

Single Channel 12-Bit 500Msps Analog to Digital Converter

Check for Samples: [ADS5403](#)

FEATURES

- Single Channel
- 12-Bit Resolution
- Maximum Clock Rate: 500 Msps
- Low Swing Fullscale Input: 1.0 Vpp
- Analog Input Buffer with High Impedance Input
- Input Bandwidth (3dB): >1.2GHz
- Data Output Interface: DDR LVDS
- 196-Pin BGA Package (12x12mm)
- Power Dissipation: 1 W
- Spectral Performance at $f_{IN} = 230$ MHz IF
 - SNR: 60.6 dBFS
 - SFDR: 80 dBc
- Spectral Performance at $f_{IN} = 700$ MHz IF
 - SNR: 59.5 dBFS
 - SFDR: 72 dBc

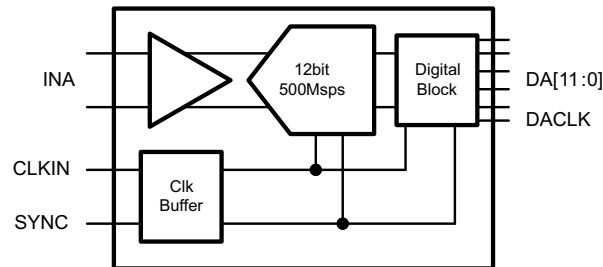
APPLICATIONS

- Test and Measurement Instrumentation
- Ultra-Wide Band Software Defined Radio
- Data Acquisition
- Power Amplifier Linearization
- Signal Intelligence and Jamming
- Radar and Satellite Systems
- Microwave Receivers

Device Part No.	Number of Channels	Speed Grade
ADS5402	2	800Msps
ADS5401	1	800Msps
ADS5404	2	500Msps
ADS5403	1	500Msps

DESCRIPTION

The ADS5403 is a high linearity single channel 12-bit, 500 MSPS analog-to-digital converter (ADC) easing front end filter design for wide bandwidth receivers. The analog input buffer isolates the internal switching of the on-chip track-and-hold from disturbing the signal source as well as providing a high-impedance input. Optionally the output data can be decimated by two. Designed for high SFDR, the ADC has low-noise performance and outstanding spurious-free dynamic range over a large input-frequency range. The device is available in a 196pin BGA package and is specified over the full industrial temperature range (–40°C to 85°C).



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

DETAILED BLOCK DIAGRAM

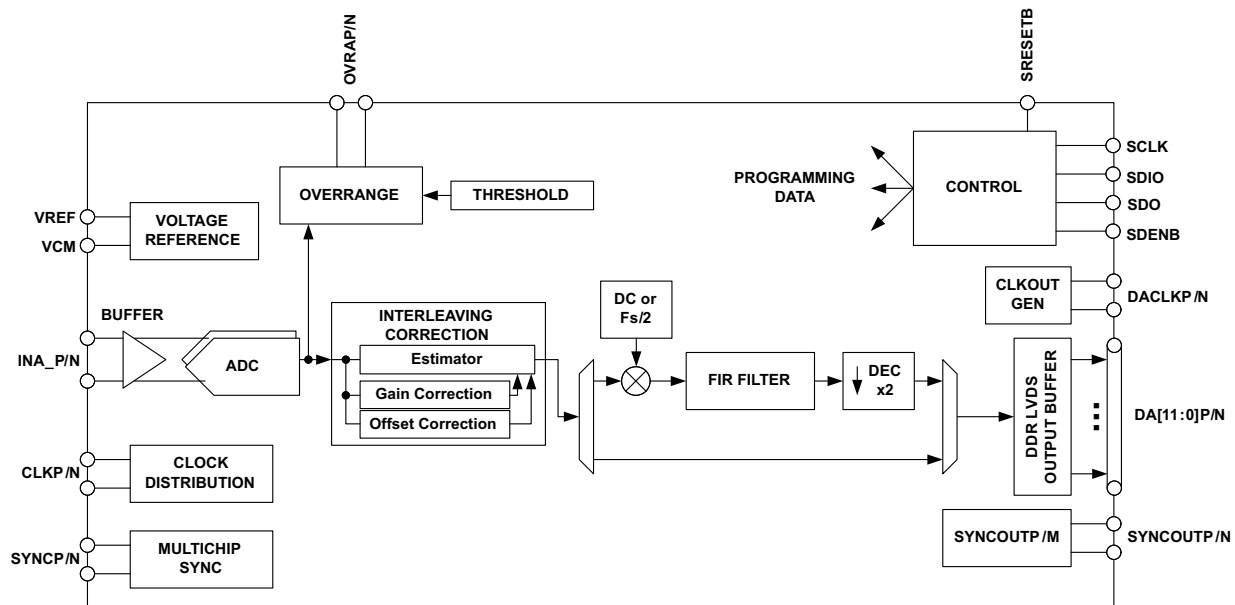


Figure 1. Detailed Block Diagram

PINOUT INFORMATION

	A	B	C	D	E	F	G	H	J	K	L	M	N	P	
14	VREF	VCM	GND	NC	NC	GND	AVDDC	AVDDC	GND	INA_P	INA_N	GND	GND	CLKINP	14
13	SDENB	TEST MODE	GND	GND	GND	GND	GND	GND	GND	GND	GND	GND	GND	CLKINN	13
12	SCLK	SRESET B	GND	AVDD33	AVDD33	AVDD33	AVDD33	AVDD33	AVDD33	AVDD33	AVDD33	GND	AVDD33	AVDD33	12
11	SDIO	ENABLE	GND	AVDD18	AVDD18	AVDD18	AVDD18	AVDD18	AVDD18	AVDD18	AVDD18	GND	AVDD18	AVDD18	11
10	SDO	IOVDD	GND	AVDD18	GND	GND	GND	GND	GND	GND	AVDD18	GND	NC	NC	10
9	DVDD	DVDD	GND	GND	GND	GND	GND	GND	GND	GND	GND	GND	SYNCP	SYNCP	9
8	DVDD	DVDD	DVDD	DVDD	GND	GND	GND	GND	GND	GND	DVDD	DVDD	DVDD	DVDD	8
7	NC	NC	DVDD LVDS	DVDD LVDS	GND	GND	GND	GND	GND	GND	DVDD LVDS	DVDD LVDS	NC	NC	7
6	NC	NC	DVDD LVDS	DVDD LVDS	GND	GND	GND	GND	GND	GND	DVDD LVDS	DVDD LVDS	NC	NC	6
5	NC	NC	NC	NC	GND	GND	GND	GND	GND	GND	OVRAN	OVRAP	SYNC OUTN	SYNC OUTP	5
4	NC	NC	NC	NC	NC	NC	NC	DA0P	DA2P	DA4P	DA6P	DA8P	NC	NC	4
3	NC	NC	NC	NC	NC	NC	NC	DA0N	DA2N	DA4N	DA6N	DA8N	DA11N	DA11P	3
2	NC	NC	NC	NC	NC	NC	NC	DA0P	DA1P	DA3P	DA5P	DA7P	DA10N	DA10P	2
1	NC	NC	NC	NC	NC	NC	NC	DA0N	DA1N	DA3N	DA5N	DA7N	DA9N	DA9P	1
	A	B	C	D	E	F	G	H	J	K	L	M	N	P	

Figure 2. Pinout in DDR output mode (top down view)

PIN ASSIGNMENTS

PIN		I/O	DESCRIPTION
NAME	NUMBER		
INPUT/REFERENCE			
INA_P/N	K14, L14	I	Analog ADC A differential input signal.
VCM	B14	O	Output of the analog input common mode (nominally 1.9V). A 0.1μF capacitor to AGND is recommended.
VREF	A14	O	Reference voltage output. A 0.1μF capacitor to AGND is recommended, but not required.
CLOCK/SYNC			
CLKINP/N	P14, P13	I	Differential input clock
SYNCP/N	P9, N9	I	Synchronization input. Inactive if logic low. When clocked in a high state initially, this is used for resetting internal clocks and digital logic and starting the SYNCOUT signal. Internal 100Ω termination.
CONTROL/SERIAL			
SRESET	B12	I	Serial interface reset input. Active low. Initialized internal registers during high to low transition. Asynchronous. Internal 50kΩ pull up resistor to IOVDD.
ENABLE	B11	I	Chip enable – active high. Power down function can be controlled through SPI register assignment. Internal 50kΩ pull up resistor to IOVDD.
SCLK	A12	I	Serial interface clock. Internal 50kΩ pull-down resistor.

PIN ASSIGNMENTS (continued)

PIN		I/O	DESCRIPTION
NAME	NUMBER		
SDIO	A11	I/O	Bi-directional serial data is 3 pin mode (default). In 4-pin interface mode (register x00, D16), the SDIO pin is an input only. Internal 50kΩ pull-down.
SDENB	A13	I	Serial interface enable. Internal 50kΩ pull-down resistor.
SDO	A10	O	Uni-directional serial interface data in 4 pin mode (register x00, D16). The SDO pin is tri-stated in 3-pin interface mode (default). Internal 50kΩ pull-down resistor.
TESTMODE	B13	–	Factory internal test, do not connect
DATA INTERFACE			
DA[11:0]P/N	P3, N3, P2, N2, P1, N1, M4, M3, M2, M1, L4, L3, L2, L1, K4, K3, K2, K1, J4, J3, J2, J1, H4, H3	O	ADC A Data Bits 11 (MSB) to 0 (LSB) in DDR output mode. Standard LVDS output.
DACLKP/N	H2, H1	O	DDR differential output data clock for Bus A. Register programmable to provide either rising or falling edge to center of stable data nominal timing.
SYNCOUP/N	F2, F1, P5, N5	O	Synchronization output signal for synchronizing multiple ADCs. Can be disabled via SPI.
OVRAP/N	M5, L5	O	Bus A, Overrange indicator, LVDS output. A logic high signals an analog input in excess of the full-scale range. Optional SYNC output.
NC	A1, A2, A3, A4, A5, A6, A7, B1, B2, B3, B4, B5, B6, B7, C1, C2, C3, C4, C5, D1, D2, D3, D4, D5, D14, E1, E2, E3, E4, E14, F3, F4, G1, G2, G3, G4, N4, N6, N7, N10, P4, P6, P7, P10	-	Don't connect to pin
POWER SUPPLY			
AVDD33	D12, E12, F12, G12, H12, J12, K12, L12, N12, P12	I	3.3V analog supply
AVDDC	G14, H14	I	1.8V supply for clock input
AVDD18	D10, D11, E11, F11, G11, H11, J11, K11, L10, L11, N11, P11	I	1.8V analog supply
DVDD	A8, A9, B8, B9, C8, D8, L8, M8, N8, P8	I	1.8V supply for digital block
DVDDLVDs	C6, C7, D6, D7, L6, L7, M6, M7	I	1.8V supply for LVDS outputs
IOVDD	B10	I	1.8V for digital I/Os
GND		I	Ground

PACKAGE/ORDERING INFORMATION

PRODUCT	PACKAGE-LEAD	PACKAGE DESIGNATOR	SPECIFIED TEMPERATURE RANGE	ECO PLAN ⁽²⁾	LEAD/BALL FINISH	PACKAGE MARKING	ORDERING NUMBER	TRANSPORT MEDIA, QUANTITY
ADS5403	196-BGA	ZAY	–40C to 85C	GREEN (RoHS & no Sb/Br)		ADS5403I	ADS5403IZAY	Tray
							ADS5403IZAYR	Tape and Reel

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)

		VALUE		UNIT
		MIN	MAX	
Supply voltage range, AVDD33		–0.5	4	V
Supply voltage range, AVDDC		–0.5	2.3	V
Supply voltage range, AVDD18		–0.5	2.3	V
Supply voltage range, DVDD		–0.5	2.3	V
Supply voltage range, DVDDLVD		–0.5	2.3	V
Supply voltage range, IOVDD		–0.5	4	V
Voltage applied to input pins	INA/B_P, INA/B_N	–0.5	AVDD33 + 0.5	V
	CLKINP, CLKINN	–0.5	AVDDC + 0.5	V
	SYNCP, SYNCN	–0.5	AVDD33 + 0.5	V
	SRESET, SDENB, SCLK, SDIO, SDO, ENABLE	–0.5	IOVDD + 0.5	V
Operating free-air temperature range, T _A		–40	85	°C
Operating junction temperature range, T _J			150	°C
Storage temperature range		–65	150	°C
ESD, Human Body Model			2	kV

THERMAL INFORMATION

THERMAL METRIC ⁽¹⁾		ADS5403	UNITS
		QFN	
		196 PINS	
θ_{JA}	Junction-to-ambient thermal resistance ⁽²⁾	37.6	°C/W
θ_{JCTop}	Junction-to-case (top) thermal resistance ⁽³⁾	6.8	
θ_{JB}	Junction-to-board thermal resistance ⁽⁴⁾	16.8	
ψ_{JT}	Junction-to-top characterization parameter ⁽⁵⁾	0.2	
ψ_{JB}	Junction-to-board characterization parameter ⁽⁶⁾	16.4	

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

(2) The junction-to-ambient thermal resistance under natural convection is obtained in a simulation on a JEDEC-standard, high-K board, as specified in JESD51-7, in an environment described in JESD51-2a.

