

## **RF3863** WIDE BANDWIDTH. HIGH LINEARITY

## LOW NOISE AMPLIFIER

Package Style: QFN, 16-Pin, 3mmx3mm



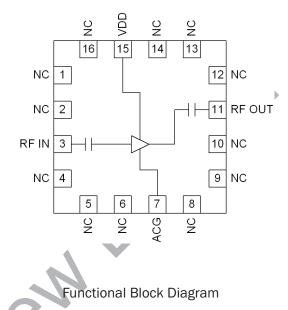


### Features

- Low Noise and High Intercept Point
- Adjustable Bias Current for Enhanced IP3
- Single 2.5V to 6.0V Power Supply
- 700 MHz to 3800 MHz Operation
- QFN16 3mmx3mm Package

### **Applications**

- First Stage CDMA, PCS, GSM/EDGE, UMT S LNA/Linear Driver
- First Stage WLAN LNA/Linear Driver
- First Stage WiMAX LNA/Linear Driver
- General Purpose Amplification



## **Product Description**

The RF3863 is a low noise amplifier with a high output IP3. The amplifier is self-biased from a single voltage supply with 50 $\Omega$  input and output ports. The useful frequency range is from 700MHz to 3800MHz. A 0.8dB noise figure and 36dBm OIP3 performance is achieved with a 5V  $V_{DD},$  90mA. Current can be increased to raise OIP3 while having minimal effect on noise figure. The IC is featured in a standard QFN, 16-pin, 3mmx3mm package.

#### **Ordering Information**

RF3863 RF3863PCK-410 Wide Bandwidth, High Linearity Low Noise Amplifier Fully Assembled Evaluation Board with 5 Sample Parts 1.5GHz to 2.7GHz

#### **Optimum Technology Matching® Applied**

🗌 GaAs HBT	□ SiGe BiCMOS	🗹 GaAs pHEMT	🗌 GaN HEMT
GaAs MESFET	🗌 Si BiCMOS	🗌 Si CMOS	□ RF MEMS
🗌 InGaP HBT	SiGe HBT	🗌 Si BJT	

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#### **Absolute Maximum Ratings**

-					
Parameter	Rating	Unit			
Supply Voltage	6	V			
Input RF Level	+10	dBm			
Current Drain, I <sub>DD</sub>	150	mA			
Operating Ambient Temperature	-40 to +85	°C			
Storage Temperature	-40 to +150	°C			

Note 1: Max continuous RF IN is +10dBm. The max transient RF IN is +20dBm.



Caution! ESD sensitive device.

Exceeding any one or a combination of the Absolute Maximum Rating conditions may cause permanent damage to the device. Extended application of Absolute Maximum Rating conditions to the device may reduce device reliability. Specified typical perfor-mance or functional operation of the device under Absolute Maximum Rating condi-tions is not implied.

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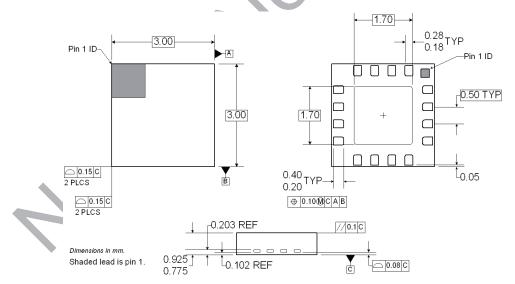
	Specification					
Parameter	Min.	Тур.	Max.	Unit	Condition	
High Band		196.				
Frequency	3.3		3.8	GHz	V <sub>DD</sub> =5V	
Current		90	110	mA		
Gain		10	110	dB		
Noise Figure		0.9		dB		
OIP3		37.0		dBm	+25°C, V <sub>DD</sub> =5V, I <sub>DD</sub> =90mA, 3500MHz	
		01.0		d Bill	unless specified	
OP1dB		22.0		dBm		
S11		-11		dB		
S22		-18		dB	f <sub>1</sub> =3500MHz, f <sub>2</sub> =3501MHz	
Mid Band						
Frequency	1.5		2.7	GHz		
Current		90	110	mA	V <sub>DD</sub> =5V	
Gain	14	15	16.5	dB	+25°C, V <sub>DD</sub> =5V, I <sub>DD</sub> =90mA, 2000MHz unless specified	
Noise Figure		0.8	1.0	dB		
OIP3	33.0	35.5		dBm	f <sub>1</sub> =2000 MHz, f <sub>2</sub> =2001 MHz	
OP1dB	21.0	22.5	25.0	dBm		
S11		-10		dB		
S22		-17		dB		
Low Band						
Frequency	700		1100	MHz		
Current	r	90	110	mA	V <sub>DD</sub> =5V	
Gain		18		dB	+25°C, V <sub>DD</sub> =5V, I <sub>DD</sub> =90mA, 850MHz unless specified	
Noise Figure		0.9		dB		
OIP3		35		dBm	f <sub>1</sub> =850MHz, f <sub>2</sub> =851MHz	
OP1dB		22		dBm		
S11		-10		dB		
S22		-17		dB		
Thermal						
Theta <sub>JC</sub>		51		°C/W		
Power Supply						
Device Operating Voltage	2.5	5.0	6.0	V		
Operating Current	65	90	110	mA	V <sub>DD</sub> =5.0V, R2=open	



