

868/915MHz Receiver Evaluation Board Description

Features

J	Single-co	nversion su	ıperhet ar	chitecture	for low	/ external	component	count

- ☐ FSK demodulation with phase-coincidence demodulator
- ☐ Low current consumption in active mode and very low standby current
- ☐ Switchable LNA gain for improved dynamic range
- ☐ RSSI allows signal strength indication and ASK detection

Ordering Information

Part No.

EVB71111-868-FSK-C EVB71111-868-ASK-C EVB71111-915-FSK-C EVB71111-915-ASK-C

Application Examples

- ☐ Tire Pressure Monitoring Systems (TPMS)
- □ Remote Keyless Entry (RKE)
- □ Wireless access control
- □ Alarm and security systems
- □ Garage door openers
- □ Remote Controls
- □ Home and building automation
- □ Low-power telemetry systems

Evaluation Board



General Description

The TH71111 FSK/ASK single-conversion superheterodyne receiver IC is designed for applications in the European 868 MHz industrial-scientific-medical (ISM) band, according to the EN 300 220 telecommunications standard. It can also be used for any other system with carrier frequencies ranging from 800 MHz to 930 MHz (e.g. for applications to FCC part 15).

^{*} EVB71111-XXX-YYY-C with XXX = Reception frequency (868.3MHz or 915MHz) and YYY = Modulation (FSK or ASK).
** The evaluation board is supplied with a SMA connector.



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1 Theory of Operation

1.1 General

With the TH71111 receiver chip, various circuit configurations can be arranged in order to meet a number of different customer requirements. For FSK reception the IF tank used in the phase coincidence demodulator can be constituted by an external ceramic discriminator. In ASK configuration, the RSSI signal is fed to an ASK detector, which is constituted by the operational amplifier.

A double-conversion variant, called TH71112, is also available. This receiver IC allows a higher degree of image rejection, achieved in conjunction with an RF front-end filter. Both RXICs have the same die. At the TH71112, the second mixer (MIX2) is used to down-convert the first IF (IF1) to the second IF (IF2). At the TH71111, MIX2 operates as an amplifier.

Efficient RF front-end filtering is realized by using a SAW, ceramic or helix filter in front of the LNA and by adding an LC filter at the LNA output.

The TH71111 receiver IC consists of the following building blocks:

- PLL synthesizer (PLL SYNTH) for generation of the local oscillator signal LO, parts of the PLL SYNTH
 are: the high-frequency VCO1, the feedback divider DIV_32, a phase-frequency detector (PFD) with
 charge pump (CP) and a crystal-based reference oscillator (RO)
- Low-noise amplifier (LNA) for high-sensitivity RF signal reception
- First mixer (MIX1) for down-conversion of the RF signal to the IF
- IF pre amplifier which is a mixer cell (MIX2) that operates as an amplifier
- IF amplifier (IFA) to amplify and limit the IF signal and for RSSI generation
- Phase coincidence demodulator (DEMOD) with third mixer (MIX3) to demodulate the IF signal
- Operational amplifier (OA) for data slicing, filtering and ASK detection
- Bias circuitry for bandgap biasing and circuit shutdown

1.2 EVB Technical Data Overview

	Input frequency range: 800 MHz to 930 MHz Power supply range: 2.3 V to 5.5 V @ ASK		Maximum input level: -10 dBm @ ASK 0 dBm @ FSK
	2.7 V to 5.5 V @ FSK		Image rejection: > 45 dB (e.g. with 868.3 MHz
	Temperature range: -40 °C to +85 °C		SAW front-end filter and at 10.7 MHz IF)
	Standby current: 50 nA		Spurious emission: < -70 dBm
	Operating current: 7.5 mA @ low gain mode		Input frequency acceptance range: up to ±100 kHz
	9.2 mA @ high gain mode		RSSI range: 70 dB
	Sensitivity: -109 dBm @ ASK 1)		FSK deviation range: ±2.5 kHz to ±80 kHz
	-103 dBm @ FSK 2)		_
	Range of IF: 400 kHz to 22 MHz		
1)	at 4 kbps NRZ, BER = 3·10 ⁻³ , 180 kHz IF filter BV	V, in	cl. 3 dB SAW front-end-filter loss
,	,	•	