(3) The junction-to-case (top) thermal resistance is obtained by simulating a cold plate test on the package top. No specific JEDEC-standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

(4) The junction-to-board thermal resistance is obtained by simulating in an environment with a ring cold plate fixture to control the PCB temperature, as described in JESD51-8.

(5) The junction-to-top characterization parameter, ψ_{JT} , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining θ_{JA} , using a procedure described in JESD51-2a (sections 6 and 7).

(6) The junction-to-board characterization parameter, ψ_{JB} , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining θ_{JA} , using a procedure described in JESD51-2a (sections 6 and 7).

RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range (unless otherwise noted)

RECOMMENDED OPERATING CONDITIONS (continued)

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
T_J	Recommended operating junction temperature			105	°C
	Maximum rated operating junction temperature ⁽¹⁾	125			
T_A	Recommended free-air temperature	–40	25	85	°C

(1) Prolonged use at this junction temperature may increase the device failure-in-time (FIT) rate.

ELECTRICAL CHARACTERISTICS

Typical values at $T_A = 25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = 85^\circ\text{C}$, ADC sampling rate = 500Msps, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVDs/IOVDD = 1.8V, –1dBFS differential input (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNITS
ADC Clock Frequency			40		500	MSPS
Resolution			12			Bits
SUPPLY						
AVDD33			3.15	3.3	3.45	V
AVDDC, AVDD18, DVDD, DVDDLVS			1.7	1.8	1.9	V
IOVDD			1.7	1.8	3.45	V
POWER SUPPLY						
I _{AVDD33}	3.3V Analog supply current			150	170	mA
I _{AVDD18}	1.8V Analog supply current			55	65	mA
I _{AVDDC}	1.8V Clock supply current			27	35	mA
I _{DVDD}	1.8V Digital supply current	Auto Correction Disabled		87	140	mA
I _{DVDD}	1.8V Digital supply current	Auto Correction Enabled		147	170	mA
I _{DVDD}	1.8V Digital supply current	Auto Correction Enabled, decimation filter enabled		164		mA
I _{DVDDLVS}	1.8V LVDS supply current			73	90	mA
I _{IOVDD}	1.8V I/O Voltage supply current			1	2	mA
P _{dis}	Total power dissipation	Auto Correction Enabled, decimation filter disabled		1.04	1.2	W
PSRR		250kHz to 500MHz	40			dB
Shut-down power dissipation				7		mW
Shut-down wake up time				2.5		ms
Standby power dissipation				7		mW
Standby wake up time				100		μs
Deep-sleep mode power dissipation	Auto correction disabled			220		mW
	Auto correction enabled			305		mW
Deep-sleep mode wakeup time				20		μs
Light-sleep mode power dissipation	Auto correction disabled			367		mW
	Auto correction enabled			448		mW
Light-sleep mode wakeup time				2		μs

ELECTRICAL CHARACTERISTICS

Typical values at $T_A = 25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = 85^\circ\text{C}$, ADC sampling rate = 500Msps, 50% clock duty cycle, AVDD3V = 3.3V, AVDD/DRVDD/IOVDD = 1.8V, –1dBFS differential input (unless otherwise noted).

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS
ANALOG INPUTS					
Differential input full-scale			1 1.25		V _{pp}
Input common mode voltage			1.9	±0.1	V
Input resistance	Differential at DC		1		kΩ
Input capacitance	Each input to GND		2		pF
VCM common mode voltage output			1.9		V
Analog input bandwidth (3dB)			1200		MHz
DYNAMIC ACCURACY					
Offset Error	Auto Correction Disabled	-20	-7.5	20	mV
	Auto Correction Enabled	-1	0	1	mV
Offset temperature coefficient			-611		μV/°C
Gain error		-5		5	%FS
Gain temperature coefficient			0.005		%FS/°C
Differential nonlinearity	$f_{\text{IN}} = 230 \text{ MHz}$	-1	±0.9	2	LSB
Integral nonlinearity	$f_{\text{IN}} = 230 \text{ MHz}$	-5	±1.5	5	LSB
CLOCK INPUT					
Input clock frequency		40		500	MHz
Input clock amplitude			2		V _{pp}
Input clock duty cycle		40	50	60	%
Internal clock biasing			0.9		V

ELECTRICAL CHARACTERISTICS

Typical values at $T_A = 25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = 85^\circ\text{C}$, ADC sampling rate = 500MSPS, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVD/IOVDD = 1.8V, –1dBFS differential input (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
Auto Correction			Enabled			Disabled			Vpp
DYNAMIC AC CHARACTERISTICS ⁽¹⁾ – Burst Mode Enabled: 12bit High Resolution Output Data									
SNR	Signal to Noise Ratio	f _{IN} = 10 MHz		60.8		60.8		dBFS	
		f _{IN} = 100 MHz		60.7		60.8			
		f _{IN} = 230 MHz	59	60.6		60.7			
		f _{IN} = 450 MHz		60.2		60.6			
		f _{IN} = 700 MHz		59.5		60.1			
HD2,3	Second and third harmonic distortion	f _{IN} = 10 MHz		83		84		dBc	
		f _{IN} = 100 MHz		85		83			
		f _{IN} = 230 MHz	70	82		82			
		f _{IN} = 450 MHz		81		82			
		f _{IN} = 700 MHz		76		74			
Non HD2,3	Spur Free Dynamic Range (excluding second and third harmonic distortion)	f _{IN} = 10 MHz		79		81		dBc	
		f _{IN} = 100 MHz		78		82			
		f _{IN} = 230 MHz	70	80		81			
		f _{IN} = 450 MHz		75		82			
		f _{IN} = 700 MHz		72		80			
IL	Fs/2-Fin interleaving spur	f _{IN} = 10 MHz		87		81		dBc	
		f _{IN} = 100 MHz		81		79			
		f _{IN} = 230 MHz	70	84		80			
		f _{IN} = 450 MHz		80		77			
		f _{IN} = 700 MHz		74		73			
SINAD	Signal to noise and distortion ratio	f _{IN} = 10 MHz		60.6		60.7		dBFS	
		f _{IN} = 100 MHz		60.6		60.7			
		f _{IN} = 230 MHz	57.5	60.6		60.7			
		f _{IN} = 450 MHz		60.1		60.5			
		f _{IN} = 700 MHz		59.3		59.9			
THD	Total Harmonic Distortion	f _{IN} = 10 MHz		76		78		dBc	
		f _{IN} = 100 MHz		78		78			
		f _{IN} = 230 MHz	68	79		77			
		f _{IN} = 450 MHz		76		77			
		f _{IN} = 700 MHz		74		73			
IMD3	Inter modulation distortion	F _{in} = 129.5 and 130.5MHz, -7dBFS		82		82		dBFS	
		F _{in} = 349.5 and 350.5MHz, -7dBFS		80		80			
Crosstalk				90		90		dB	
ENOB	Effective number of bits	f _{IN} = 230 MHz		9.8		9.8		Bit	

(1) SFDR and SNR calculations do not include the DC or Fs/2 bins when Auto Correction is disabled.

ELECTRICAL CHARACTERISTICS

Typical values at $T_A = 25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = 85^\circ\text{C}$, ADC sampling rate = 500Msps, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVS/IOVDD = 1.8V, –1dBFS differential input (unless otherwise noted).

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS
OVER-DRIVE RECOVERY ERROR					
Input overload recovery	Recovery to within 5% (of final value) for 6dB overload with sine wave input		2		ns
SAMPLE TIMING CHARACTERISTICS					
rms Aperture Jitter	Sample uncertainty		100		fs rms
Data Latency	ADC sample to digital output, auto correction disabled		38		Clock Cycles
	ADC sample to digital output, auto correction enabled		50		
	ADC sample to digital output, Decimation filter enabled, Auto correction disabled		74		Sampling clock Cycles
Over-range Latency	ADC sample to over-range output		12		Clock Cycles

ELECTRICAL CHARACTERISTICS

The DC specifications refer to the condition where the digital outputs are not switching, but are permanently at a valid logic level 0 or 1. AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVS/IOVDD = 1.8V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS	
DIGITAL INPUTS – SRESET, SCLK, SDENB, SDIO, ENABLE						
High-level input voltage	All digital inputs support 1.8V and 3.3V logic levels.	0.7 x IOVDD			V	
Low-level input voltage		0.3 x IOVDD			V	
High-level input current		–50		200	μA	
Low-level input current		–50		50	μA	
Input capacitance		5			pF	
DIGITAL OUTPUTS – SDO						
High-level output voltage	Iload = -100uA	IOVDD – 0.2			V	
	Iload = -2mA	0.8 x IOVDD				
Low-level output voltage	Iload = 100uA	0.2			V	
	Iload = 2mA	0.22 x IOVDD				
DIGITAL INPUTS – SYNCP/N, TRIGGERP/N						
V _{ID}	Differential input voltage	250	350	450	mV	
V _{CM}	Input common mode voltage	1.125	1.2	1.375	V	
t _{SU}		500			ps	
DIGITAL OUTPUTS – DA[11:0]P/N, DACLK/P/N, OVRAP/N, SYNCOUTP/N, TRDYP/N, HRESP/N, DB[11:0]P/N, DBCLKP/N, OVRBP/N						
V _{OD}	Output differential voltage	Iout = 3.5mA	250	350	450	mV
V _{OCM}	Output common mode voltage	Iout = 3.5mA	1.125	1.25	1.375	V
t _{suA}		F _s = 500Mpsps, Data valid to zero-crossing of DACLK	600	800		ps
t _{hA}		F _s = 500Mpsps, Zero-crossing of DACLK to data becoming invalid	600	790		ps
t _{PD}		F _s = 500Mpsps, CLKIN falling edge to DACLK, DBCLK rising edge	3.28	3.48	3.74	ns
t _{RISE}		10% - 90%	100	150	200	ps
t _{FALL}		90% - 10%	100	150	200	ps

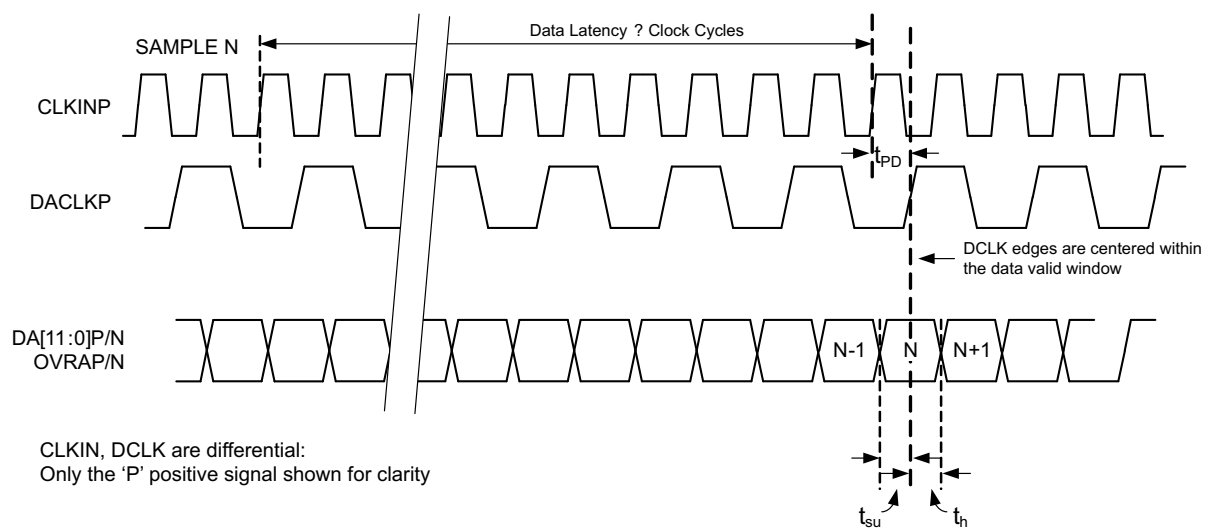


Figure 3. Timing Diagram for 12-bit DDR Output

TYPICAL CHARACTERISTICS

Typical values at $T_A = +25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = +85^\circ\text{C}$, ADC sampling rate = 500Mps, 50% clock duty cycle, $\text{AVDD33} = 3.3\text{V}$, $\text{AVDDC}/\text{AVDD18}/\text{DVDD}/\text{DVDDLVS}/\text{IOVDD} = 1.8\text{V}$, -1dBFS differential input, unless otherwise noted.

