied together to form a triple 8:1 MUX. With the internal resistors valued at approximately 800Ω, the gain error is about -0.57 dB, or about -6%.

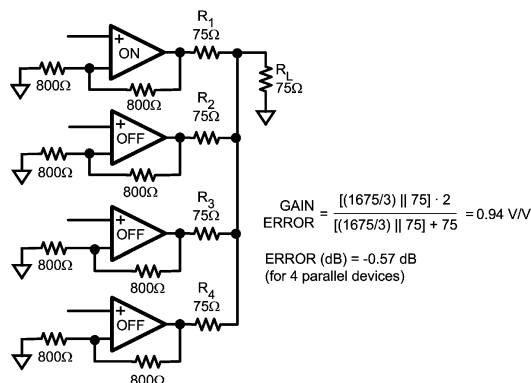


Figure 22. Multiplexer Input Expansion by Combining Outputs

An alternate approach would be to tie the outputs directly together and let all devices share a common back termination resistor in order to alleviate the gain error issue above.

The drawback in this case is the increased capacitive load presented to the output of each LMH6572 due to the offstate capacitance of the LMH6572.

## Other Applications

The LMH6572 may be utilized in systems that involve a single RGB channel as well whenever there is a need to switch between different “flavors” of a single RGB input.

Here are some examples:

1. RGB positive polarity, negative polarity switch
2. RGB full resolution, high-pass filter switch

In each of these applications, the same RGB input occupies one set of inputs to the LMH6572 and the other “flavor” would be tied to the other input set.

## Driving Capacitive Loads

Capacitive output loading applications will benefit from the use of a series output resistor,  $R_{OUT}$ , to stabilize the amplifier output under capacitive loading. Figure 23 shows the use of a series output resistor,  $R_{OUT}$ , to stabilize the amplifier output under capacitive loading. Capacitive loads of 5 to 120 pF are the most critical, causing ringing, frequency response peaking and possible oscillation. Figure 24 gives a recommended value for selecting a series output resistor for mitigating capacitive loads. The values suggested in the charts are selected for .5 dB or less of peaking in the frequency response. This gives a good compromise between settling time and bandwidth. For applications where maximum frequency response is needed and some peaking is tolerable, the value of  $R_{OUT}$  can be reduced slightly from the recommended values.

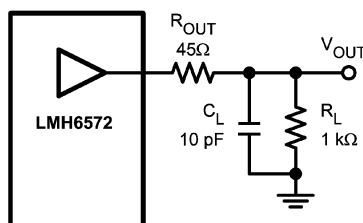


Figure 23. Decoupling Capacitive Loads

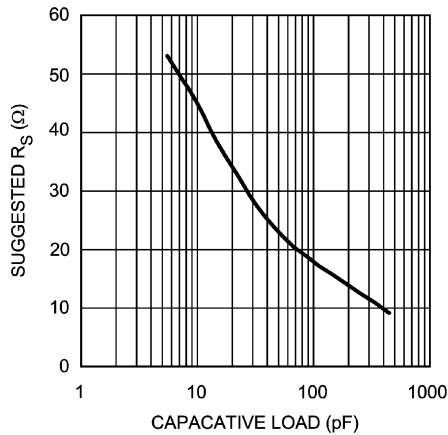
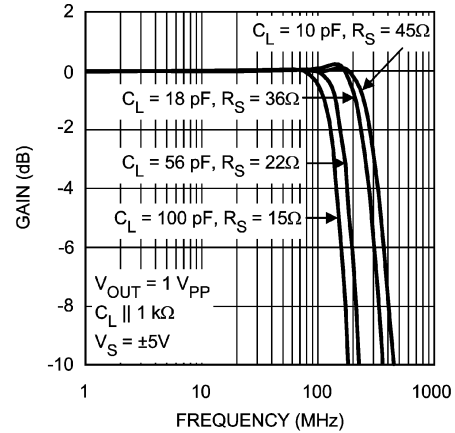
Figure 24. Recommended  $R_{OUT}$  vs. Capacitive Load