2) at 4 kbps NRZ, BER = $3\cdot10^{-3}$, \pm 20 kHz FSK deviation, 180 kHz IF filter BW, incl. 3 dB SAW front-end-filter loss

For more detailed information, please refer to the latest TH71111 data sheet revision

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1.3 Block Diagram

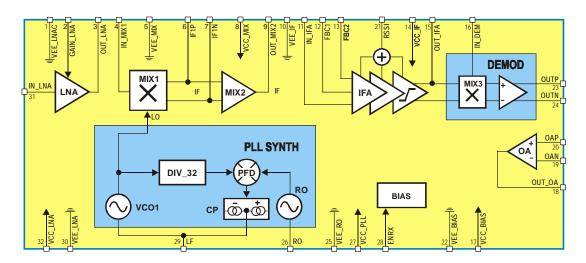


Fig. 1: TH71111 block diagram

1.4 Mode Configurations

ENRX	Mode	Description		
0	RX standby	RX disabled		
1	RX active	RX enable		

Note: ENRX are pulled down internally

1.5 LNA GAIN Control

V _{GAIN_LNA}	Mode	Description
< 0.8 V	HIGH GAIN	LNA set to high gain
> 1.4 V	LOW GAIN	LNA set to low gain

Note: hysteresis between gain modes to ensure stability

1.6 Frequency Planning

Frequency planning is straightforward for single-conversion applications because there is only one IF that can be chosen, and then the only possible choice is low-side or high-side injection of the LO signal (which is now the one and only LO signal in the receiver).

The receiver's single-conversion architecture requires careful frequency planning. Besides the desired RF input signal, there are a number of spurious signals that may cause an undesired response at the output. Among them is the image of the RF signal that must be suppressed by the RF front-end filter.

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By using the internal PLL synthesizer of the TH71111 with the fixed feedback divider ratio of N = 32 (DIV_32), two types of down-conversion are possible: low-side injection of LO and high-side injection of LO. The following table summarizes some equations that are useful to calculate the crystal reference frequency (REF) and the LO frequency, for a given RF and IF.

Injection type	low	high		
REF	(RF – IF)/32	(RF + IF)/32		
LO	32 • REF	32 • REF		
IF	RF – LO	LO – RF		
RF image	RF – 2IF	RF + 2IF		

1.6.1 Selected Frequency Plans

The following table depicts crystal, LO and image signals considering the examples of 868.3 MHz and 915 MHz RF reception at IF = 10.7 MHz.

Signal type	RF = 868.3 MHz	RF = 868.3 MHz	RF = 915 MHz	RF = 915 MHz	
Injection type	low	high	low	high	
REF / MHz	26.80000	27.46875	28.25938	28.92813	
LO / MHz	857.6	879.0	904.3	925.7	
RF image / MHz	846.9	889.7	893.6	936.4	

The selection of the reference crystal frequency is based on some assumptions. As for example: the image frequency should not be in a radio band where strong interfering signals might occur (because they could represent parasitic receiving signals), the LO signal should be in the range of 800 MHz to 930 MHz (because this is the optimum frequency range of the VCO1). Furthermore the IF should be as high as possible to achieve highest RF image rejection. The columns in bold depict the selected frequency plans to receive at 868.3 MHz and 915 MHz, respectively.

1.6.2 Maximum Frequency Coverage

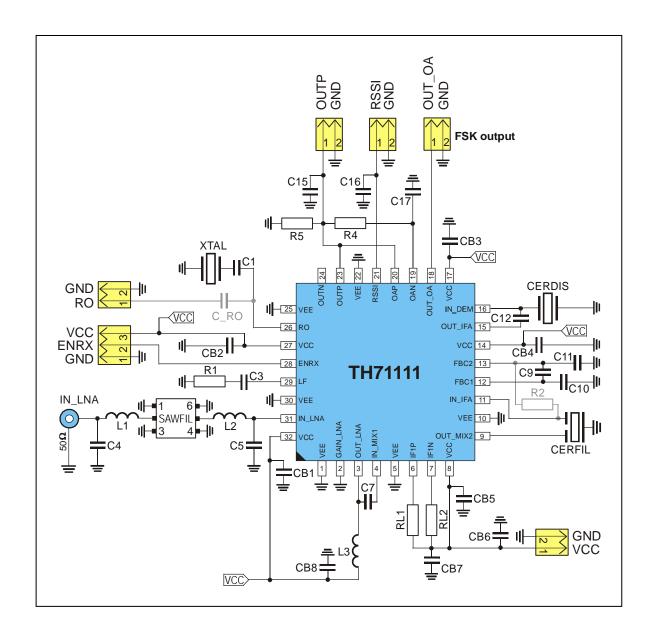
Parameter	f _{min}	f _{max}		
Injection type	high	low		
RF / MHz	789.3	940.7		
REF / MHz	25.0	29.0625		
LO / MHz	800	930		
IF/ MHz	10.7	10.7		