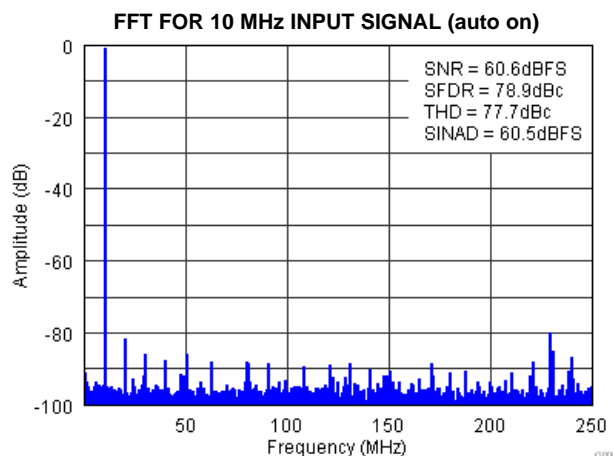


Figure 4.

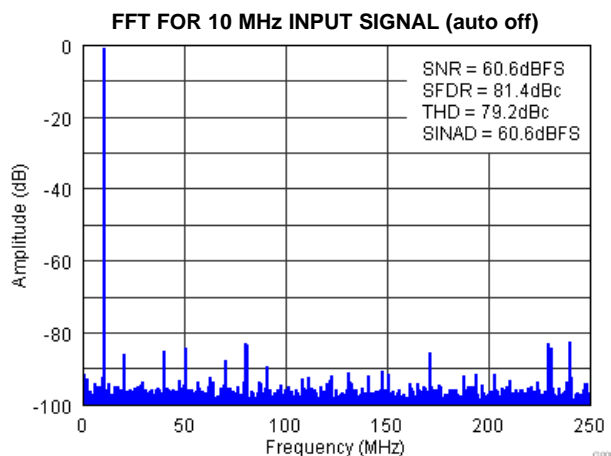


Figure 5.

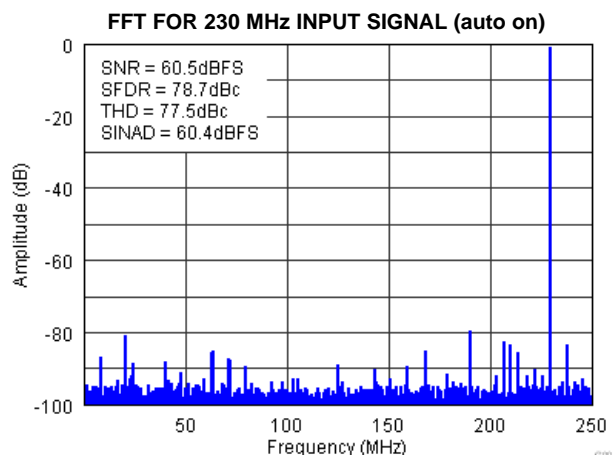


Figure 6.

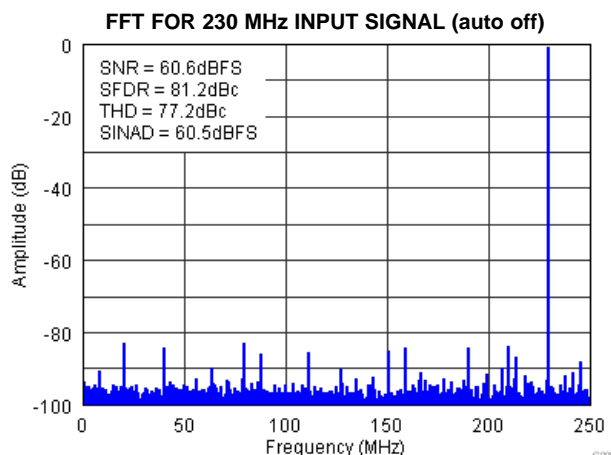


Figure 7.

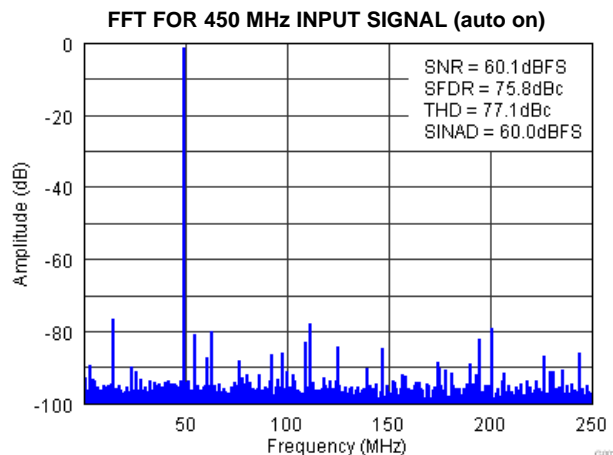


Figure 8.

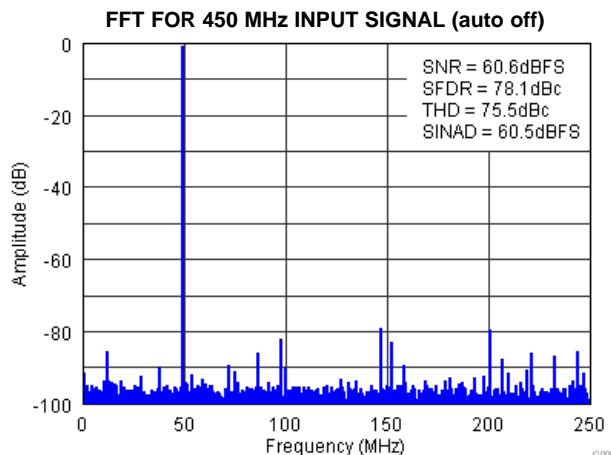


Figure 9.

TYPICAL CHARACTERISTICS (continued)

Typical values at $T_A = +25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = +85^\circ\text{C}$, ADC sampling rate = 500Msps, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVS/IOVDD = 1.8V, -1dBFS differential input, unless otherwise noted.

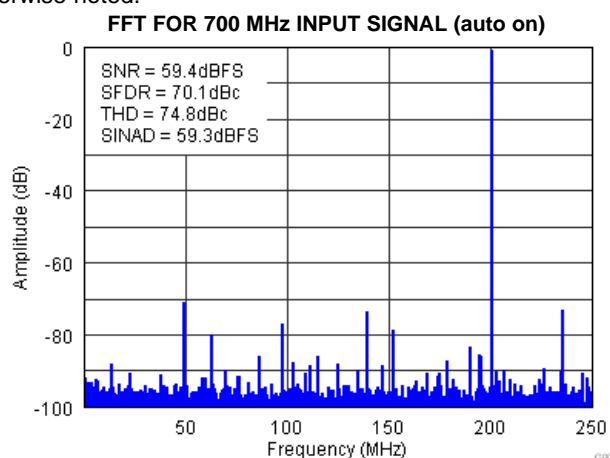


Figure 10.

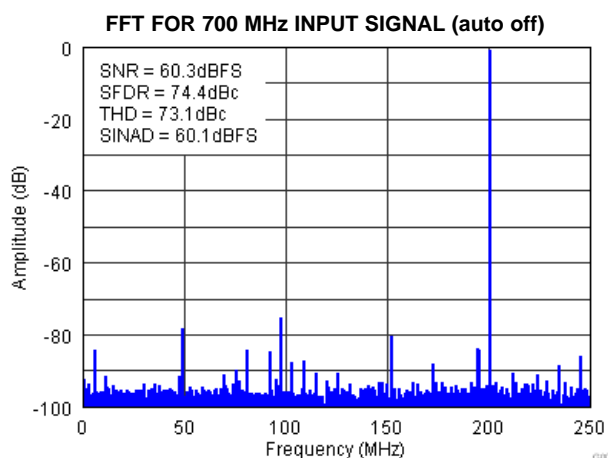


Figure 11.

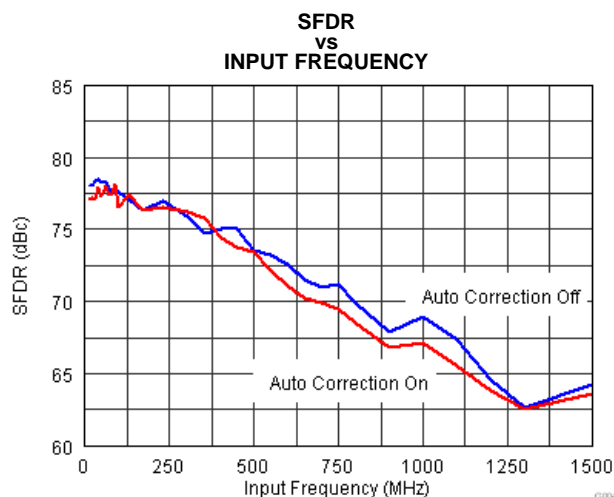


Figure 12.

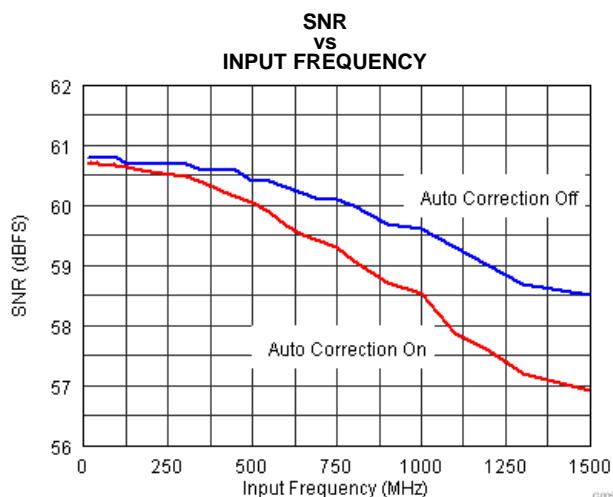


Figure 13.

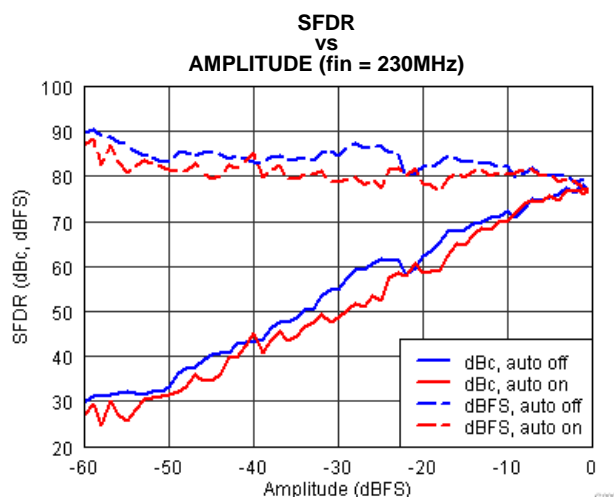


Figure 14.

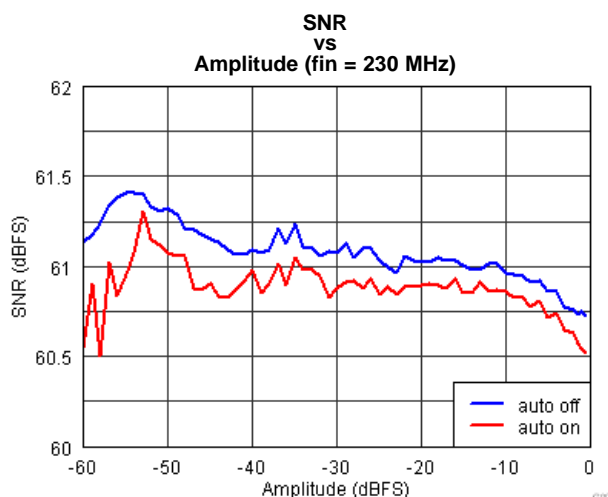


Figure 15.

TYPICAL CHARACTERISTICS (continued)

Typical values at $T_A = +25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = +85^\circ\text{C}$, ADC sampling rate = 500Mps, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVS/IOVDD = 1.8V, -1dBFS differential input, unless otherwise noted.

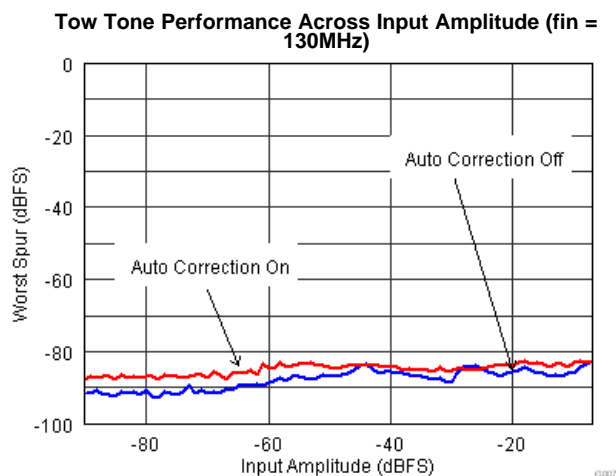


Figure 16.