Figure 25. Frequency Response vs. Capacitive Load

### Layout Considerations

Whenever questions about layout arise, use the LMH730151 evaluation board as a guide. To reduce parasitic capacitances, ground and power planes should be removed near the input and output pins. For long signal paths controlled impedance lines should be used, along with impedance matching elements at both ends. Bypass capacitors should be placed as close to the device as possible. Bypass capacitors from each rail to ground are applied in pairs. The larger electrolytic bypass capacitors can be located farther from the device; however, the smaller ceramic capacitors should be placed as close to the device as possible. In Figure 18 and Figure 19, the capacitor between  $V^+$  and  $V^-$  is optional, but is recommended for best second harmonic distortion. Another way to enhance performance is to use pairs of .01  $\mu$ F and 0.1  $\mu$ F ceramic capacitors for each supply bypass.

### Power Dissipation

The LMH6572 is optimized for maximum speed and performance in the small form factor of the standard SSOP package. To achieve its high level of performance, the LMH6572 consumes 23 mA of quiescent current, which cannot be neglected when considering the total package power dissipation limit. To ensure maximum output drive and highest performance, thermal shutdown is not provided. Therefore, it is of utmost importance to make sure that the  $T_{JMAX}$  is never exceeded due to the overall power dissipation.

Follow these steps to determine the Maximum power dissipation for the LMH6572:

1. Calculate the quiescent (no-load) power:

$$P_{AMP} = I_{CC} * (V_S)$$

where

- $V_S = V^+ - V^-$  (1)

2. Calculate the RMS power dissipated in the output stage:

$$P_D (rms) = rms ((V_S - V_{OUT}) * I_{OUT})$$

where

- $V_{OUT}$  and  $I_{OUT}$  are the voltage across and the current through the external load
- $V_S$  is the total supply voltage (2)

3. Calculate the total RMS power:

$$P_T = P_{AMP} + P_D \quad (3)$$

The maximum power that the LMH6572 package can dissipate at a given temperature can be derived with the following equation:

$$P_{MAX} = (150^\circ - T_{AMB}) / \theta_{JA}$$

where

- $T_{AMB}$  = Ambient temperature ( $^\circ\text{C}$ )
- $\theta_{JA}$  = Thermal resistance, from junction to ambient, for a given package ( $^\circ\text{C}/\text{W}$ )
- For the SSOP package  $\theta_{JA}$  is 125  $^\circ\text{C}/\text{W}$  (4)

## ESD Protection

The LMH6572 is protected against electrostatic discharge (ESD) on all pins. The LMH6572 will survive 2000V Human Body model and 200V Machine model events. Under normal operation the ESD diodes have no effect on circuit performance. There are occasions, however, when the ESD diodes will be evident. If the LMH6572 is driven by a large signal while the device is powered down the ESD diodes will conduct. The current that flows through the ESD diodes will either exit the chip through the supply pins or will flow through the device, hence it is possible to power up a chip with a large signal applied to the input pins. Shorting the power pins to each other will prevent the chip from being powered up through the input.

## Evaluation Boards

Texas Instruments provides the following evaluation boards as a guide for high frequency layout and as an aid in device testing and characterization. Many of the datasheet plots were measured with these boards.

Device	Package	Evaluation Board Part Number
LMH6572	SSOP	LMH730151

An evaluation board can be shipped when a device sample request is placed with Texas Instruments.

REVISION HISTORY

Changes from Revision E (April 2013) to Revision F	Page
• Changed layout of National Data Sheet to TI format .....	<a href="#">13</a>

PRODUCT PREVIEW

## PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LMH6572MQ/NOPB	ACTIVE	SSOP	DBQ	16	95	Green (RoHS & no Sb/Br)	CU 100% TIN PLATING	Level-1-260C-UNLIM	-40 to 85	LH65 72MQ	<a href="#">Samples</a>
LMH6572MQX/NOPB	ACTIVE	SSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU 100% TIN PLATING	Level-1-260C-UNLIM	-40 to 85	LH65 72MQ	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