2 Application Circuits

2.1 FSK Application Circuit

2.1.1 Circuit Diagram for FSK Reception



Circuit Features

- Tolerates input frequency variations
- Well-suited for NRZ, Manchester and similar codes



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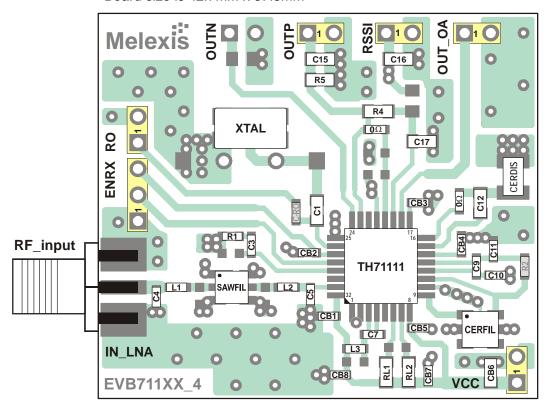
2.1.2 Board Component Values for FSK

Part	Size	Value @ 868.3 MHz	Value @ 915 MHz	Tolerance	Description
C1	0805	22 pF	22 pF	±5%	crystal series capacitor
C3	0603	1 nF	1 nF	±10%	loop filter capacitor
C4	0603	4.7 pF	NIP	±5%	capacitor to match SAW filter input
C5	0603	2.7 pF	NIP	±5% capacitor to match SAW filter outp	
C7	0603	1.0 pF	0.68 pF	±5%	MIX1 input matching capacitor
C9	0603	33 nF	33 nF	±10%	IFA feedback capacitor
C10	0603	1 nF	1 nF	±10%	IFA feedback capacitor
C11	0603	1 nF	1 nF	±10%	IFA feedback capacitor
C12	0805	10 pF	10 pF	±5%	DEMOD phase-shift capacitor
C15	0805	100 pF	100 pF	±5%	demodulator output low-pass capacitor, this value for data rates < 20 kbps NRZ, for higher data rates decrease the value
C16	0805	1.5 nF	1.5 nF	±10%	RSSI output low-pass capacitor
C17	0805	10 nF	10 nF	±10%	data slicer capacitor, this value for data rates > 0.8 kbps NRZ, for lower data rates increase the value
CB1 to CB5 CB7 to CB8	0603	330 pF	330 pF	±10%	de-coupling capacitor
CB6	0805	33 nF	33 nF	±10%	de-coupling capacitor
C_RO	0603	330 pF	330 pF	±5%	optional capacitor, to couple external RO signal
R1	0603	10 kΩ	10 kΩ	±5%	loop filter resistor
R2	0603	330 Ω	330 Ω	±5%	optional CERFIL output matching resistor
R4	0805	330 kΩ	330 kΩ	±5%	data slicer resistor
R5	0805	220 kΩ	220 kΩ	±5%	loading resistor
RL1	0805	470 Ω	470 Ω	±5%	MIX1 bias resistor
RL2	0805	470 Ω	470Ω	±5%	MIX1 bias resistor
L1	0603	22 nH	0 Ω	±5%	SAW filter matching inductor
L2	0603	22 nH	0 Ω	±5%	from Würth-Elektronik (WE-KI series), or equivalent part
L3	0603	10 nH	10 nH	±5%	LNA output tank inductor from Würth-Elektronik (WE-KI series), or equivalent part
XTAL	SMD 6x3.5	26.80000 MHz @ RF = 868.3 MHz	28.25938 MHz @ RF = 915 MHz	±25ppm cal. ±30ppm temp.	fundamental-mode crystal from Telcona/Horizon (HEX22 series) or equivalent par
	HC49 SMD	© 1(1 = 000.5 Wil 12	© 1(1 - 515 WH12	±зорріп тетір.	fundamental-mode crystal, C_{load} = 10 pF to 15 pF, $C_{0, max}$ = 7 pF, $R_{1, max}$ = 50 Ωl
SAWFIL	SMD 3x3	SAFCC868MSL0X00 (f ₀ =868.3 MHz)		B _{3dB} = 2 MHz	low-loss SAW filter from Murata or equivalent part
SAVVIL	SMD 3.8x3.8		SAFCH915MAL0N00 $(f_0 = 915 \text{ MHz})$	B _{3dB} = 40 MHz	
CERFIL	SMD 3.45x3.1	SFECF10	DM7HA00	B _{3dB} = 180 kHz	ceramic filter from Murata, or equivalent part
CERDIS	SMD 4.5x2	CDSCB10	M7GA135		ceramic discriminator from Murata, or equivalent part