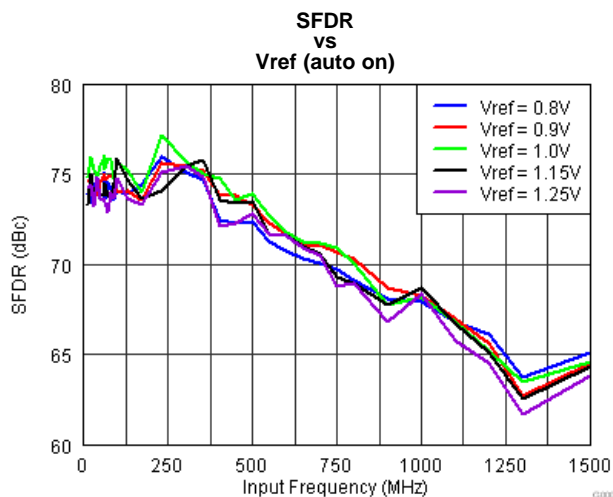


Figure 17.

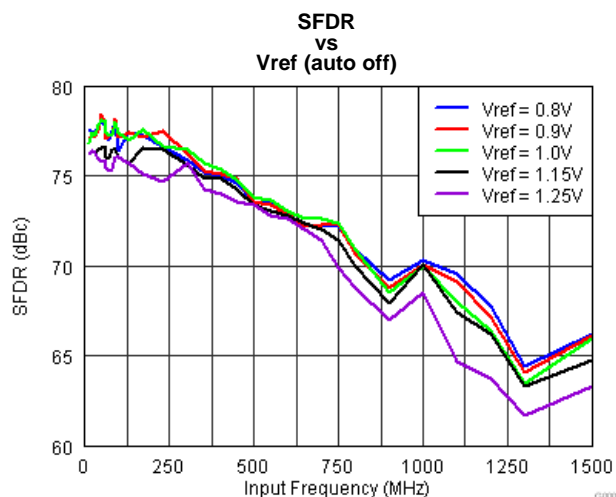


Figure 18.

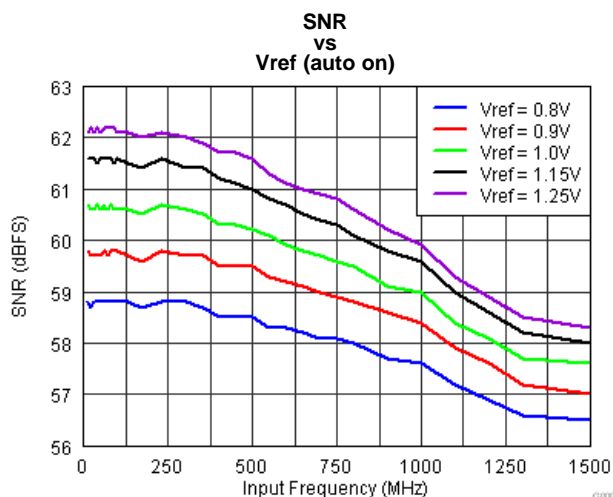


Figure 19.

TYPICAL CHARACTERISTICS (continued)

Typical values at $T_A = +25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = +85^\circ\text{C}$, ADC sampling rate = 500Msps, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVS/IOVDD = 1.8V, -1dBFS differential input, unless otherwise noted.

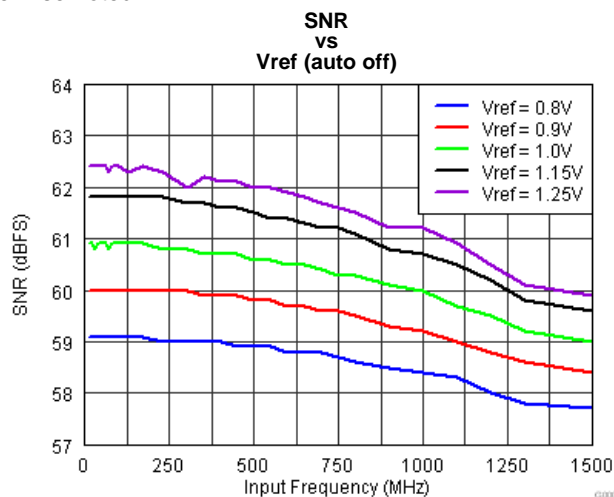


Figure 20.

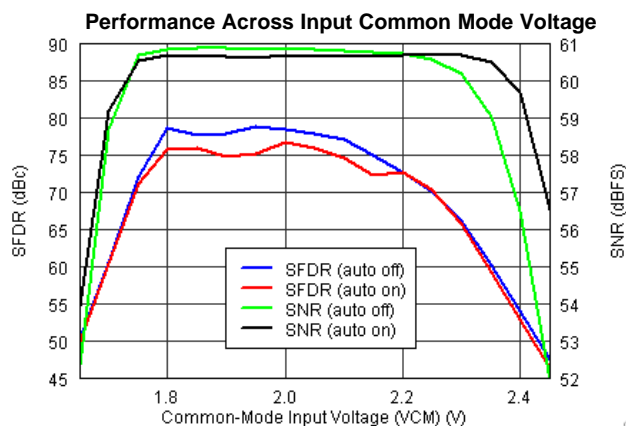


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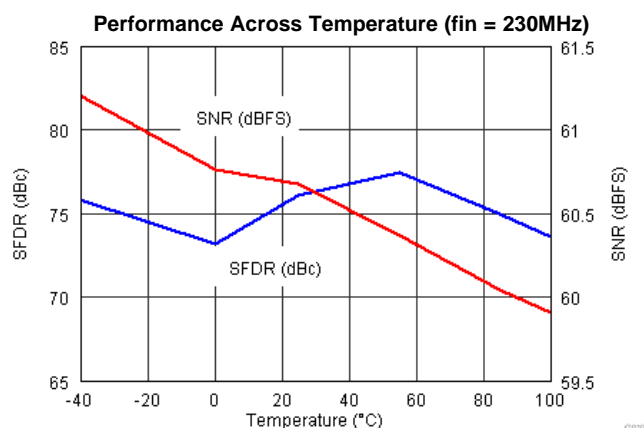


Figure 22.

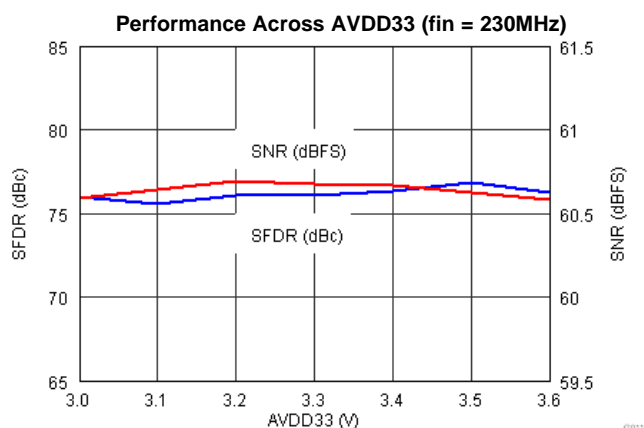


Figure 23.

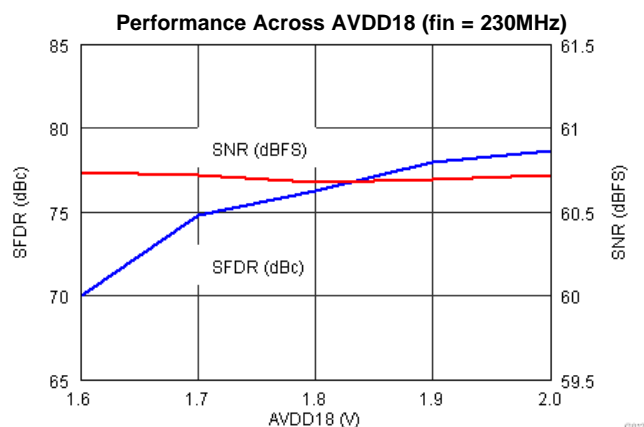


Figure 24.

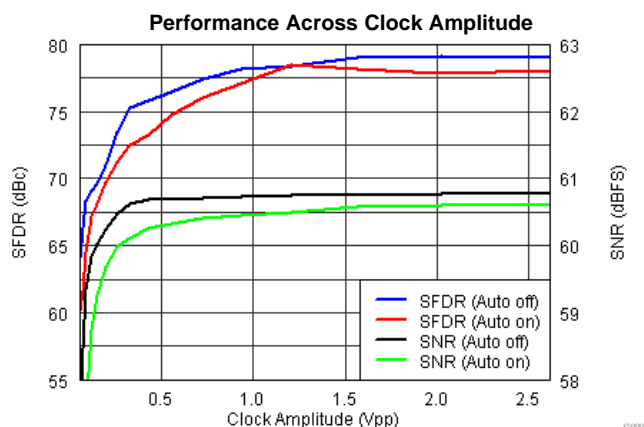


Figure 25.

TYPICAL CHARACTERISTICS (continued)

Typical values at $T_A = +25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = +85^\circ\text{C}$, ADC sampling rate = 500Mps, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVS/IOVDD = 1.8V, -1dBFS differential input, unless otherwise noted.

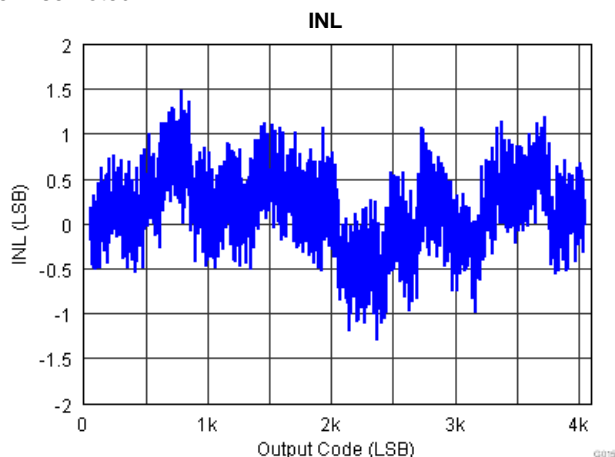


Figure 26.

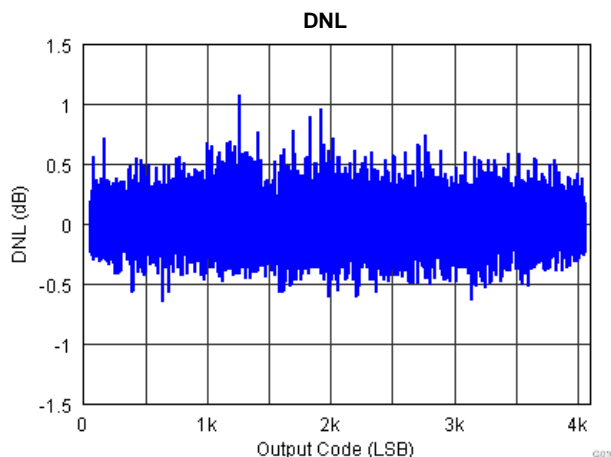


Figure 27.

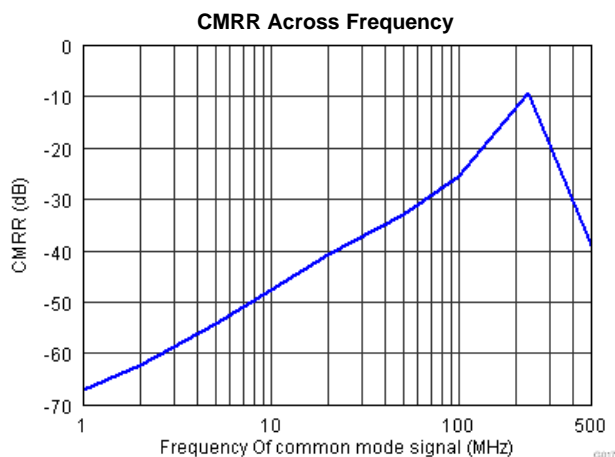


Figure 28.