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Pin	Function	Description	Interface Schematic
1	NC	Not connected.	
2	NC	Not connected.	
3	RF IN	RF input pin. 50 $\Omega$ matched. This pin is DC-blocked.	
4	NC	Not connected.	
5	NC	Not connected.	
6	NC	Not connected.	
7	ACG	AC ground. Shunt cap may be added for tuning. Shunt resistor may be added to increase $\rm I_{\rm DD}/\rm IP3.$	
8	NC	Not connected.	
9	NC	Not connected.	
10	NC	Not connected.	
11	<b>RF OUT</b>	RF output pin. 50 $\Omega$ matched. This pin is DC-blocked.	
12	NC	Not connected.	
13	NC	Not connected.	
14	NC	Not connected.	
15	VD	Bias voltage. 2.5V to 6.0V applied through bias inductor.	
16	NC	Not connected.	
Pkg Base	GND	Ground connection.	

## Package Drawing

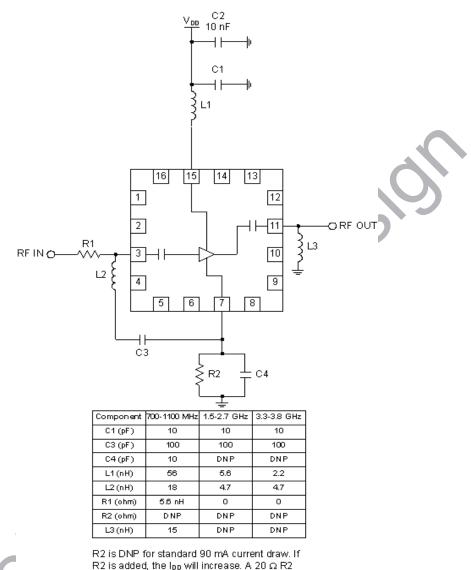


**RF3863** 





### Evaluation Board Schematic 700MHz to 3800MHz



R2 is bown for standard so mA current draw. If R2 is added, the  $l_{DD}$  will increase. A 20  $\Omega$  R2 will raise the current to achieve higher linearity.