2.1.3 Component Arrangement Top Side for FSK Reception

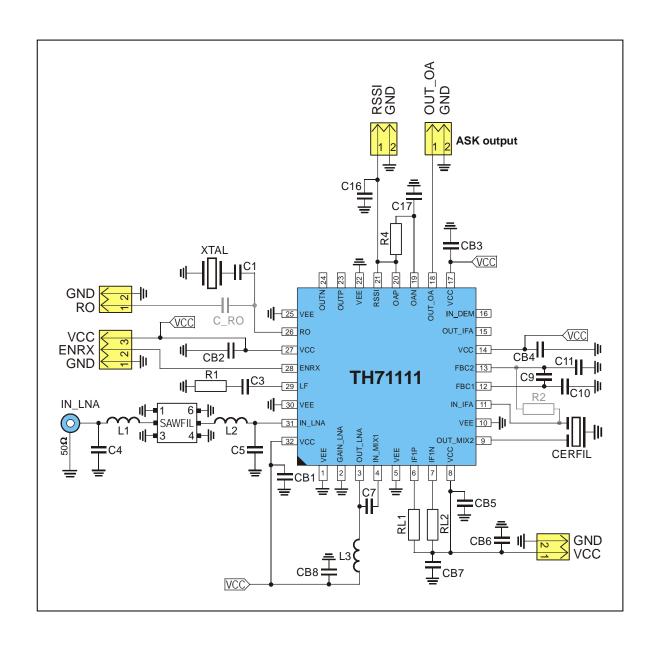
Board size is 42.7mm x 37.5mm





2.2 ASK Application Circuit

2.2.1 Circuit Diagram for ASK Reception





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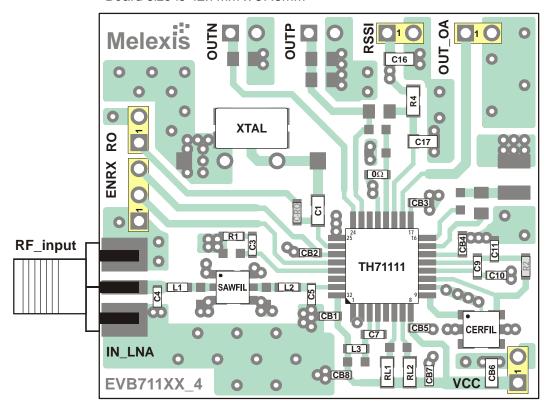
2.2.2 Board Component Values for ASK