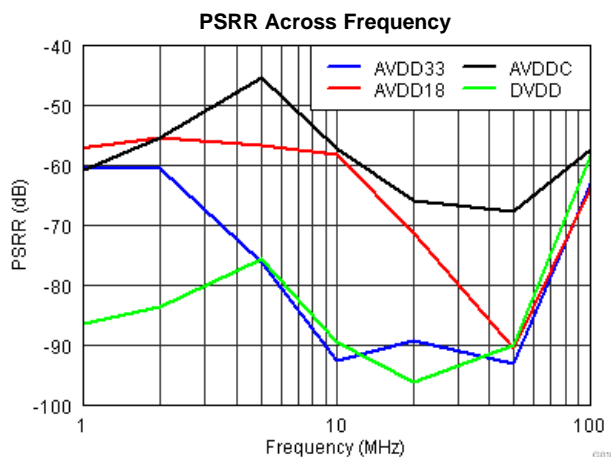


Figure 29.

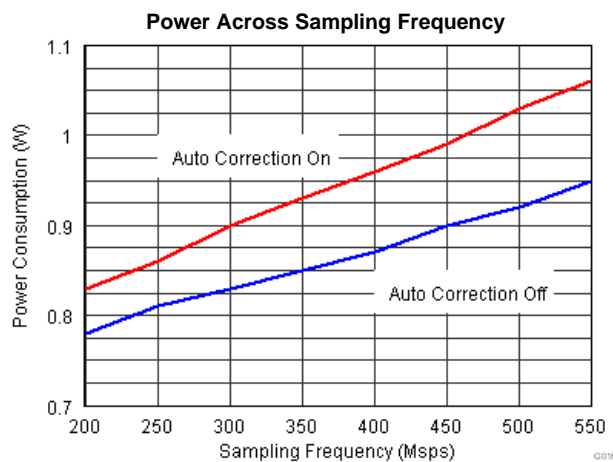


Figure 30.

TYPICAL CHARACTERISTICS (continued)

Typical values at $T_A = +25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = +85^\circ\text{C}$, ADC sampling rate = 500MSPS, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVS/IOVDD = 1.8V, -1dBFS differential input, unless otherwise noted.

SFDR Across Input and Sampling Frequencies (auto on)

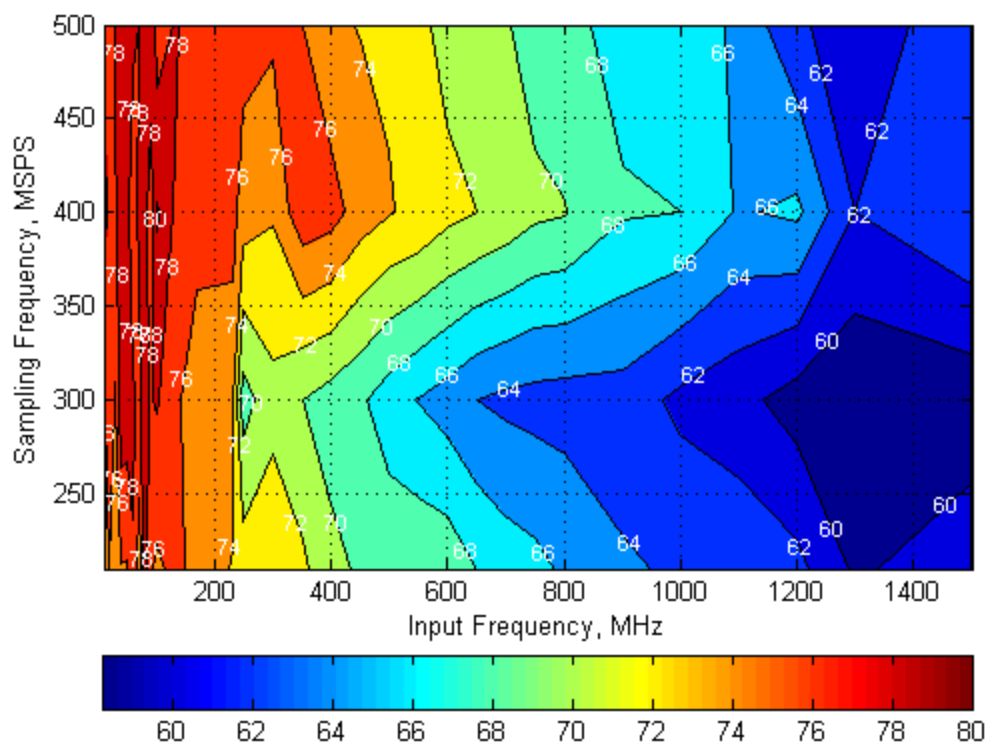


Figure 31.

TYPICAL CHARACTERISTICS (continued)

Typical values at $T_A = +25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = +85^\circ\text{C}$, ADC sampling rate = 500MSPS, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVS/IOVDD = 1.8V, -1dBFS differential input, unless otherwise noted.

SFDR Across Input and Sampling Frequencies (auto off)

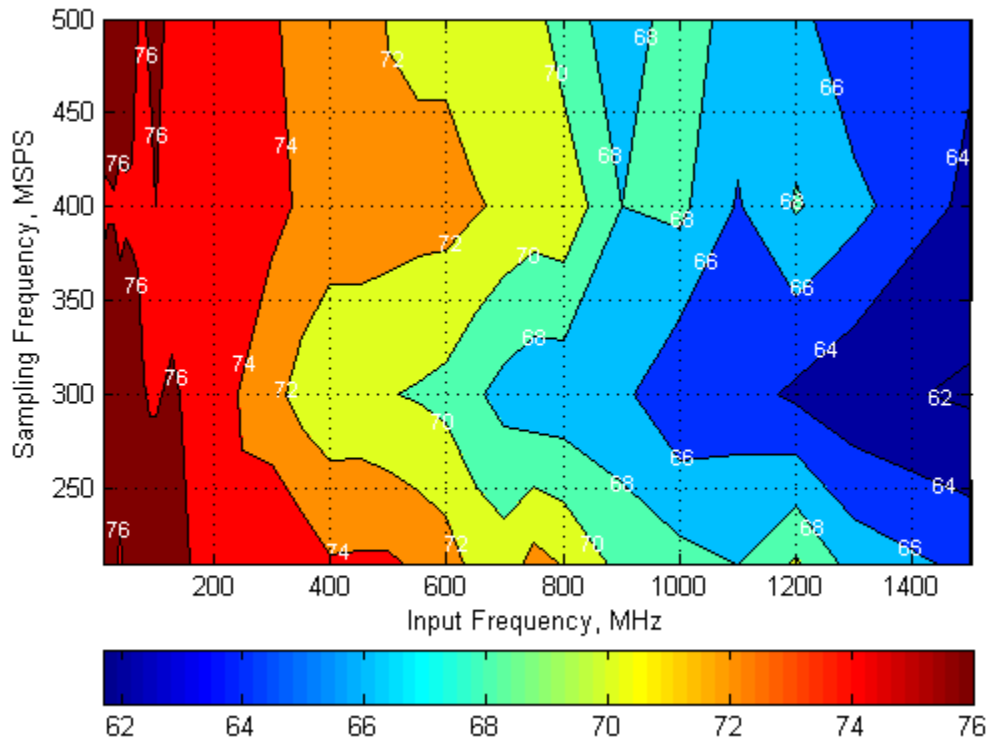
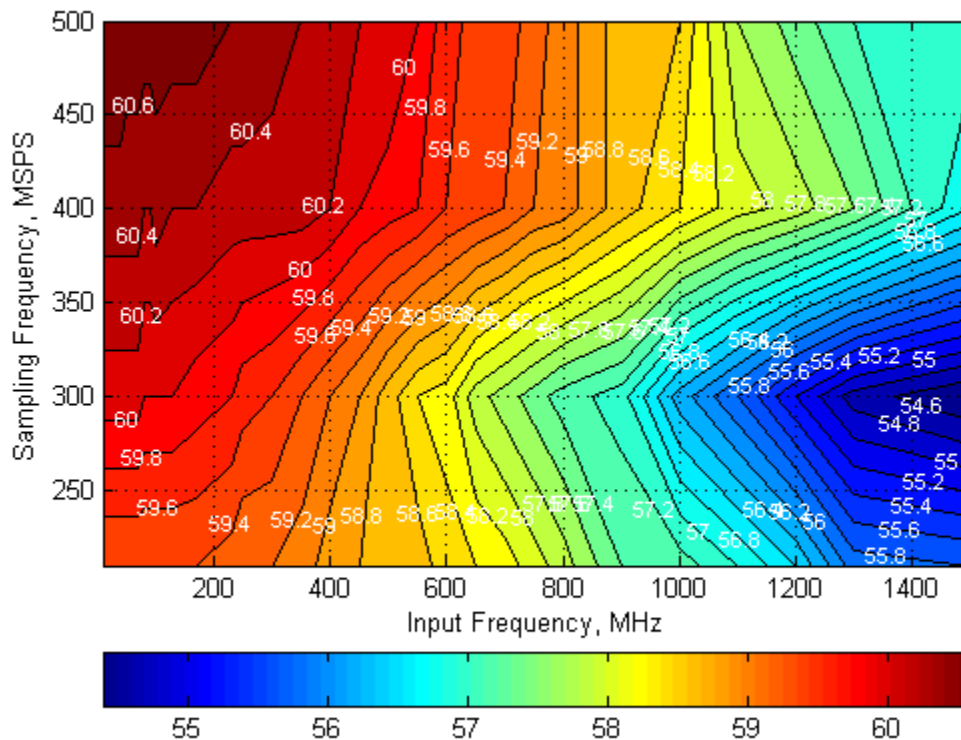


Figure 32.

TYPICAL CHARACTERISTICS (continued)

Typical values at $T_A = +25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = +85^\circ\text{C}$, ADC sampling rate = 500MSPS, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVD/IOVDD = 1.8V, -1dBFS differential input, unless otherwise noted.

SNR Across Input and Sampling Frequencies (auto on)



TYPICAL CHARACTERISTICS (continued)

Typical values at $T_A = +25^\circ\text{C}$, full temperature range is $T_{\text{MIN}} = -40^\circ\text{C}$ to $T_{\text{MAX}} = +85^\circ\text{C}$, ADC sampling rate = 500Mps, 50% clock duty cycle, AVDD33 = 3.3V, AVDDC/AVDD18/DVDD/DVDDLVS/IOVDD = 1.8V, -1dBFS differential input, unless otherwise noted.

SNR Across Input and Sampling Frequencies (auto on)

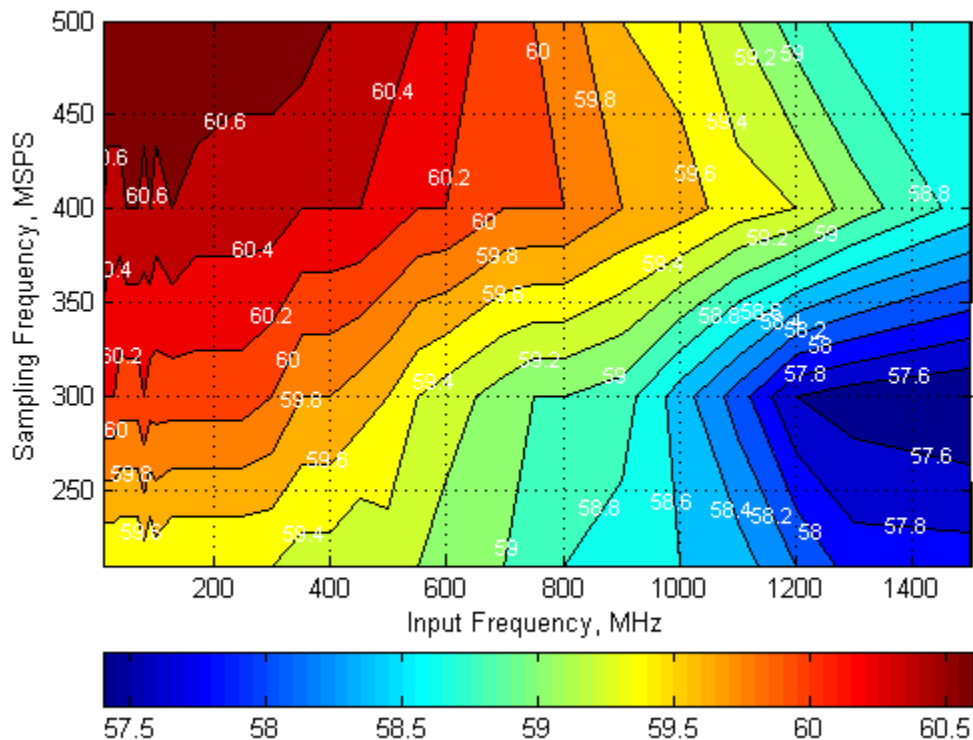


Figure 34.

FEATURES

POWER DOWN MODES

The ADS5403 can be configured via SPI write (address x37) to a stand-by, light or deep sleep power mode which is controlled by the ENABLE pin. The sleep modes are active when the ENABLE pin goes low. Different internal functions stay powered up which results in different power consumption and wake up time between the two sleep modes.