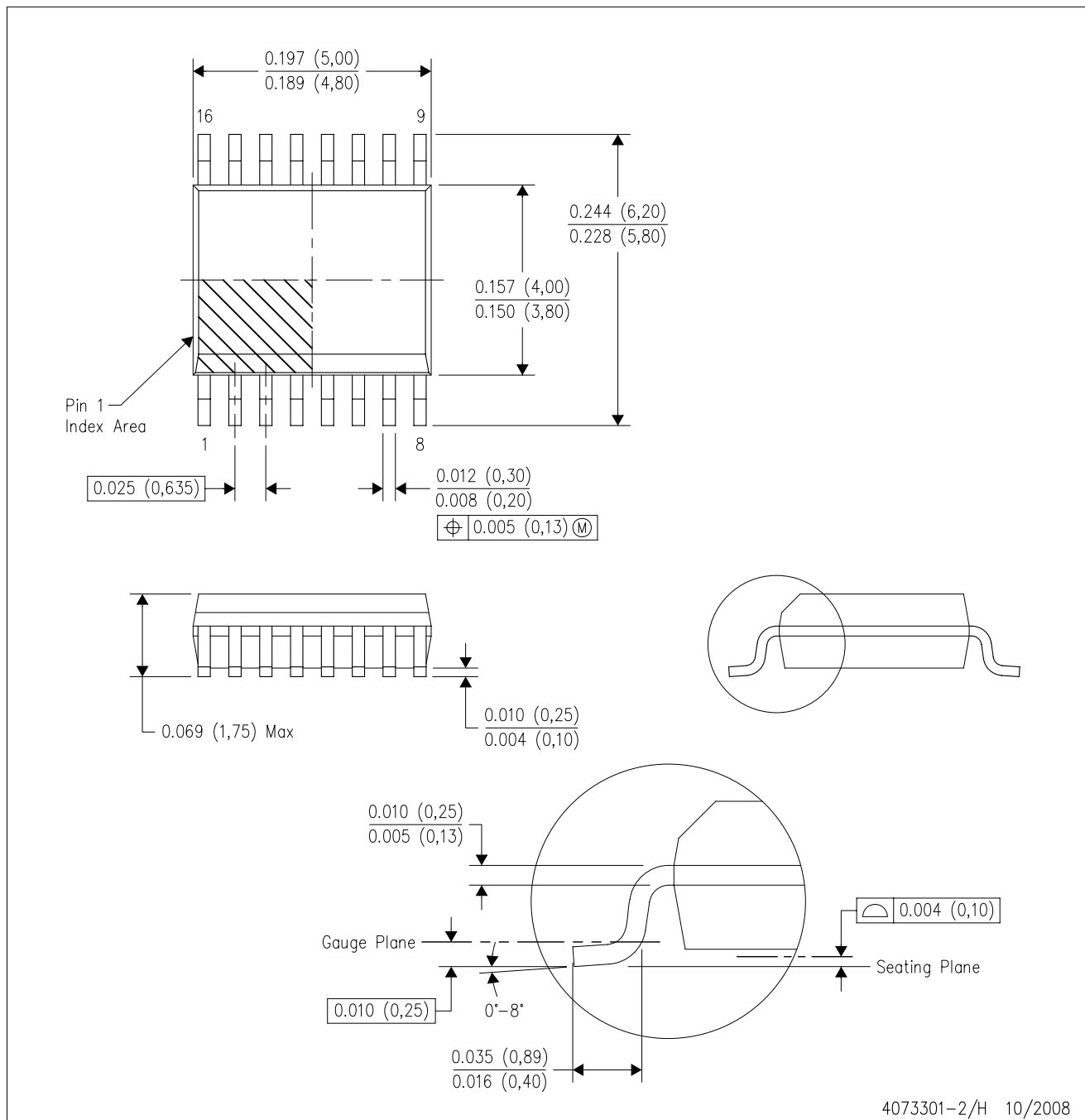
(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

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DBQ (R-PDSO-G16)

PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
  - B. This drawing is subject to change without notice.
  - C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15) per side.
  - D. Falls within JEDEC MO-137 variation AB.



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