Part	Size	Value @ 868.3 MHz	Value @ 915 MHz	Tolerance	Description
C1	0805	22 pF	22 pF	±5%	crystal series capacitor
C3	0603	1 nF	1 nF	±10%	loop filter capacitor
C4	0603	4.7 pF	NIP	±5%	capacitor to match SAW filter input
C5	0603	2.7 pF	NIP	±5%	capacitor to match SAW filter output
C7	0603	1.0 pF	0.68 pF	±5%	MIX1 input matching capacitor
C9	0603	33 nF	33 nF	±10%	IFA feedback capacitor
C10	0603	1 nF	1 nF	±10%	IFA feedback capacitor
C11	0603	1 nF	1 nF	±10%	IFA feedback capacitor
C16	0805	1.5 nF	1.5 nF	±10%	RSSI output low-pass capacitor, this value for data rates < 10 kbps NRZ, for higher data rates decrease the value
C17	0805	10 nF	10 nF	±10%	data slicer capacitor, this value for data rates > 0.8 kbps NRZ, for lower data rates increase the value
CB1 to CB5 CB7 to CB8	0603	330 pF	330 pF	±10%	de-coupling capacitor
CB6	0805	33 nF	33 nF	±10%	de-coupling capacitor
C_RO	0603	330 pF	330 pF	±5%	optional capacitor, to couple external RO signal
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R4	0805	330 kΩ	330 kΩ	±5%	data slicer resistor
RL1	0805	470 Ω	470 Ω	±5%	MIX1 bias resistor
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	HC49 SMD	© 101 = 000.0 WHZ	© 101 - 515 WH12	±зорріп ієтр.	fundamental-mode crystal, C_{load} = 10 pF to 15 pF, $C_{0, max}$ = 7 pF, $R_{1, max}$ = 50 Ωl
SAWFIL	SMD 3x3	SAFCC868MSL0X00 (f ₀ =868.3 MHz)		B _{3dB} = 2 MHz	low-loss SAW filters from Murata or equivalent part
SAWFIL	SMD 3.8x3.8		SAFCH915MAL0N00 $(f_0 = 915 \text{ MHz})$	B _{3dB} = 40 MHz	
CERFIL	SMD 3.45x3.1	SFECF10)M7HA00	B _{3dB} = 180 kHz	ceramic filter from Murata, or equivalent part



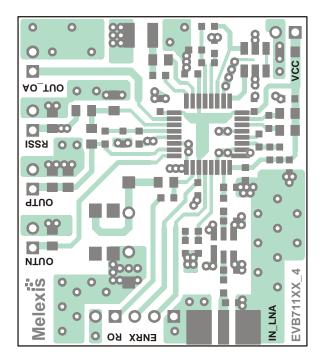
2.2.3 Component Arrangement Top Side for ASK Reception

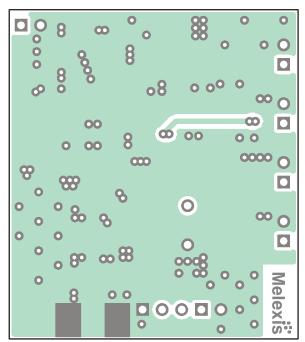
Board size is 42.7mm x 37.5mm



3 Evaluation Board Layouts

Board layout data in Gerber format is available, board size is 37.5mm x 42.7mm.





PCB top view

PCB bottom view



4 Package Description



The device TH71111 is RoHS compliant.

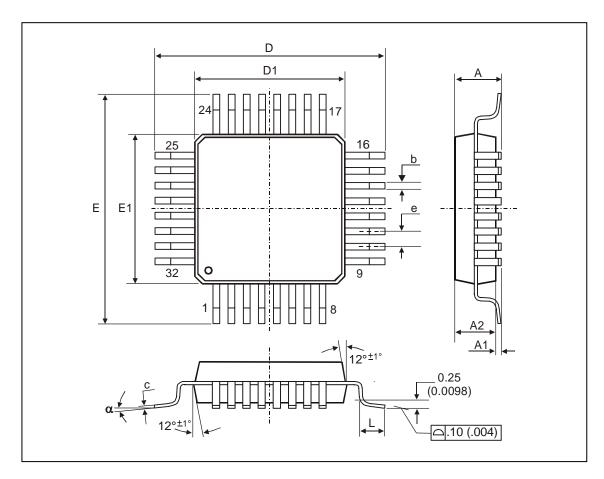


Fig. 2: LQFP32 (Low profile Quad Flat Package)

All Dim	All Dimension in mm, coplanaríty < 0.1mm										
	E1, D1	E, D	Α	A1	A2	е	b	С	L	α	
min	7.00	9.00	1.40	0.05	1.35	0.8	0.30	0.09	0.45	0°	
max	7.00	9.00	1.60	0.15	1.45	0.6	0.45	0.20	0.75	7°	
All Dim	All Dimension in inch, coplanaríty < 0.004"										
min	0.276	0.354	0.055	0.002	0.053	0.031	0.012	0.0035	0.018	0°	
max	0.270	0.554	0.063	0.006	0.057	0.031	0.018	0.0079	0.030	7°	

4.1 Soldering Information

 The device TH71111 is qualified for MSL3 with soldering peak temperature 260 deg C according to JEDEC J-STD-2.



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