Sleep mode	Wake up time	Power Consumption Auto correction disabled	Power Consumption Auto correction enabled
Complete Shut Down	2.5 ms	7mW	7mW
Stand-by	100μs	7mW	7mW
Deep Sleep	20μs	220mW	305mW
Light Sleep	2μs	367mW	448mW

TEST PATTERN OUTPUT

The ADS5403 can be configured to output different test patterns that can be used to verify the digital interface is connected and working properly. To enable the test pattern mode, the high performance mode 1 has to be disabled first via SPI register write. Then different test patterns can be selected by configuring registers x3C, x3D and x3E. All three registers must be configured for the test pattern to work properly.

First set HP1 = 0 (Addr 0x01, D01)

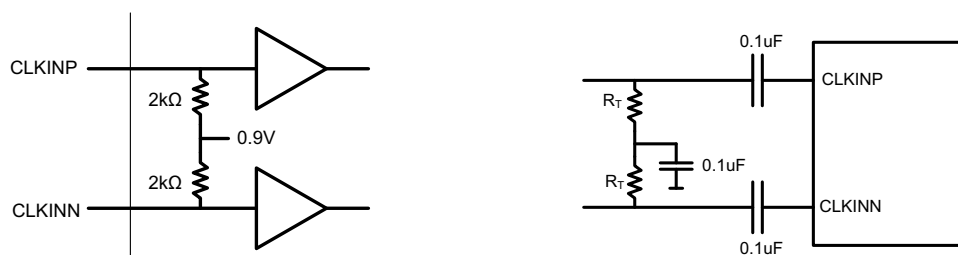
Register Address	All 0s	All 1s	Toggle (0xAAA => 0x555)	Toggle (0xFFFF => 0x000)
0x3C	0x8000	0xBFFC	0x9554	0xBFFC
0x3D	0x0000	0x3FFC	0x2AA8	0x0000
0x3E	0x0000	0x3FFC	0x1554	0x3FFC

Register Address	Custom Pattern															
	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
x3C	1	0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	0	0
x3D	0	0													0	0
x3E	0	0													0	0

For normal operation, set HP1 = 1 (Addr 0x01, D01) and 0x3C, 0x3D, 0x3E all to 0.

CLOCK INPUT

The ADS5403 clock input can be driven differentially with a sine wave, LVPECL or LVDS source with little or no difference in performance. The common mode voltage of the clock input is set to 0.9V using internal 2kΩ resistors. This allows for AC coupling of the clock inputs. The termination resistors should be placed as close as possible to the clock inputs in order to minimize signal reflections and jitter degradation.



Recommended differential clock driving circuit

Figure 35. Recommended Differential Clock Driving Circuit

SNR AND CLOCK JITTER

The signal to noise ratio of the ADC is limited by three different factors: the quantization noise is typically not noticeable in pipeline converters and is 74dB for a 12bit ADC. The thermal noise limits the SNR at low input frequencies while the clock jitter sets the SNR for higher input frequencies.

$$\text{SNR}_{\text{ADC}}[\text{dBc}] = -20 \times \log \sqrt{\left(10 - \frac{\text{SNR}_{\text{Quantization_Noise}}}{20}\right)^2 + \left(10 - \frac{\text{SNR}_{\text{ThermalNoise}}}{20}\right)^2 + \left(10 - \frac{\text{SNR}_{\text{Jitter}}}{20}\right)^2} \quad (1)$$

The SNR limitation due to sample clock jitter can be calculated as following:

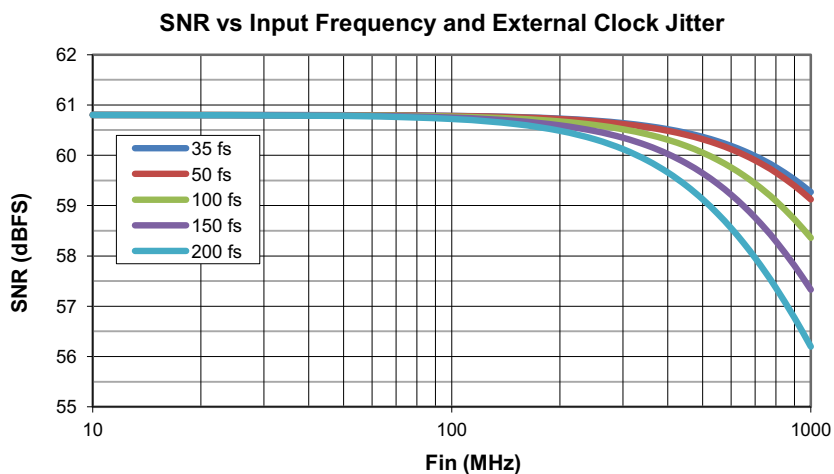
$$\text{SNR}_{\text{Jitter}}[\text{dBc}] = -20 \times \log(2\pi \times f_{\text{IN}} \times t_{\text{Jitter}}) \quad (2)$$

The total clock jitter (TJitter) has three components – the internal aperture jitter (100fs for ADS5403) which is set by the noise of the clock input buffer, the external clock jitter and the jitter from the analog input signal. It can be calculated as following:

$$T_{\text{Jitter}} = \sqrt{(T_{\text{Jitter,Ext.Clock_Input}})^2 + (T_{\text{Aperture_ADC}})^2} \quad (3)$$

External clock jitter can be minimized by using high quality clock sources and jitter cleaners as well as bandpass filters at the clock input while a faster clock slew rate improves the ADC aperture jitter.

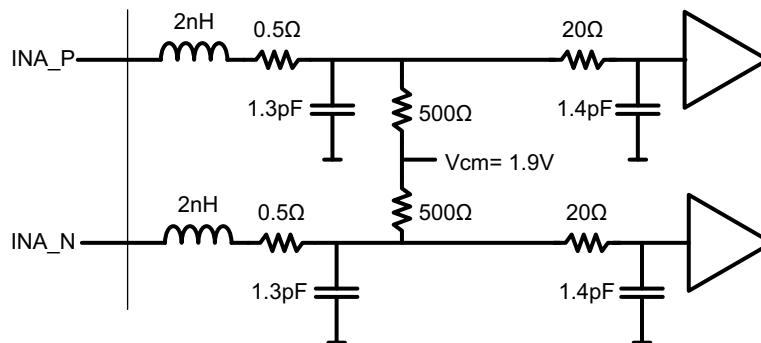
The ADS5403 has a thermal noise of 60.8 dBFS and internal aperture jitter of 100fs. The SNR depending on amount of external jitter for different input frequencies is shown in the following figure.



ANALOG INPUTS

The ADS5403 analog signal input is designed to be driven differentially. The analog input pins have internal analog buffers that drive the sampling circuit. As a result of the analog buffer, the input pins present a high impedance input across a very wide frequency range to the external driving source which enables great flexibility in the external analog filter design as well as excellent 50Ω matching for RF applications. The buffer also helps to isolate the external driving circuit from the internal switching currents of the sampling circuit which results in a more constant SFDR performance across input frequencies.

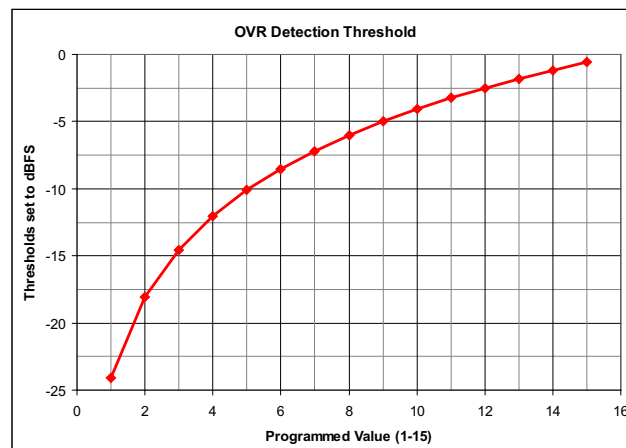
The common-mode voltage of the signal input is internally biased to 1.9V using 500Ω resistors which allows for AC coupling of the input drive network. Each input pin (INP, INM) must swing symmetrically between $(V_{CM} + 0.25V)$ and $(V_{CM} - 0.25V)$, resulting in a 1.0Vpp (default) differential input swing. The input sampling circuit has a 3dB bandwidth that extends up to 1.2GHz.



OVER-RANGE INDICATION

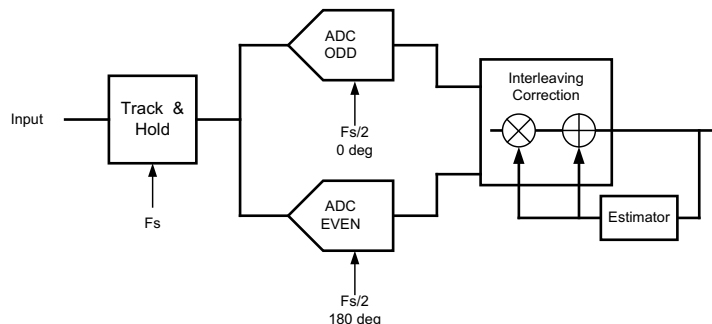
The ADS5403 provides a fast over-range indication on the OVRA/B pins. The fast OVR is triggered if the input voltage exceeds the programmable overrange threshold and it gets presented after just 12 clock cycles enabling a quicker reaction to an overrange event. The OVR threshold can be configured using SPI register writes.

The input voltage level at which the overload is detected is referred to as the threshold and is programmable using the Over-range threshold bits. The threshold at which fast OVR is triggered is $(\text{full-scale} \times [\text{the decimal value of the FAST OVR THRESH bits}] / 16)$. After reset, the default value of the over-range threshold is set to 15 (decimal) which corresponds to a threshold of 0.56dB below full scale ($20 \cdot \log(15/16)$).



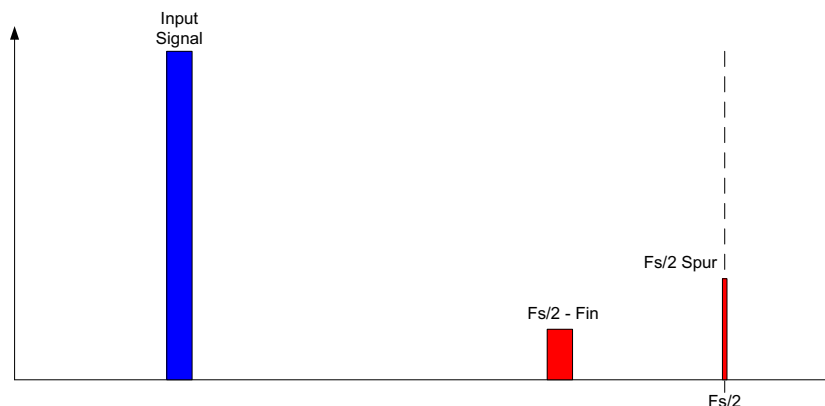
INTERLEAVING CORRECTION

The data converter channel consists of two interleaved ADCs each operating at half of the ADC sampling rate but 180° out of phase from each other. The front end track and hold circuitry is operating at the full ADC sampling rate which minimizes the timing mismatch between the two interleaved ADCs. In addition the ADS5403 is equipped with internal interleaving correction logic that can be enabled via SPI register write.



The interleaving operation creates 2 distinct and interleaving products:

- $F_s/2 - F_{in}$: this spur is created by gain timing mismatch between the ADCs. Since internally the front end track and hold is operated at the full sampling rate, this component is greatly improved and mostly dependent on gain mismatch.
- $F_s/2$ Spur: due to offset mismatch between ADCs



The auto correction loop can be enabled via SPI register write in address 0x01 and resetting the correction circuit in address 0x03 and 0x1A. By default it is disabled for lowest possible power consumption. The default settings for the auto correction function should work for most applications. However please contact Texas Instruments if further fine tuning of the algorithm is required.

The auto correction function yields best performance for input frequencies below 250MHz. For input frequencies greater than 250MHz it is recommended to disable the auto gain correction loop.

RECEIVE MODE: DECIMATION FILTER

There is an optional digital decimation filter in the data path as shown in Figure 36. The filter can be programmed as a low-pass or a high-pass filter and the normalized frequency response of both filters is shown in Figure 37.

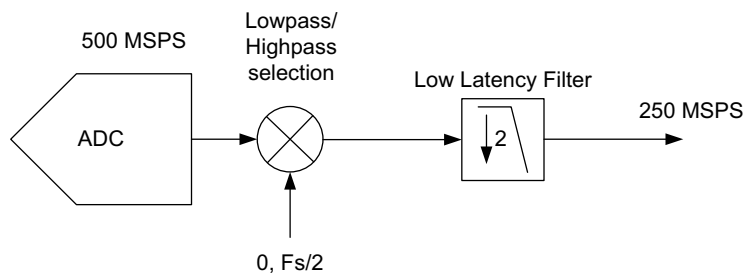


Figure 36.