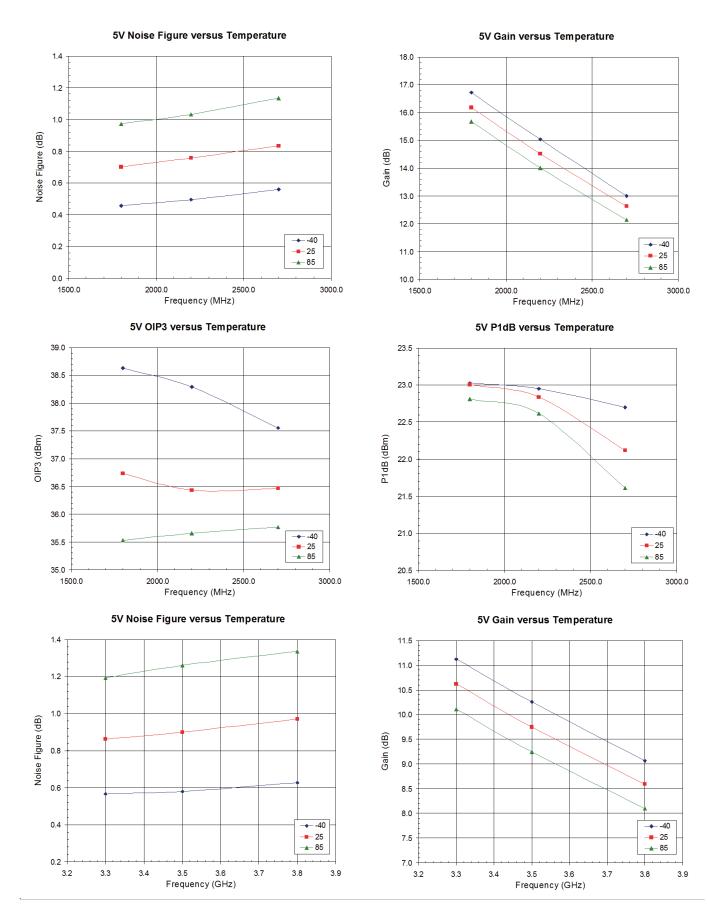


### **Theory of Operation**

Low noise figure/high IP3 make the RF3863 ideal for use as both a receive LNA and a transmit driver for cellular/DCS/PCS/UMTS and WiMax platforms, in addition to many other general purpose applications. Standard evaluation boards cover 700MHz to 1100MHz, 1500MHz to 2700MHz, and 3300MHz to 3800MHz. Viewing the data sheet evaluation board schematic, refer to the information below for purpose/function of external components:

• R1/L3 (0Ω/unpopulated on standard evaluations boards): These unused components were placed for convenience and flexibility when needed to optimize matching for an out-of-band application.





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5V OIP3 versus Temperature 5V P1dB versus Temperature 40.0 23.6 23.4 39.5 23.2 39.0 23.0 38.5 22.8 38.0 22.6 OIP3 (dBm) P1dB (dBm) 22.4 37.5 22.2 37.0 22.0 21.8 36.5 21.6 36.0 21.4 -40 35.5 - 25 21.2 . 85 35.0 21.0 3.2 3.3 3.4 3.5 3.6 3.7 3.8 3.9 3.2 3.3 3.4 3.5 3.6 3.7 3.8 3.9 Frequency (GHz) Frequency (GHz) RF3863- 410 Evaluation Board, OIP3 vs R2 value Vdd = 5.0 V Optimum OIP3 enhancement at R2 = 20 ohm. RF3863 lcc vs R2 Vdd = 5.0 V = 20 ohm for optimal OIP3 39.6 39.4 135 39.2 130 39 125 38.8 + R2 uni 120 (dBm) R2 unpopulate
R2 = 27 ohm
R2 = 20 ohm
R2 = 20 ohm
R2 = 18 ohm
R2 = 10 ohm 115 38.6 OIP3 • R2 = 10 ohm 110 38.4 105 38.2 100 38 95 37.8 90 37.6 85 2250 1750 2150 2350 1850 1950 2050 R2 unpopulated R2 = 27 ohm R2 = 20 ohm R2 = 18 ohm R2 = 10 ohm Frequency (M Hz)



- L2/C3/C4: Placed to optimize input match and enhance out-of-band low frequency stability.
- R2: Optionally placed to increase bias current and IP3. 20Ω value is found to be the best case. (See graphs.)
- L1/C1: Influence output return loss.

The RF3863 has internal DC-blocking capacitors at RF IN and RF OUT. In addition, impedance seen looking out at pins 7/15 has been shown to influence response. As a result, two port S-parameters become non-applicable. In the event matching is desired for frequency bands outside of those provided with standard evaluation boards, the following approach can be used:

- Start with matching seen for standard evaluation board closest to the desired band of operation.
- Optimize values at L2/C4/L1 to obtain response/performance.

When considering use of RF3863 outside of standard frequency bands, it is advisable to consider RF3861 as an option. We can describe the difference between the two designs here to illustrate that point. That difference, in terms of performance, can be outlined by two statements:

1. RF3863, in the primary bands of interest, shows a lower noise figure.

2. In the trade-off here, RF3861 generally sees better return loss, and matching is more easily accomplished.

"Out-of-band" would primarily involve frequencies below 700 MHz. Refer to the RF3861 data sheet's "Theory of Operation" and "Application Schematic" sections where the 700 MHz to 1100 MHz standard evaluation board is specified from 400 MHz to 1300 MHz. Excellent performance is seen, providing an option for lower frequency operation. In addition, RF3861/RF3863 noise figure delta is not present below 700 MHz. The conclusion then, for the case of low frequency of operation, is that RF3861 offers the most easily implemented solution. Simply use the standard RF3861 700 MHz to 1100 MHz evaluation board.



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