The decimation filter response has a 0.1dB pass band ripple with approximately 41% pass-band bandwidth. The stop-band attenuation is approximately 40dB.

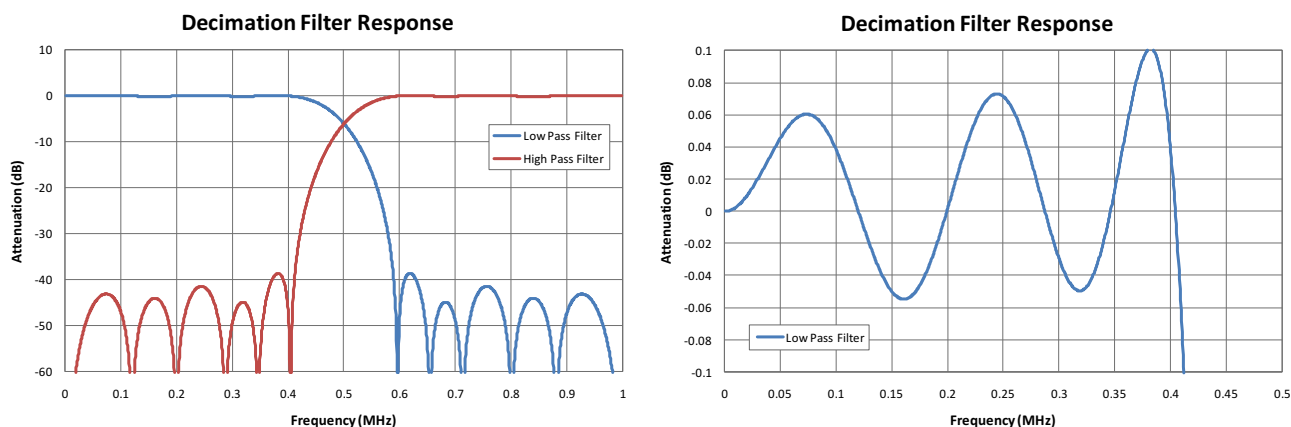
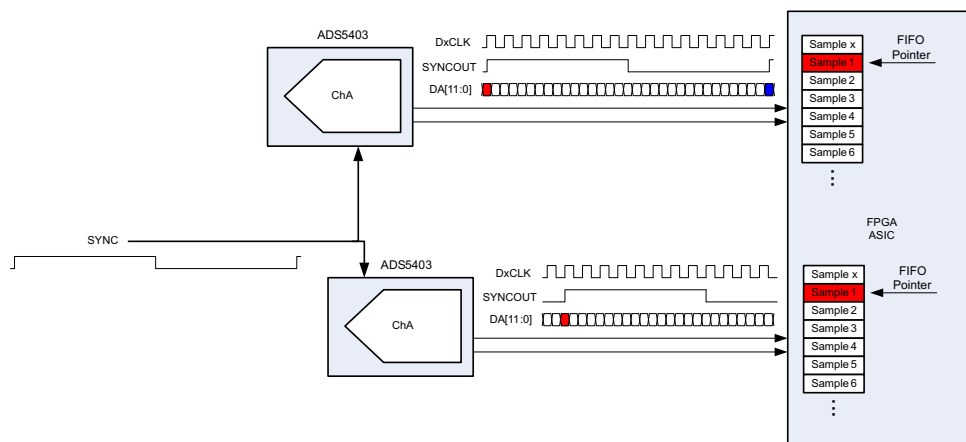


Figure 37.

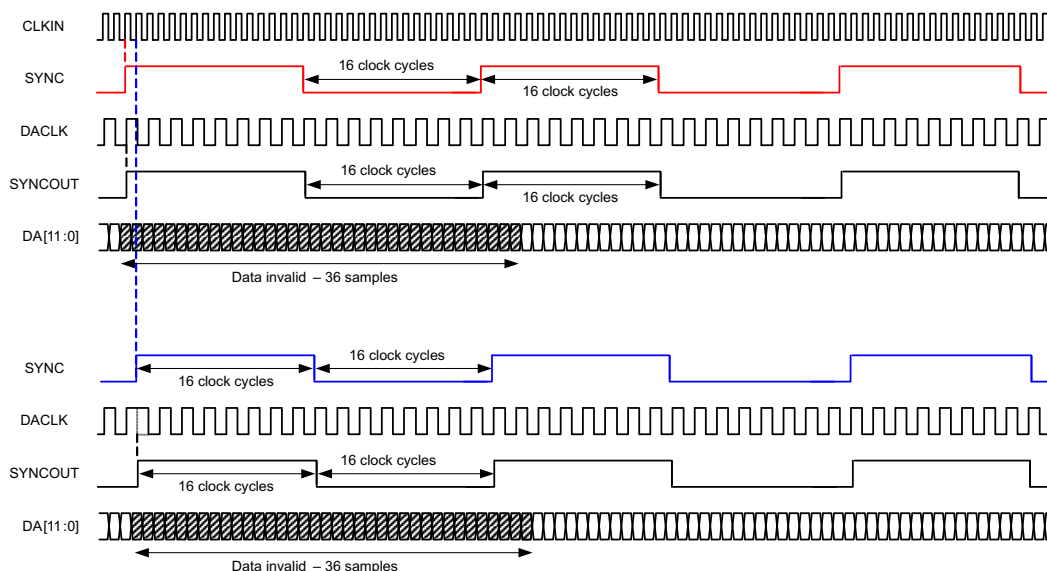
MULTI DEVICE SYNCHRONIZATION

The ADS5403 simplifies the synchronization of data from multiple ADCs in one common receiver. Upon receiving the initial SYNC input signal, the ADS5403 resets all the internal clocks and digital logic while also starting a SYNCOUT signal which operates on a 5bit counter (32 clock cycles). Therefore by providing a common SYNC signal to multiple ADCs their output data can be synchronized as the SYNCOUT signal marks a specific sample with the same latency in all ADCs. The SYNCOUT signal then can be used in the receiving device to synchronize the FIFO pointers across the different input data streams. Thus the output data of multiple ADCs can be aligned properly even if there are different trace lengths between the different ADCs.



The SYNC input signal should be a periodic signal repeating every 32 CLKIN clock cycles. It gets registered on the rising edge of the ADC input clock (CLKIN). Upon registering the initial rising edge of the SYNC signal, the internal clocks and logic get reset which results in invalid output data for 36 samples (1 complete sync cycle and 4 additional samples). The SYNCOUT signal starts with the next output clock (DACLK) rising edge and operates on a 5-bit counter independent from the SYNC signal frequency and duty cycle.

Since the ADS5403 output interface operates with a DDR clock, the synchronization can happen on the rising edge or falling edge sample. Synchronization on the falling edge sample will result in a half cycle clock stretch of DACLK. For convenience the SYNCOUT signal is available on the ChA output LVDS bus. When using decimation the SYNCOUT signal still operates on 32 clock cycles of CLKIN but since the output data is decimated by 2, only the first 18 samples should be discarded.



PROGRAMMING INTERFACE

The serial interface (SIF) included in the ADS5403 is a simple 3 or 4 pin interface. In normal mode, 3 pins are used to communicate with the device. There is an enable (SDENB), a clock (SCLK) and a bi-directional IO port (SDIO). If the user would like to use the 4 pin interface one write must be implemented in the 3 pin mode to enable 4 pin communications. In this mode, the SDO pin becomes the dedicated output. The serial interface has an 8-bit address word and a 16-bit data word. The first rising edge of SCLK after SDENB goes low will latch the read/write bit. If a high is registered then a read is requested, if it is low then a write is requested. SDENB must be brought high again before another transfer can be requested. The signal diagram is shown below:

Register Initialization

After power up, the internal registers must be initialized to the default values. This initialization can be accomplished in one of two ways:

1. Either through hardware reset by applying a low pulse on SRESET pin
2. By applying a software reset. When using the serial interface, a reset can be performed by addressing register x2C. This setting initializes the internal registers to the default values and then self-resets the RESET register to 0. In this case the SRESET pin can be kept high.

Serial Register Write

The internal register of the ADS5403 can be programmed following these steps:

1. Drive SDENB pin low
2. Set the R/W bit to '0' (bit A7 of the 8 bit address)
3. Initiate a serial interface cycle specifying the address of the register (A6 to A0) whose content has to be written
4. Write 16bit data which is latched on the rising edge of SCLK

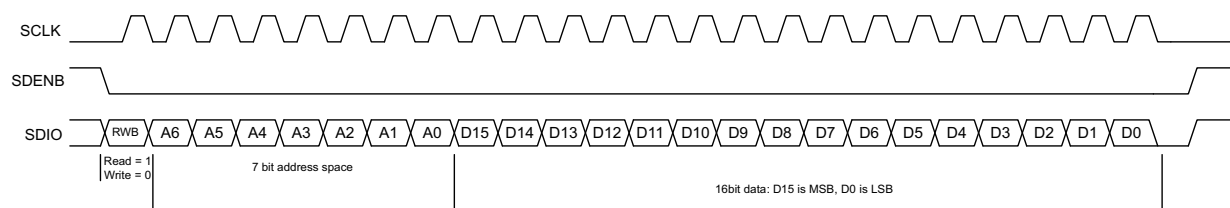


Figure 38. Serial Register Write Timing Diagram

PARAMETER		MIN	TYP ⁽¹⁾	MAX	UNIT
f_{SCLK}	SCLK frequency (equal to $1/t_{SCLK}$)	>DC		20	MHz
t_{SLOADS}	SDENB to SCLK setup time	25			ns
t_{SLOADH}	SCLK to SDENB hold time	25			ns
t_{DSU}	SDIO setup time	25			ns
t_{DH}	SDIO hold time	25			ns

(1) Typical values at +25°C; minimum and maximum values across the full temperature range: TMIN = -40°C to TMAX = +85°C, AVDD3V = 3.3V, AVDD, DRVDD = 1.9V, unless otherwise noted.

Serial Register Readout

The device includes a mode where the contents of the internal registers can be read back using the SDO/SDIO pins. This read-back mode may be useful as a diagnostic check to verify the serial interface communication between the external controller and the ADC.

1. Drive SDENB pin low
2. Set the RW bit (A7) to '1'. This setting disables any further writes to the registers
3. Initiate a serial interface cycle specifying the address of the register (A6 to A0) whose content has to be read.
4. The device outputs the contents (D15 to D0) of the selected register on the SDO/SDIO pin
5. The external controller can latch the contents at the SCLK rising edge.
6. To enable register writes, reset the RW register bit to '0'.

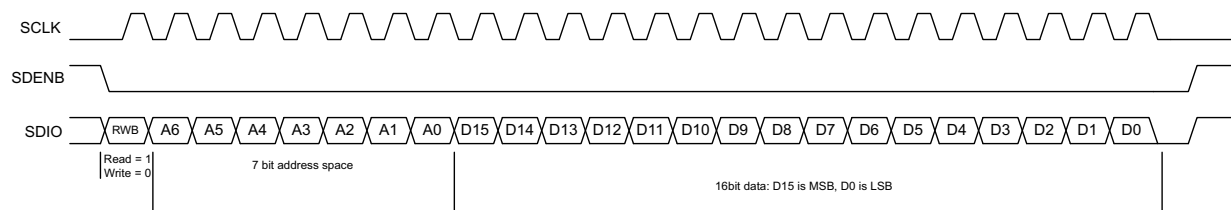


Figure 39. Serial Register Read Timing Diagram

SERIAL REGISTER MAP⁽²⁾

(2) Multiple functions in a register can be programmed in a single write operation.

Register Address	Register Data															
A7-A0 IN HEX	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	3/4 Wire SPI	Decimation Filter EN	0	ChA High/Low Pass	0	0	0	0	0	0	0	0	0	0	0	0
1	ChA Corr EN	0	0	0	0	0	0	0	0	0	0	0	Data Format	0	Hp Mode1	0
2	0	0	0	0	0	Over-range threshold				0	0	0	0	0	0	0
3	0	Start Auto Corr ChA	0	0	1	0	1	1	0	0	0	1	1	0	0	0
E	Sync Select															
F	Sync Select				0	0	0	0	0	VREF Set			0	0	0	0
2B	0	0	0	0	0	0	0	Temp Sensor								
2C	Reset															
37	Sleep Modes		0	0	0	0	0	0	0	0	0	0	0	0	0	0
38	HP Mode2									BIAS EN	SYNC EN	LP Mode 1	0	0	0	0
3A	LVDS Current Strength			LVDS SW		Internal LVDS Termination		0	0	0	0	DACLK EN	LP Mode 2	0	OVRA EN	LP Mode 3
66	LVDS Output Bus A EN															

DESCRIPTION OF SERIAL INTERFACE REGISTERS

Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	3/4 Wire SPI	Decimation Filter EN	0	ChA High/Low Pass	0	0	0	0	0	0	0	0	0	0	0	0

- D15 3/4 Wire SPI** Enables 4-bit serial interface when set
Default 0
- 0 3 wire SPI is used with SDIO pin operating as bi-directional I/O port
- 1 4 wire SPI is used with SDIO pin operating as data input and SDO pin as data output port.
- D14 Decimation Filter EN** 2x decimation filter is enabled when bit is set
Default 0
- 0 Normal operation with data output at full sampling rate
- 1 2x decimation filter enabled
- D12 ChA High/Low Pass** (Decimation filter must be enabled first: set bit D14)
Default 0
- 0 Low Pass
- 1 High Pass

Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	ChA Corr EN	0	0	0	0	0	0	0	0	0	0	0	Data Format	0	HP Mode1	0

D15 ChA Corr EN (should be enabled for maximum performance)

Default 0

0 auto correction disabled

1 auto correction enabled

D3 Data Format

Default 0

0 Two's complement

1 Offset Binary

D1 HP Mode 1

Default 0

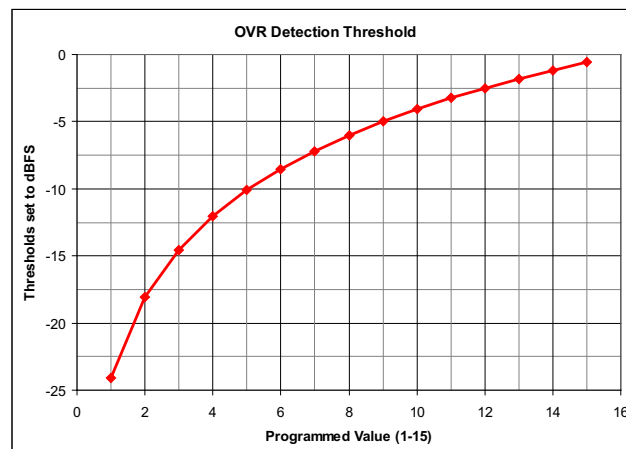
1 Must be set to 1 for optimum performance

Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
2	0	0	0	0	0	Over-range threshold					0	0	0	0	0	0

D10-D7 Over-range threshold

The over-range detection is triggered 12 output clock cycles after the overload condition occurs. The threshold at which the OVR is triggered = $1.0V \times [\text{decimal value of } \langle \text{Over-range threshold} \rangle] / 16$. After power up or reset, the default value is 15 (decimal) which corresponds to a OVR threshold of 0.56dB below fullscale ($20 \times \log(15/16)$). This OVR threshold is applicable to both channels.

Default 1111



Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
3	0	Start Auto Corr ChA	0	0	1	0	1	1	0	0	0	1	1	0	0	0

D14 Start Auto Corr ChA Starts DC offset and Gain correction loop for ChA
 Default 1
 0 Starts offset and Gain correction loop for ChA
 1 Clears DC offset correction value to 0 and Gain correction value to 1
D11, 9, 8, 4, 3 Must be set to 1 for maximum performance
 Default 1

Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
E	Sync Select														0	0

D15-D2 Sync Select Sync selection for the clock generator block (also need to see address 0x0F)
 Default 1010 1010
 1010 10
 0000 0000 0000 00 Sync is disabled
 0101 0101 0101 01 Sync is set to one shot (one time synchronization only)
 1010 1010 1010 10 Sync is derived from SYNC input pins
 1111 1111 1111 11 not supported

Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
F	Sync Select				0	0	0	0	0	VREF Sel			0	0	0	0

D15-D12 Sync Select Sync selection for the clock generator block
 Default 1010
 0000 Sync is disabled
 0101 Sync is set to one shot (one time synchronization only)
 1010 Sync is derived from SYNC input pins
 1111 not supported
D6-D4 VREF SEL Internal voltage reference selection
 Default 000
 000 1.0V
 001 1.25V
 010 0.9V
 011 0.8V
 100 1.15V
 Others external reference

Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
2B	0	0	0	0	0	0	0	Temp Sensor								

D8-D0 **Temp Sensor** Internal temperature sensor value – read only

Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
2C	Reset															

D15-D0 **Reset** This is a software reset to reset all SPI registers to their default value. Self Default 0000 clears to 0.

1101001011110000 Perform software reset

Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
37	Sleep Modes						0	0	0	0	0	0	0	0	0	0

D15-D14 **Sleep Modes** Sleep mode selection which is controlled by the ENABLE pin. Sleep modes are active when Default 00 ENABLE pin goes low.

000000 Complete shut down Wake up time 2.5 ms

100000 Stand-by mode Wake up time 100 μ s

110000 Deep sleep mode Wake up time 20 μ s

110101 Light sleep mode Wake up time 2 μ s

Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
38	HP Mode 2									Bias EN	SYNC EN	LP Mode 1	0	0	0	0

D15-D7 HP Mode 2

Default 11111111

1 Set to 1 for normal operation

D6 BIAS EN

Default 1

Enables internal fuse bias voltages – can be disabled after power up to save power.

0 Internal bias powered down

1 Internal bias enabled

D5 SYNC EN

Default 1

Enables the SYNC input buffer.

0 SYNC input buffer disabled

1 SYNC input bffer enabled

D4 LP Mode

Default 1

Low power mode 1 to disable internal unused input buffer.

0 Internal input buffer disabled

1 Internal input buffer enabled

Register Address	Register Data															
A7-A0 in hex	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
3A	LVDS Current Strength			LVDS SW		Internal LVDS Termination		0	0	0	0	DACLK EN	LP Mode 2	0	OVRA EN	LP Mode 3

D15-D13 **LVDS Current Strength** LVDS output current strength.

Default 000

000	2 mA	100	3 mA
001	2.25 mA	101	3.25 mA
010	2.5 mA	110	3.5 mA
011	2.75 mA	111	3.75 mA

D12-D11 **LVDS SW** LVDS driver internal switch setting – correct range must be set for setting in D15-D13

Default 01

01	2 mA to 2.75 mA
11	3mA to 3.75mA

D10-D9 **Internal LVDS Termination** Internal termination

Default 00

00	2 k Ω
01	200 Ω
10	200 Ω
11	100 Ω

D4 **DACLK EN** Enable DACLK output buffer

Default 1

0	DACLK output buffer powered down
1	DACLK output buffer enabled

D3 **LP Mode 2** Low power mode to disable unused internal output buffer

Default 1

0	Internal output buffer disabled
1	Internal output buffer enabled

D1 **OVRA EN** Enable OVRA output buffer

Default 1

0	OVRA output buffer powered down
1	OVRA output buffer enabled

D0 **LP Mode 3** Low power mode to disable unused internal output buffer

Default 1

0	Internal output buffer disabled
1	Internal output buffer enabled

Register Address	Register Data															
	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
A7-A0 in hex																
66	LVDS Output Bus A EN															

D15-D10	LVDS Output Bus A EN	Individual LVDS output pin power down for channel A
	Default FFFF	
0	Output is powered down	
1	Output is enabled	
D15	Pins N7, P7 (no connect pins) which are not used and should be powered down for power savings	
D14	Pins N6, P6 (no connect pins) which are not used and should be powered down for power savings	
D13	SYNCOUTP/N (pins N5, P5)	
D12	Pins N4, P4 (no connect pins) which are not used and should be powered down for power savings	
D11-D0	corresponds to DA11-D0	

REVISION HISTORY

Changes from Original (February 2013) to Revision A	Page
• Changed D15-10 in register 66 From: Individual LVDS output pin power down for channel B To: Individual LVDS output pin power down for channel A	34
• Changed D13 in Register 66 From: SYNCOUTP/N (pins F1, F2) To: SYNCOUTP/N (pins N5, P5)	34
• Changed D12 in egister 66 From: "Pins E3, E4..." To: "Pins N4, P4..."	34
• Changed D11-D10 - corresponds to DB11-DB0 in Register 66 To: D11-D0 - corresponds to DA11-D0	34

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
ADS5403IZAY	ACTIVE	NFBGA	ZAY	196	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 85	ADS5403I	Samples
ADS5403IZAYR	ACTIVE	NFBGA	ZAY	196	1000	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 85	ADS5403I	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

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TAPE AND REEL INFORMATION


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ADS5403IZAYR	NFBGA	ZAY	196	1000	330.0	24.4	12.3	12.3	2.3	16.0	24.0	Q1

TAPE AND REEL BOX DIMENSIONS

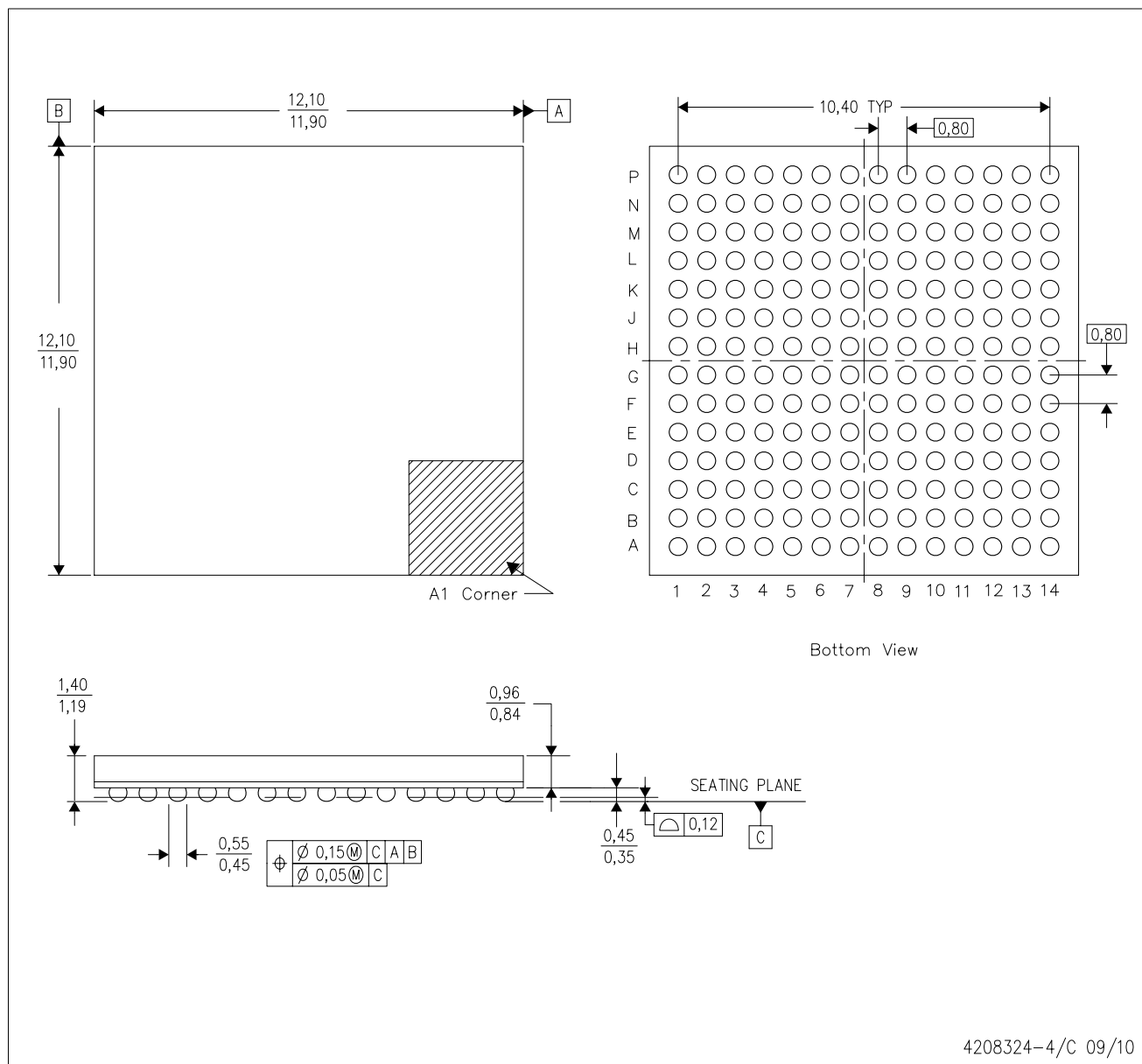


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ADS5403IZAYR	NFBGA	ZAY	196	1000	336.6	336.6	31.8

ZAY (S-PBGA-N196)

PLASTIC BALL GRID ARRAY



- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - B. This drawing is subject to change without notice.
 - C. This is a Pb-free solder ball design.

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